

SKRIPSI

PERENCANAAN DAN PEMBUATAN MASTER KONTROL LAB. BAHASA DELAPAN CHANNEL BERBASIS MIKROKONTROLLER AT89S8252



Disusun Oleh :

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NIM : 01.17.044**



**KONSENTRASI TEKNIK ELEKTRONIKA
JURUSAN TEKNIK ELEKTRO S-1
FAKULTAS TEKNOLOGI INDUSTRI
INSTITUT TEKNOLOGI NASIONAL
MALANG
2007**

LEMBAR PERSETUJUAN



PERENCANAAN DAN PEMBUATAN MASTER KONTROL LAB. BAHASA DELAPAN CHANNEL BERBASIS MIKROKONTROLLER AT89S8252

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Gelar Sarjana Teknik

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Alhamdulillahi Robbil Alamin, puji syukur hamba panjatkan kehadirat-Mu Ya Allah atas segala nikmat dan karunia-Mu. Sholawat serta salam semoga tercurah kepada Rasulullah SAW., keluarga, sahabat, dan para pengikutnya. Amin. Atas kehendak Allah SWT. sajalah kami dapat menyelesaikan skripsi yang berjudul **"PERENCANAAN DAN PEMBUATAN MASTER KONTROL LAB. BAHASA DELAPAN CHANNEL BERBASIS MIKROKONTROLLER AT89S8252"** ini dengan lancar. Skripsi ini merupakan persyaratan kelulusan studi di jurusan Teknik Elektro S-1 konsentrasi Elektronika ITN Malang dan untuk mencapai gelar Sarjana Teknik.

Keberhasilan penyelesaian laporan Skripsi ini tidak lepas dari dukungan dan bantuan berbagai pihak. Untuk itu penyusun menyampaikan terimakasih kepada :

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2. Bapak Prof. Dr. Ir. Abraham Lomi, MSEE. selaku Rektor ITN Malang.
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5. Teman-temanku, terimakasih atas dukungannya selama ini.
6. Berbagai pihak yang tidak bisa disebutkan satu persatu, yang telah banyak membantu hingga selesaiya laporan ini.

Penyusun telah berusaha semaksimal mungkin dan menyadari sepenuhnya akan keterbatasan pengetahuan dalam menyelesaikan laporan ini. Untuk itu penyusun mengharapkan saran dan kritik yang membangun dari pembaca demi kesempurnaan laporan ini.

Harapan penyusun, semoga laporan Skripsi ini memberikan manfaat bagi perkembangan ilmu pengetahuan dan pembaca.

Malang, March 2007
Penyusun

Isnaini Rozafi



Laboratorium Bahasa sekarang ini banyak dibutuhkan oleh para pengajar, khususnya pada bidang bahasa, contohnya pada tempat kursusan, sekolah, lembaga penelitian, kampus, kantor dan lain sebagainya. Maka dengan adanya latar belakang itu, maka penulis mencoba membuatnya dengan judul Perencanaan Dan Pembuatan Master Kontrol Lab. Bahasa Delapan Chanel Berbasis Mikrokontroller AT89S8252, disamping sebagai pembelajaran, penulis juga ingin cepat lulus kuliah karena pengajuan skripsi sebagai prasyarat lulus sarjana.

1.2 Rumusan Masalah

Permasalahan yang dihadapi saat ini ialah bagaimana cara membuat dan merancang alat Laboratorium Bahasa yang berbasis mikrokontroller AT89S8252, yang meliputi cara kerja sistem, perancangan *hardware* dan *software*-nya.

Dari perumusan masalah didapat beberapa rumusan masalah, antara lain :

- Bagaimana membuat atau merancang cara kerja sistem dari Laboratorium Bahasa.
- Bagaimana cara memanfaatkan mikrokontroller sebagai pemroses cara kerja dari sistem yang kita rancang.
- Bagaimana cara memanfaatkan relay yang diatur oleh mikrokontroller sesuai dengan cara kerja yang kita inginkan.

1.3 Tujuan

Tujuan dari penggeraan skripsi ini ialah Perencanaan Dan Pembuatan Alat Master Kontrol Lab. Bahasa Delapan Chanel Berbasis Mikrokontroller AT89S8252 sesuai dengan cara kerja yang kita inginkan dengan memanfaatkan

mikrokontroller sebagai pengontrol alat yang akan dibuat dan dapat dimanfaatkan oleh masyarakat ataupun instansi terkait.

1.4 Batasan Masalah

Agar permasalahan dari pembuatan alat ini tidak meluas, maka perlu dibatasi pada hal-hal sebagai berikut :

- Pengontrol utama menggunakan Mikrokontroller AT89S8252.
- Tidak membahas catu daya.
- Tidak membahas audio amplifier.
- Hanya memakai sebuah rangkaian pengajar (Guru) dan delapan rangkaian murid.

1.5 Metodologi Penulisan

Dalam penulisan skripsi ini, langkah-langkah yang dilakukan sebagai berikut :

1. Studi Pustaka

Dimana penulis dengan membaca, mengumpulkan bahan-bahan materi yang digunakan sebagai referensi dan acuan dalam penyusunan skripsi

2. Studi Lapangan

Dalam hal ini, memperoleh data dengan cara praktik langsung dalam perencanaan alat.

3. Pengolahan Data

Mengolah data dengan jalan membuat analisa dan menarik kesimpulan dari hasil data yang didapat.

4. Pengujian Alat

Menguji alat yang telah dibuat apakah sesuai dengan perencanaan atau belum

5. Penyusunan Buku

Menyusun laporan hasil dari pelaksanaan skripsi.

1.6 Sistematika Penulisan

Sistematika penulisan dan gambaran secukupnya yang terdapat dalam setiap bab sebagai berikut :

BAB I : PENDAHULUAN

Membahas tentang latar belakang, rumusan masalah, tujuan, batasan masalah, metodologi perencanaan dan sistematika penulisan.

BAB II : LANDASAN TEORI

Membahas teori beserta pustaka yang relevan dan menunjang perancangan dan pembuatan rangkaian yang digunakan.

BAB III : PERENCANAAN DAN PEMBUATAN ALAT

Membahas perancangan dan pembuatan perangkat keras dan perangkat lunak.

BAB IV : PENGUJIAN ALAT

Membahas tentang pengujian-pengujian terhadap sistem yang telah dibuat.

BAB V : PENUTUP

Merupakan bagian-bagian penutup yang berisi kesimpulan dan saran yang bermanfaat bagi pengembangan lebih lanjut.

BAB II

LANDASAN TEORI

Landasan teori ini sangat membantu untuk dapat memahami suatu sistem. Disamping itu dapat juga dijadikan sebagai bahan acuan didalam merencanakan suatu sistem. Dengan pertimbangan hal-hal tersebut maka landasan teori merupakan bagian yang harus dipahami untuk pembahasan selanjutnya.

2.1 Mikrokontroler AT89S8252

Mikrokontroler AT89S8252 merupakan mikrokontroler 8 bit kompatibel dengan Standar industri MCS-51TM baik atas segi pemrograman maupun kaki tiap pin. Mikrokontroler AT89S8252 menpunyai 8 Kbyte (*Flash Programmable and Read Only Memori*) pada dasarnya mikrokontroler adalah terdiri atas mikroprosesor, *timer*, dan *counter*, perangkat I/O dan internal memori. Mikrokontroler termasuk perangkat yang sudah didesain dalam *chip* tunggal.

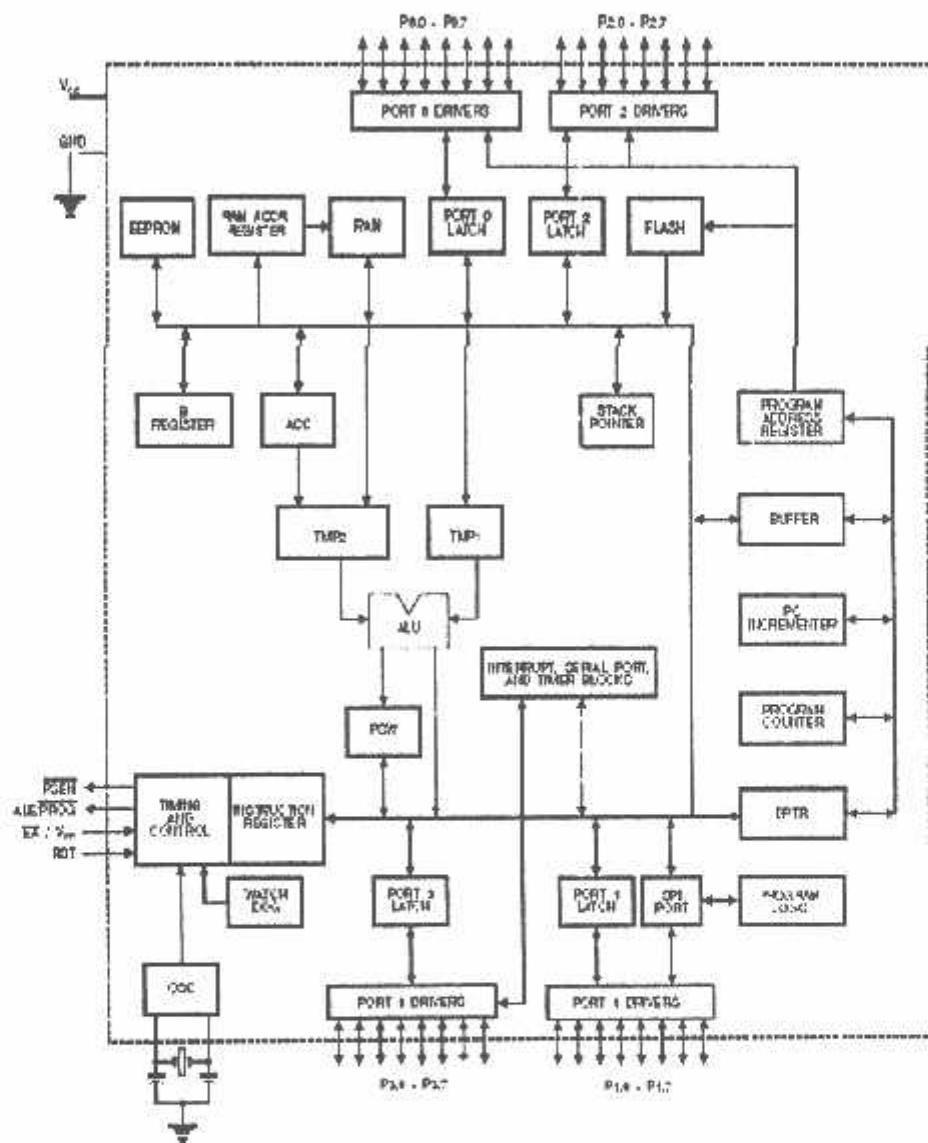
Pada dasarnya mikrokontroler mempunyai fungsi yang sama dengan mikroprosesor yaitu untuk mengontrol suatu kerja sistem. Selain itu mikrokontroler juga dikemas dalam satu *chip* (*single chip*). Didalam mikrokontroler juga terdapat CPU, ALU, PC, SP, dan register seperti dalam mikroprosesor, tetapi juga ditambah dengan perangkat-perangkat lain seperti ROM, RAM, PIO, SIO, *counter* dan sebuah rangkaian *clock*. Mikroprosesor didesain dengan instruksi-instruksi lebih luas dan 8 bit instruksi yang digunakan membaca data instruksi dari internal memori ke ALU. Sebagai suatu sistem control mikrokontroler bila dibandingkan dengan mikroprosesor memiliki kemampuan dan segi ekonomis yang bisa diandalkan karena dalam

mikrokontroler sudah terdapat RAM dan ROM. Sedangkan mikroprosesor didalamnya tidak terdapat keduanya. Terlihat bahwa mikrokontroler Atmel AT89S8252 memiliki banyak fitur yang menguntungkan. Dipakainya Downloadable flash memori memungkinkan mikrokontroler ini bekerja sendiri tanpa diperlukan tambahan *chip* lainnya. Sementara *Flash* memorinya mampu diprogram hingga seribu kali. Hal lain yang menguntungkan adalah system pemrograman menjadi lebih sederhana dan tidak memerlukan rangkaian yang rumit seperti rangkaian untuk memprogram produk Atmel lainnya. Secara umum konfigurasi yang dimiliki mikrokontroler AT89S8252 adalah sebagai berikut :

- Sebuah CPU 8 bit dengan menggunakan teknologi dari Atmel.
- 8K byte Downloadable Flash Memori.
- 2K byte EEPROM
- Sebuah *port* serial dengan control *full duplex* UART (Universal Asynchronous Receiver Transmpter).
- 256 byte RAM internal.
- 32 I/O yang dapat dipakai semua.
- 3 buah Timer/Counter 16 bit .
- SPI Serial Interface.
- Programmable Watchdog Timer .
- Dual Data Pointer.
- Frekuensi kerja 0 sampai 24 MHz
- Tegangan operasi 2,7 Volt sampai 6Volt.

- Kemampuan melaksanakan operasi perkalian, pembagian, dan operasi Boolean (bit)

Sedangkan untuk blok diagram AT89S8252 diperlihatkan dalam gambar 2-1



Gambar 2-1 Blok Diagram AT89S8252

Sumber : Data Sheet Mikrokontroller ATMEL AT89S8252

2.1.1. Penjelasan Fungsi Pin AT89S8252

Mikrokontroler AT89S8252 mempunyai 40 pin seperti yang ditunjukkan dalam gambar 2-2 Fungsi-fungsi pin dijelaskan sebagai berikut :

(T2) P1.0	1	40	VCC
(T2 EX) P1.1	2	39	P0.0 (AD0)
P1.2	3	38	P0.1 (AD1)
P1.3	4	37	P0.2 (AD2)
(SS) P1.4	5	36	P0.3 (AD3)
(MOSI) P1.5	6	35	P0.4 (AD4)
(MISO) P1.6	7	34	P0.5 (AD5)
(SCK) P1.7	8	33	P0.6 (AD6)
RST	9	32	P0.7 (AD7)
(RXD) P3.0	10	31	EA/VPP
(TXD) P3.1	11	30	ALE/PROG
(INT0) P3.2	12	29	PSEN
(INT1) P3.3	13	28	P2.7 (A15)
(T0) P3.4	14	27	P2.6 (A14)
(T1) P3.5	15	26	P2.5 (A13)
(WR) P3.6	16	25	P2.4 (A12)
(RD) P3.7	17	24	P2.3 (A11)
XTAL2	18	23	P2.2 (A10)
XTAL1	19	22	P2.1 (A9)
GND	20	21	P2.0 (A8)

Gambar 2-2 Susunan Pin AT89S8252

Sumber : Data Sheet Mikrokontroller ATMEL AT89S8252

➤ Pin 1 sampai 8

Port 1 yang terdiri atas pin 1 sampai 8 merupakan saluran masukan/keluaran dua arah.

➤ Pin 9

RST merupakan saluran dua masukan untuk mereset mikrokontroler dengan cara memberi masukan logika tinggi.

➤ Pin 10 sampai 17

Port 3 yang terdiri atas pin 10 sampai pin 17 merupakan saluran masukan/keluaran dua arah dan mempunyai fungsi khusus seperti yang terlihat dalam tabel 2-1

➤ Pin 18 dan 19

XTAL₁ dan *XTAL₂* merupakan saluran untuk mengatur pewaktuan system. Untuk pewaktuan dapat menggunakan pewaktuan internal maupun eksternal .

➤ Pin 20

V_{ss} merupakan hubungan ke *ground* dari rangkaian.

➤ Pin 21 sampai 28

Port 2 yang terdiri atas pin 21 sampai 28 merupakan saluran masukan/keluaran dua arah. *Port* ini mengeluarkan 8 bit bagian alamat tinggi (A8-A15) selama pengambilan instruksi dari memori program eksternal dan pengambilan data memori eksternal yang menggunakan mode pengalamanan 16 bit.

➤ Pin 29

PSEN (Program Store Enable) merupakan sinyal baca untuk mengaktifkan memori program eksternal.

➤ Pin 30

ALE/PROG (Address Latch Enable) merupakan pulsa yang bersfungsi untuk menahan alamat rendah (A0-A7) dalam *port 0*, selama proses baca/tulis memori eksternal. Frekuensi ALE adalah 1/6 kali frekuensi

osilator, dan digunakan sebagai perekam. Pin ini juga berfungsi sebagai saluran program selama dilakukan pemrograman jika menggunakan memori program eksternal.

➤ **Pin 31**

EA/VPP (*External Access Enable*) untuk mengatur penggunaan memori program eksternal dan internal. Pin ini harus dihubungkan dengan *ground* bila menggunakan memori program eksternal dan dihubungkan dengan VPP sebesar 12 Volt jika menggunakan memori program eksternal.

➤ **Pin 32 sampai 39**

Port 0 yang terdiri atas pin 32 sampai 39 merupakan saluran masukan/keluaran. *Port 0* merupakan saluran alamat rendah (A0-A7) yang dimultiplex dengan saluran bus data (D0-D7).

➤ **Pin 40**

Vcc merupakan saluran masukan untuk catu daya positif sebesar 5 Volt DC dengan toleransi kurang lebih 10 %.

2.1.2. Masukan dan Keluaran

Untuk masukan dan keluaran terdapat 4 buah *port* yang masing-masing 8 bit. Saluran ini bersifat dua arah (*bidirectional*) yang berarti dapat difungsikan sebagai masukan atau keluaran, serta dapat dialamat per bit. *Port 3* selain digunakan sebagai *port* masukan dan keluaran juga dapat digunakan sebagai fungsi pengganti sebagaimana yang terdapat dalam tabel 2-1.

Sedangkan AT89S8252 memiliki fitur tambahan yang terdapat pada *port 1* seperti pada tabel 2-2.

Tabel 2-1 Fungsi Pengganti Port 3

Port Pin	Alternate Functions
P3.0	RXD (serial input port)
P3.1	TXD (serial output port)
P3.2	INT0 (external interrupt 0)
P3.3	INT1 (external interrupt 1)
P3.4	T0 (timer 0 external input)
P3.5	T1 (timer 1 external input)
P3.6	WR (external data memory write strobe)
P3.7	RD (external data memory read strobe)

Sumber : Data Sheet Mikrokontroller ATMEL AT89S8252

Tabel 2-2 Fungsi khusus pada port 1 AT89S8252

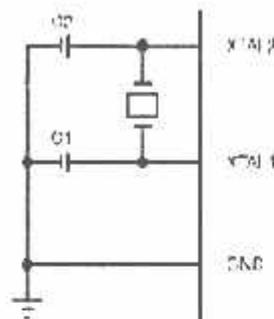
Port Pin	Alternate Functions
P1.0	T2 (external count input to Timer/Counter 2; clock-out)
P1.1	T2EX (Timer/Counter 2 capture/reload trigger and direction control)
P1.4	SS (Slave port select input)
P1.5	MOSI (Master data output, slave data input pin for SPI channel)
P1.6	MISO (Master data input, slave data output pin for SPI channel)
P1.7	SCK (Master clock output, slave clock input pin for SPI channel)

Sumber : Data Sheet Mikrokontroller ATMEL AT89S8252

2.1.3. Osilator

Jantung dari AT89S8252 adalah rangkaian yang membangkitkan pulsa clock yang mesinkronkan semua operasi internal. Mikrokontroler AT89S8252 memiliki osilator internal (*on chip oscillator*) yang dapat digunakan sebagai sumber waktu (*clock*) bagi CPU. Untuk menggunakan internal diperlukan

sebuah kristal atau resonator keramik antara pin XTAL1 dan pin XTAL2 dan sebuah kapasitor ke *ground*. Konfigurasinya dapat dilihat pada gambar berikut.

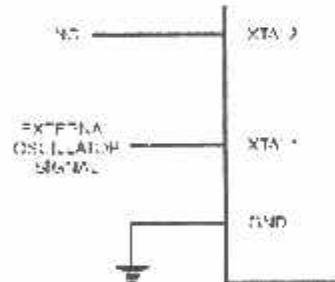


Gambar 2-3 Konfigurasi Osilator Menggunakan Kristal

Sumber : Data Sheet Mikrokontroller ATMEL AT89S8252

Nilai C1 dan C2 adalah 10 pF – 30 pF bila menggunakan kristal, dan bernilai 10 pF – 40 pF bila menggunakan resonator keramik.

Untuk penggunaan dengan external clock, XTAL2 harus dibiarkan dalam kondisi tidak terhubung. Konfigurasinya dapat dilihat pada gambar berikut ini :

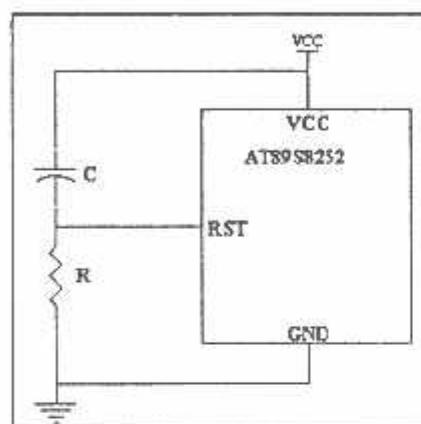


Gambar 2-4 Konfigurasi Osilator Menggunakan External Oscillator Signal

Sumber : Data Sheet Mikrokontroller ATMEL AT89S8252

2.1.4. Reset

Rangkaian *power on reset* diperlukan untuk mereset mikrokontroler secara otomatis setiap catu daya *on*. Gambar 2-5 menunjukkan rangkaian *power on reset*. Ketika catu daya diaktifkan, rangkaian *reset* menahan logika tinggi pin RST dengan jangka waktu yang ditentukan oleh besarnya pengisian muatan C.



Gambar 2-5 Rangkaian *Power On Reset*

Sumber : Data Sheet Mikrokontroller ATMEL AT89S8252

2.1.5. Data Memori (EEPROM) Dan RAM

Berbeda dengan mikrokontroler standard MCS-51, mikrokontroler Atmel AT89S8252 juga dilengkapi dengan data memori yang berupa EEPROM (*Electrically Erasable Programmable Read Only Memory*). EEPROM yang dimaksud ini besarnya *2 klo byte* (2K) dan dipakai untuk penyimpanan data.

EEPROM *on-chip* ini diakses dengan mengeset bit EEMEN pada register WMCON pada alamat 96H. Alamat EEPROM ini adalah 000H sampai 7FFH. Instruksi move digunakan untuk mengakses EEPROM internal ini. Bit EEMWE pada register WMCON harus diset ke-1 sebelum sebarang lokasi pada EEPROM dapat ditulisi. Program pengguna harus *mereset* bit EEMWE ke '0' jika proses

ABSTRAKSI

PERENCANAAN DAN PEMBUATAN MASTER KONTROL

LAB. BAHASA DELAPAN CHANNEL BERBASIS MIKROKONTROLLER AT89S8252

(Nama : Isnaini Rozafi, NIM : 01.17.044, Teknik Elektronika S1)

(Dosen Pembimbing : Joseph Dedi Irawan ST., MT.)

Pada perkembangan teknologi saat ini pada bidang pendidikan, sekarang ini sedang tren adanya Laboratorium Bahasa guna pembelajaran-pembelajaran bahasa, baik bahasa asing ataupun bahasa dalam negeri, maka dengan adanya Laboratorium Bahasa, seorang siswa atau pemakai dapat belajar lebih mudah dan fleksibel, guna memudahkan para pengajar dalam hal penyampaian pelajarannya.

Selain itu, dengan adanya Laboratorium Bahasa, suatu instansi baik instansi negeri ataupun instansi swasta dapat menambah bobot (pandangan atau gengsi), sehingga dengan adanya Laboratorium Bahasa diharapkan dapat menarik minat masyarakat, khususnya calon siswa.

Laboratorium Bahasa sekarang ini banyak dibutuhkan oleh para pengajar, khususnya pada bidang bahasa, seperti pada tempat kursusan, sekolah-an, lembaga penelitian, kampus, kantor dan lain sebagainya. Maka dengan adanya latar belakang itu, maka penulis mencoba untuk mengajukannya dengan judul Perencanaan Dan Pembuatan Master Kontrol Lab. Bahasa Delapan Channel Berbasis Mikrokontroller AT89S8252, disamping sebagai pembelajaran, penulis juga ingin cepat lulus kuliah karena pengajuan skripsi sebagai prasyarat lulus sarjana.

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DAFTAR PUSTAKA

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- Wasito S. Vademekum Elektronika Edisi Kedua, Penerbit PT. Gramedia Pustaka Utama, Jakarta, 2004
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LAMPIRAN

BAB V

PENUTUP

5.1 Kesimpulan

Dari percobaan dan pengujian, didapatkan kesimpulan sebagai berikut :

- Penggunaan relay pada rangkaian *audio* yang dekat dengan *audio amplifier* dapat menghasilkan suara dengung (*noise*).
- Pada sistem *audio* sebaiknya menggunakan kabel jenis *scarm* untuk input dan output rangkaian *audio* guna meredam noise akibat *fluksi* dari trafo.
- Untuk mendapatkan sistem *audio* yang bagus, sebaiknya sistem *ground* terhubung dengan baik pada bodi dan rangkaian serta pada kabel *power* sebaiknya menggunakan kabel besar, guna mendapatkan nilai *resistansi* yang lebih kecil.

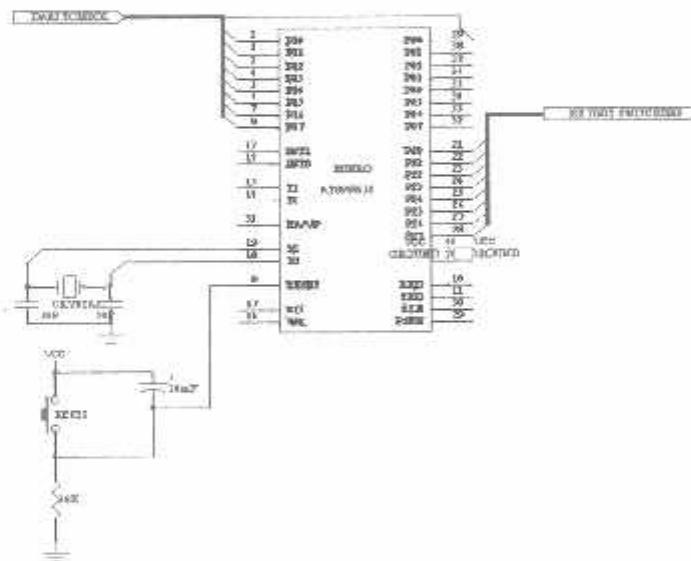
5.2. Saran-saran

1. Agar siswa dapat dengan mudah dalam menggunakannya, dapat ditambahkan suatu tampilan, seperti LCD ataupun dapat dihubungkan ke PC agar lebih canggih, mengikuti perkembangan zaman.
2. Guna pengembangan yang lebih lanjut, dapat ditambahkan beberapa fitur yang dapat membantu dalam pengoperasiannya seperti sistem kuis maupun pada ruang seminar agar lebih efisien.

3.5. Mikrokontroller AT89S8252

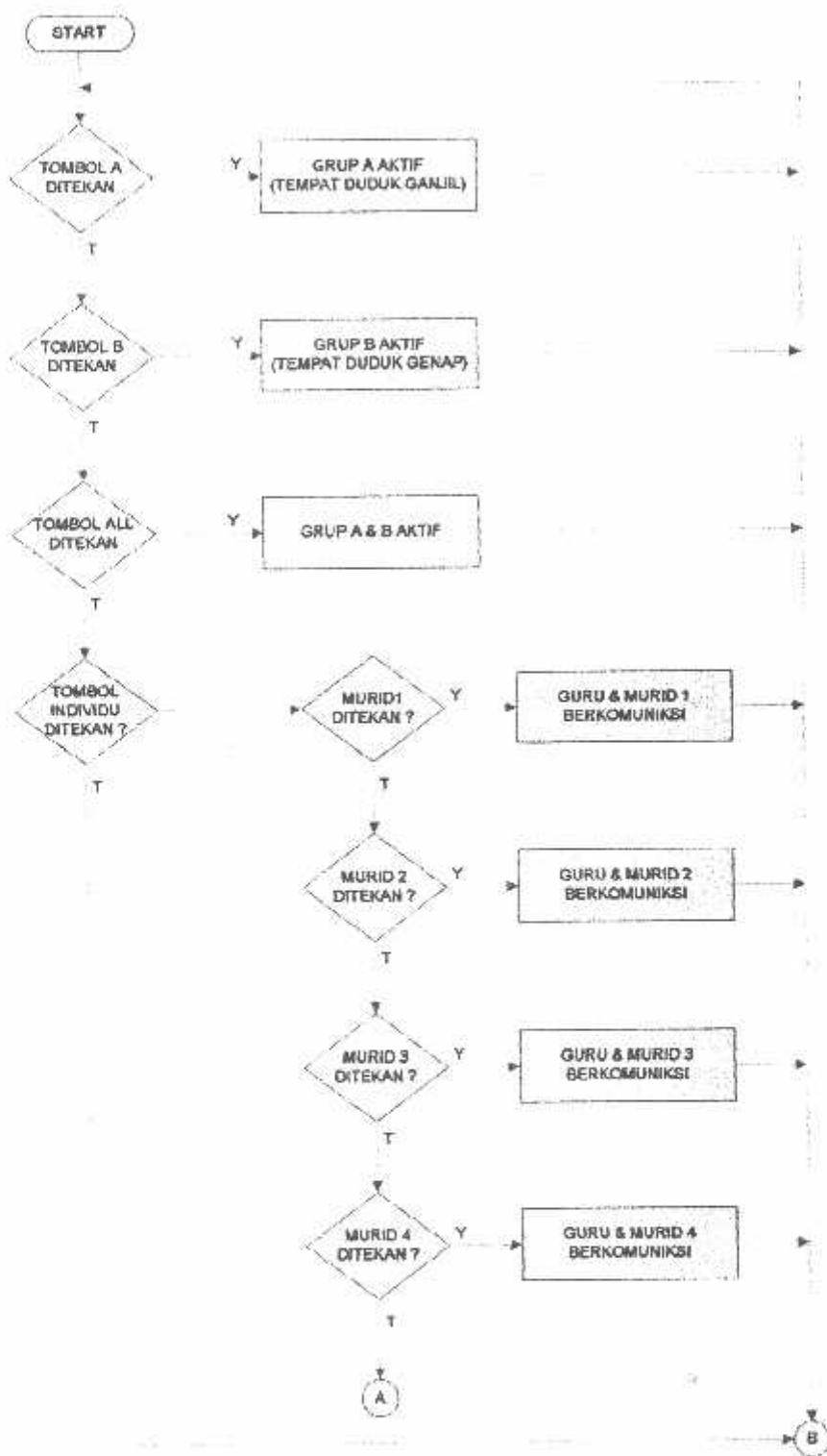
Disini rangkaian mikrokontroller berfungsi sebagai pengolah data dan pengendali alat agar sistem dapat bekerja dengan optimal, maka dalam perancangan alat ini pin-pin yang akan digunakan sebagai berikut :

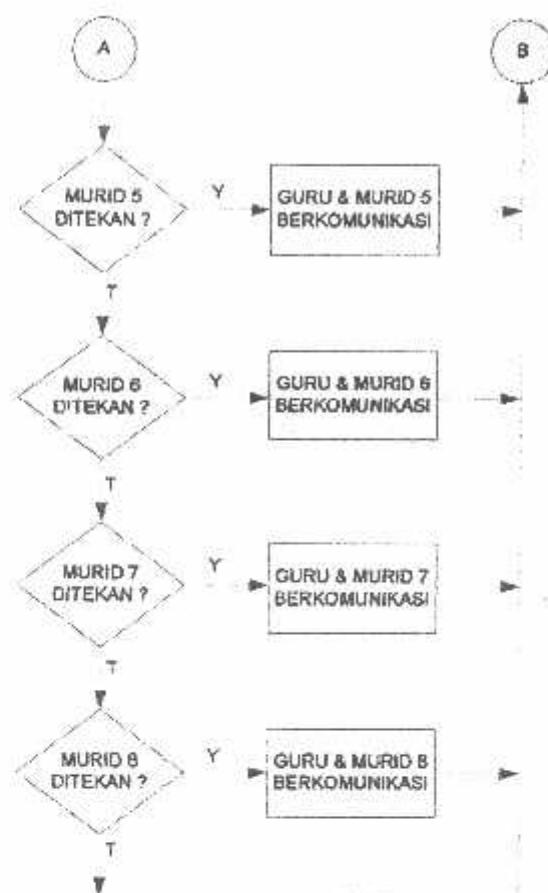
- Port 1.0-1.7 digunakan sebagai inputan dari tombol
- Port 0.0-0.7 digunakan sebagai outputan menuju relay
- Port 18 dan 19 sebagai rangkaian clock
- Port 9 sebagai reset



Gambar 3-4 Rangkaian Mikrokontroller

3.6. Flowchart





BAB IV

PENGUJIAN ALAT

4.1. Umum

Dalam bab ini akan dibahas tentang pengujian alat yang telah dibuat berdasarkan perencanaan dan pembuatan. Maksud dan tujuan pengujian alat ini adalah untuk mengetahui apakah alat yang telah direncanakan dan dibuat dapat bekerja dengan baik.

Pengujian alat ini meliputi pengujian terhadap :

1. Pengujian rangkaian pre-amp mic
2. Pengujian rangkaian mixer
3. pengujian rangkaian IC switch
4. Pengujian seluruh sistem.

4.2. Pengujian Rangkaian Pre-amp mic.

Tujuan pengujian pre-amp mic adalah untuk mengetahui besarnya penguatan dengan cara mengukur Vout dari pre-amp mic. Penguatan telah ditentukan sebesar 50 kali.

Peralatan yang digunakan antara lain catu daya, multimeter digital dan variabel resistor. Pengujian penguatan dapat dilakukan dengan menggunakan multimeter digital yang berfungsi untuk mengetahui besarnya nilai Vout dari pre-amp

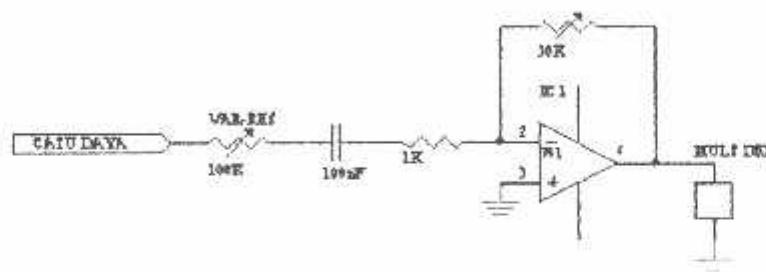
mic. Dengan menentukan besarnya penguatan sebesar 50 kali, maka R_f dapat dicari dengan rumus :

$$V_{out} = A_v \times V_{in}$$

$$A_v = R_f/R_1$$

$$50 = R_f/1K$$

Bila ditentukan $R_1 = 1K$, maka $R_f = 50K$



Gambar 4.1.

Pengujian rangkaian pre-amp mic.

Langkah-langkah pengujian pre-amp mic dengan menggunakan multimeter digital adalah sebagai berikut :

1. menyiapkan rangkaian seperti gambar, serta memastikan bahwa catudaya telah terpasang dan hubungan rangkaian telah benar.
2. Setelah susunan dan hubungan antar rangkaian telah benar, aktifkan sumber tegangan
3. Gunakan multimeter digital untuk mengetahui tegangan pada V_{out} pre-amp mic.

Tabel 4-1

Hasil Pengukuran Pre-Amp Mic

V _{in} (mV)	V _{out} (V)	A _V (kali)
20,45	0,95	46,89
31,66	1,50	47,51
40,80	2,03	49,79
52,30	2,57	49,32
61,68	3,07	49,83

Dari pengujian didapat hasil seperti dalam tabel 4-1, dari data telah ditentukan besarnya penguatan sebesar 50 kali, maka untuk melihat prosentase kesalahan dari penguatan digunakan rumus :

$$\text{Error} = \frac{\text{perencanaan} - \text{pengukuran}}{\text{perencanaan}} \times 100\%$$

$$\text{Error1} = \frac{50 - 46,89}{50} \times 100\%$$

$$= 6,22 \%$$

$$\text{Error2} = \frac{50 - 47,51}{50} \times 100\%$$

$$= 4,98 \%$$

$$Error\ 3 = \frac{50 - 49,79}{50} \times 100\%$$

$$= 0,42\%$$

$$Error\ 4 = \frac{50 - 49,32}{50} \times 100\%$$

$$= 1,36\%$$

$$Error\ 5 = \frac{50 - 49,83}{50} \times 100\%$$

$$= 0,34\%$$

Dari data diatas didapatkan error rata-rata sebesar :

$$\frac{6,22 + 4,98 + 0,42 + 1,36 + 0,34}{5} = 2,66\%$$

dan penguatan rata-rata sebesar :

$$\frac{46,89 + 47,51 + 49,79 + 49,32 + 49,83}{5} = 48,66 \text{ kali}$$

4.3. Pengujian rangkaian mixer

Tujuan pengujian rangkaian mixer adalah untuk mengetahui besarnya penjumlahan dengan mengukur V_{out} dari mixer. Penjumlahan ditentukan untuk penguatan yang pertama sebesar 1 kali dan untuk penguatan yang kedua sebesar 5 kali.

Peralatan yang digunakan antara lain catu daya, variable resistor dan multimeter digital. Pengujian penguatan dapat dilakukan dengan cara menggunakan multimeter digital untuk mengetahui nilai V_{out} dari mixer. Dengan menentukan

besarnya penguatan pertama sebesar 1 kali dan penguatan yang kedua sebesar 5 kali, maka untuk penguatan yang pertama digunakan rumus :

$$V_{out1} = - \frac{V_{in1}}{R_{in1}} \times R_{f1}$$

$$\Delta V = - \frac{R_{f1}}{R_{in1}}$$

$$= \frac{V_{out1}}{V_{in1}}$$

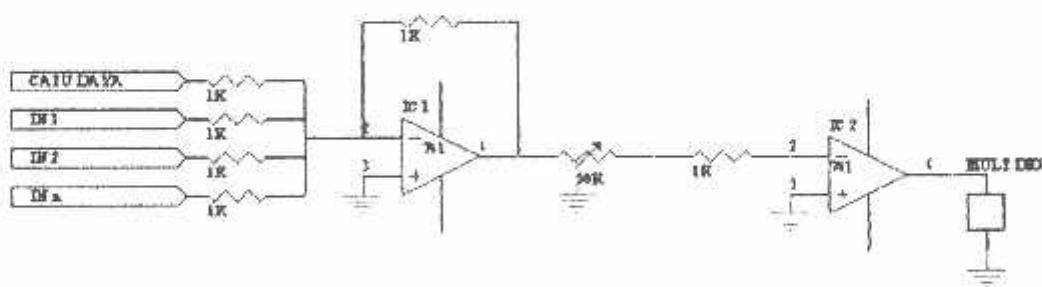
Bila ditentukan $R1 = 1K$, maka $Rf = 1K$

Sedangkan untuk penguatan yang kedua sebesar 5 kali menggunakan rumus :

$$V_{out2} = - \frac{V_{in2}}{R_{in2}} \times R_{f2}$$

$$\Delta V = - \frac{R_{f2}}{R_{in2}}$$

$$= \frac{V_{out1}}{V_{in1}}$$



Gambar 4.2.
Pengujian Rangkaian Mixer.

Tabel 4-2.
Hasil pengukuran rangkain mixer

Vin (mV)	Vout (mV)	AV (kali)
21,24	61,38	2,89
30,90	90,53	2,93
39,68	114,27	2,88
50,41	146,69	2,91
59,96	175,08	2,92

Dari pengujian didapat hasil seperti dalam tabel 4-2, dari data telah ditentukan besarnya penguatan sebesar 3 kali, maka untuk melihat prosentase kesalahan dari penguatan digunakan rumus :

$$Error = \frac{perencanaan - pengukuran}{perencanaan} \times 100\%$$

$$Error1 = \frac{3 - 2,89}{3} \times 100\%$$

$$= 3,6 \%$$

$$Error2 = \frac{3 - 2,93}{3} \times 100\%$$

$$= 2,3 \%$$

$$Error3 = \frac{3 - 2,88}{3} \times 100\%$$

$$= 4 \%$$

$$Error4 = \frac{3 - 2,91}{3} \times 100\%$$

$$= 3 \%$$

$$Error5 = \frac{3 - 2,92}{3} \times 100\%$$

$$= 2,6 \%$$

Dari data diatas didapatkan error rata-rata sebesar :

$$\frac{3,6 + 2,3 + 4 + 3 + 2,6}{5} = 3,10\%$$

sedangkan penguatan rata-ratanya didapat :

$$\frac{2,89 + 2,93 + 2,88 + 2,91 + 2,92}{5} = 2,90 \text{ kali}$$

4.4. Pengujian seluruh sistem

Tujuan pengujian seluruh sistem adalah untuk mengetahui apakah sistem telah bekerja sesuai dengan apa yang diharapkan, langkah-langkah pengujian seluruh sistem adalah dengan mempersiapkan semua rangkaian, kemudian melihat hasil keluaran yang ditampilkan melalui lampu led maupun output audionya.

Tabel 4-3

Hasil pengujian tombol individu saat menu individu ON

IND	Kondisi Led							
	1	2	3	4	5	6	7	8
1	ON	OFF						
2	OFF	ON	OFF	OFF	OFF	OFF	OFF	OFF
3	OFF	OFF	ON	OFF	OFF	OFF	OFF	OFF
4	OFF	OFF	OFF	ON	OFF	OFF	OFF	OFF
5	OFF	OFF	OFF	OFF	ON	OFF	OFF	OFF
6	OFF	OFF	OFF	OFF	OFF	ON	OFF	OFF
7	OFF	OFF	OFF	OFF	OFF	OFF	ON	OFF
8	OFF	OFF	OFF	OFF	OFF	OFF	OFF	ON

Tabel 4-4
Hasil pengujian tombol CALL

IND	Kondisi Led (berkedip)							
	1	2	3	4	5	6	7	8
1	ON	OFF						
2	OFF	ON	OFF	OFF	OFF	OFF	OFF	OFF
3	OFF	OFF	ON	OFF	OFF	OFF	OFF	OFF
4	OFF	OFF	OFF	ON	OFF	OFF	OFF	OFF
5	OFF	OFF	OFF	OFF	ON	OFF	OFF	OFF
6	OFF	OFF	OFF	OFF	OFF	ON	OFF	OFF
7	OFF	OFF	OFF	OFF	OFF	OFF	ON	OFF
8	OFF	OFF	OFF	OFF	OFF	OFF	OFF	ON

Tabel 4-5
Hasil pengujian tombol A (ganjil)*

A	Kondisi Audio							
	1	2	3	4	5	6	7	8
1	ON	OFF						
2	OFF	OFF	OFF	OFF	OFF	OFF	OFF	OFF
3	OFF	OFF	ON	OFF	OFF	OFF	OFF	OFF
4	OFF	OFF	OFF	OFF	OFF	OFF	OFF	OFF
5	OFF	OFF	OFF	OFF	ON	OFF	OFF	OFF
6	OFF	OFF	OFF	OFF	OFF	OFF	OFF	OFF
7	OFF	OFF	OFF	OFF	OFF	OFF	ON	OFF
8	OFF	OFF	OFF	OFF	OFF	OFF	OFF	OFF

Tabel 4-6
Hasil pengujian tombol B (genap)*

Kondisi Audio								
B	1	2	3	4	5	6	7	8
1	OFF							
2	OFF	ON	OFF	OFF	OFF	OFF	OFF	OFF
3	OFF							
4	OFF	OFF	OFF	ON	OFF	OFF	OFF	OFF
5	OFF							
6	OFF	OFF	OFF	OFF	OFF	ON	OFF	OFF
7	OFF							
8	OFF	ON						

*) ket : Untuk rangkaian tombol A dan B yang ON/OFF adalah sistem audionya, bukan masalah tentang lampu lcdnya.

Dari pengujian seluruh sistem dapat disimpulkan bahwa sistem telah bekerja sesuai dengan yang diharapkan.

instruksi tombol yang sudah ada.

- Tombol Murid, untuk komunikasi guru dengan murid sesuai dengan yang dapat berkomunikasi dengan guru.
 - Tombol B, hanya murid yang mempunyai nomor/tempat duduk genap saja yang dapat berkomunikasi dengan guru.
 - Tombol A, hanya murid yang mempunyai nomor/tempat duduk ganjil saja sesuai dengan keinginan sang guru.
 - Tombol Individu, guru dapat berkomunikasi dengan salah satu obyek.
 - Tombol All sebagai komunikasi antara guru dengan seluruh murid.
 - Tombol Power sebagai menghidupkan ataupun mematikan master kontrol.
- Ada beberapa tombol yang terdapat pada panel master kontrol, diantaranya :

instruksi

- ◆ Tombol Master Kontrol sebagai menjalankannya dengan memadamkan sertu ketemanagan dari blok-diagram sebagai berikut :

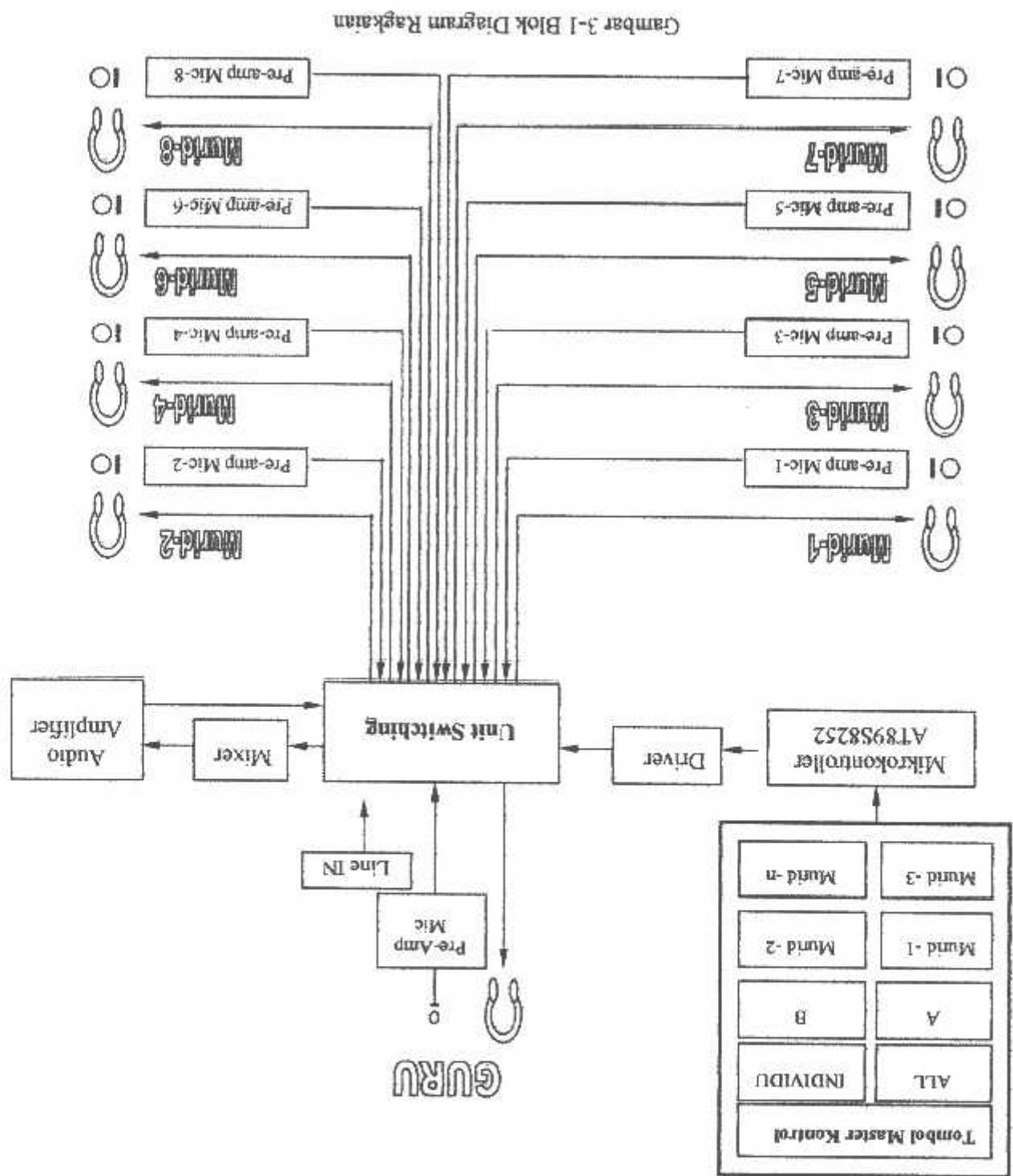
3.1 Sistematisik Secara Diagram Blok

Pembuatan perangkat lunak secara gratis besarnya meliputi: perencanaan dan pembuatan perangkat keras serta perencanaan dan direalisasikan sebagaimana fungsiya. Adapun perencanaan dan pembuatan alat Pada bagian dibahas mengenai peralatan yang ditencanakan dan akan

PERENCANAAN DAN PEMBUATAN ALAT

BAB III

- ♦ Mixer sebagai input dari tape recorder ataupun vcd.
- ♦ Audio Amplifier sebagai peningkat suara.
- ♦ Mixer sebagai pencampur audio dari gitar dan mud.
- debgan mud sesuai dengan instruksi.
- ♦ Unit Switching sebagai penghubung alatpuan pemutus audio antara gitar switching.
- ♦ Driver sebagai rangkaian penghubung antara mikrokontroler dengan unit mengendalikan output yang kita inginkan.
- memproses data inputan sesuai dengan instruksi yang masuk serta
- ♦ Mikrokontroler sebagai pengendali keseluruhan sistem, mikrokontroler akan



dapatlah diambilkan $A_V = 1$ kali dan R_i sebesar $1\text{ k}\Omega$, maka untuk mencari R_f yaitu :
 sebesar 3 kali. Dan bila R_f diambilkan $1\text{ k}\Omega$ maka R_f dapat dicari melalui rumus,
 penyelesaian op-amp yang pertama sebesar 1 kali dan untuk op-amp yang kedua
 diketahui denegan pengaruh inverting, dimana dalam rangkaian ini untuk
 mic. Rangkaian mixer menggunakan pengaruh op-amp jenis summing yang
 Mixer disini difungsikan sebagai pencampur semua masukan dari pre-amp

3.2 Mixer

$$= 50 \text{ K}$$

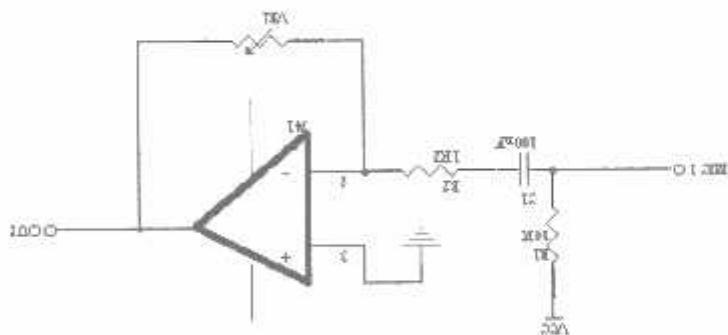
$$VR_1 = 50 \times 1\text{ k}$$

$$50 = -(VR_1 / 1\text{ k})$$

$$\text{menGGunakan rumus : } A_v = -(VR_1 / R_2)$$

Dari rangkaian dilatas diambilkan besaranya penyelesaian sebesar 50 kali dengan

Gambar 3-2 Rangkaian Pre-Amp Mic



dasar pre-amp mic adalah rangkaian pengaruh inverting seperti gambar di bawah ini
 Pre-Amp Mic dilumgsikan sebagai pengaruh bagi mic condenser, rangkaian

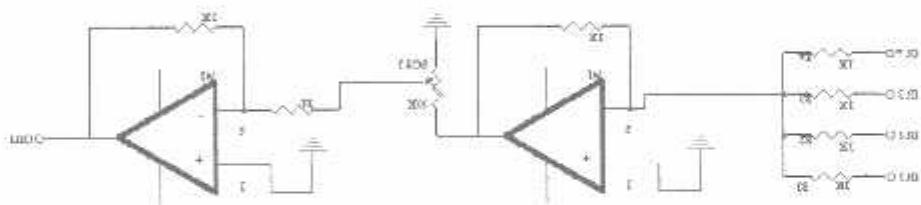
3.2 Rangkaian Pre-Amp Mic

BD 139 dengan datasheet sebagai berikut :

padah saat transistor dalam keadaan tidak aktif. Transistor yang digunakan adalah berfungsi untuk mengelelah timbulnya teganganan lebih pada transistor pengetima berfungsi untuk mengendalikan posisi relay, dioda yang diparalel dengan relay sebagai alat untuk mengendalikan rangkaian posisi relay. Sebelum masuk ke relay dipertukarkan rangkaian pengendali relay yang pada saatnya tidak aktif.

3.4. Driver Relay

Gambar 3-3 Rangkaian Mixer



Jadi untuk rangkaian mixer adalah sebagai berikut :

$$= 3\text{ k}\Omega$$

$$= 3 \times 1\text{k}$$

$$R_f = A_v \times R_i$$

Untuk mengetahui R_f adalah sebagai berikut :

Sedangkan untuk A_v diambil sebesar 3 kali dan R_i diambil 1kΩ, maka

$$= 1 \times 1\text{k}, \quad = 1\text{k}\Omega$$

$$R_f = A_v \times R_i$$

$$A_v = R_f/R_i$$

$$RB = \frac{Ib}{Vce - Vbe}$$

$$= \frac{0.33mA}{5 - 0.7}$$

$$= 13K$$

yaitu :

dengan demikian, harga RB dapat dicari jika nilai $Vce = 5$ Volt, $Vbe = 0,7$ Volt,

$$Ib = \frac{100}{hfe}$$

$$= 0.33mA$$

Maka pada saat jenuh (saturation) besar arus Id adalah :

$$Id = \frac{Vce}{Rt(\text{relay})}$$

$$= \frac{12}{360}$$

$$= 33.3mA$$

pada LC-nya, yaitu :

Jika relay dapat bekerja, transistor pada kondisi saturation, sehingga apabila dicari

♦ $VBE = 0,7$ Volt

♦ $hfe = 100$

♦ BD 139

BAB I

PENDAHULUAN

1.1 Latar Belakang

Perkembangan ilmu di bidang teknologi, khususnya teknologi elektronika merupakan suatu kenyataan yang menunjukkan bahwa manusia didalam melakukan aktifitasnya. Sesuai dengan tuntutan jaman yang semakin modern telah membawa ide-ide ke arah teknologi yang lebih maju, lebih bermanfaat serta memiliki nilai ekonomis.

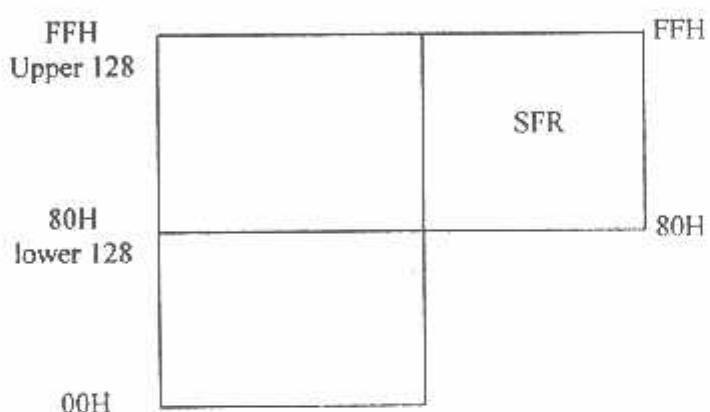
Oleh karena itu berbagai macam sebab perkembangan teknologi khususnya dibidang elektronika mempengaruhi pola kehidupan masyarakat yang dulunya memanfaatkan tenaga murni (tenaga manusia itu sendiri) kini berubah menggantikan suatu alat yang dapat bekerja sendiri yang hanya membutuhkan sedikit sentuhan tenaga manusia.

Seperti pada perkembangan teknologi saat ini pada bidang pendidikan, sekarang ini sedang tren adanya Laboratorium Bahasa guna pembelajaran-pembelajaran bahasa, seperti bahasa asing ataupun bahasa dalam negeri, sehingga dengan adanya Laboratorium Bahasa, seorang siswa atau pemakai dapat belajar lebih mudah dan fleksibel, guna memudahkan para pengajar dalam hal penyampaian pelajarannya.

Selain itu, dengan adanya Laboratorium Bahasa, suatu instansi baik instansi negeri ataupun instansi swasta, dapat menambah bobot (pandangan atau gengsi), maka dengan adanya Laboratorium Bahasa diharapkan dapat menarik minat masyarakat, khususnya calon siswa.

penulisan ke EEPROM tidak diperlukan lagi. Proses penulisan ke EEPROM dapat dilihat dengan membaca bit RDY/BSY pada SFR WMCON. Jika bit ini berlogika rendah maka berarti penulisan EEPROM sedang berlangsung, jika bit ini berlogika tinggi berarti penulisan sudah selesai dan penulisan lain dapat dimulai lagi.

Sedangkan RAM yang ada pada mikrokontroler AT89S8252 adalah berkapasitas 256 byte dan kompatibel dengan RAM yang ada pada mikrokontroler standard MCS-51.



Gambar 2-6. Memori Data Internal (RAM)

Sumber : Data Sheet Mikrokontroller ATMEIL AT89S8252

Pada *lower* 128 lokasi memori dibagi menjadi 3 bagian :

1. Register bank 0 – 3

Lokasi bank register dimulai dari alamat 00H – 1 H yang terdiri dari 32 bytes. Register bank ini terdiri dari 4 buah register 8 bit yang dapat dipilih melalui pengaturan *program status word* register.

2. Bit Addressing

Terdapat 16 bytes yang dimulai dari 20H – 2FH. Masing-masing dari 128 bit lokasi ini dapat dialamat secara langsung yaitu dari 00H sampai 7FH.

3. Scratch Pad Area

Lokasi dari alamat 30H – 7FH atau sebanyak 80 bytes yang dapat digunakan sebagai alamat bagi RAM.

2.1.6. Special Function Register (SFR)

Special Function Register merupakan register dengan tugas khusus. SFR pada mikrokontroler AT89S8252 kompatibel dengan mikrokontroler keluarga MCS-51 dan memiliki alamat 80H - FFH sehingga terdapat 128 lokasi alamat untuk SFR. Namun demikian pada mikrokontroler ini tidak berarti memiliki SFR sebanyak 128 buah. Berikut ini adalah gambar letak dari lokasi alamat SFR.

T2MOD (Timer 2 Mode dengan alamat 0C9H), WMCON (Watchdog and Memory Control Register dengan alamat 96H), SPCR (SPI Control Register dengan alamat D5H), SPSR (SPI Status Register dengan alamat AAH), SPDR (SPI Data Register dengan alamat 86H).

➤ SFR untuk Timer 2

Mikrokontroler AT89S8252 terdapat tambahan sebuah Timer/Counter yang diberi nama timer 2 (sehingga AT89S8252 memiliki 3 Timer/Counter yaitu Timer/Counter 0, Timer/Counter 1, Timer/Counter 2). Pada Timer/Counter 2 ini dikendalikan oleh Special Function Register yang bernama T2CON (Timer 2 Control), T2MOD (Timer @ MODE) dan sepasang register RCAP2H, RCAP2L merupakan register capture/reload untuk Timer 2 dalam 16 bit capture mode/auto reload mode.

➤ SFR untuk Watchdog Memori,

Untuk menggunakan Watchdog timer atau memori, maka dapat dilakukan dengan mengatur SFR yang bernama WMCON dengan alamat 96H.

➤ SFR pengontrol SPI

Berbeda dengan mikrokontroler MCS-51, AT89S8252 memiliki fasilitas SPI (Serial Peripheral Interface). Fasilitas ini memungkinkan transfer data kecepatan tinggi secara sinkron antara mikrokontroler dengan peripheral atau antar mikrokontroler AT89S8252. Fitur ini meliputi :

- a. Full Duplex, 3 kawat dengan transfer data secara sinkron ,
- b. Operasi Master atau Slave.
- c. Frekuensi maksimum 6 MHz.

d. 4 bit rate terprogram.

2.1.7. Timer dan Counter

Dalam mikrokontroler AT89S8252 terdapat tiga buah pewaktu/pencacah (timer/counter 16) 16 bit yang dapat diatur melalui perangkat lunak, yaitu pewaktu / pencacah 0 dan pewaktu / pencacah 1. Timer/counter ini diatur oleh *special function register* yaitu *Timer/Counter Control* (TCON alamat 88H), dan *Timer/Counter Mode Control* (TMOD alamat 89H). Selain itu nilai byte bawah dan byte atas dari Timer/Counter disimpan dalam register TI dan TH.

Jika difungsikan sebagai Timer, maka akan menggunakan system clock sebagai sumber masukan pulsanya. Jika sebagai Counter (pencacah), maka akan menggunakan pulsa dari luar (eksternal) sebagai masukan pulsanya. Pada Port 3 terdapat fungsi khusus yaitu TO (masukan luar untuk Timer/Counter 0) dan TI (masukan luar untuk Timer/Counter 1). Pemilihan mode Timer/Counter dikontrol oleh register TMOD. Dengan memberikan nilai tertentu pada register TMOD dapat dipilih mode operasi untuk Timer/Counter 0 dan Timer/Counter 1 seperti terlihat dalam Tabel 2-3.

Tabel 2-3 Mode Operasi Timer/Counter 0 dan 1

M2	M1	Mode	Keterangan
0	0	0	13 bit Timer
0	1	1	16 bit Timer
1	0	2	8 bit auto-reload
1	1	3	Split Mode

Pada mikrokontroler AT89S8252 terdapat tambahan Timer 2. Timer yang lain adalah Timer 0 dan Timer 1. Timer 2 ini merupakan Timer/Counter 16 bit dan memiliki 3 mode operasi yaitu *capture*, *auto reload (up down counting)* dan baud rate generator. Untuk memilih mode ini dilakukan dengan mengatur bit pada SFR T2CON (Timer 2 Control Register). Timer 2 ini terdiri dari 2 buah timer 8 bit register yaitu TH2 dan TL2. Pada fungsi Timer, register TL2 dinaikkan (increment) tiap siklus mesin. Karena siklus mesin terdiri dari 12 periode osilasi, maka count rate menjadi 1/12 dari frekuensi osilator. Sedangkan pada fungsi Counter, register dinaikkan berdasarkan tanggapan adanya transisi tinggi ke rendah pada pena yang bersesuaian (dalam hal ini pin T2 atau P1.0). Tabel berikut menunjukkan mode operasi yang dapat dijalankan pada timer 2.

Tabel 2-4 Mode Operasi Timer 2

RCLK + TCLK	CP/RL2	TR2	MODE
0	0	1	16-bit Auto-reload
0	1	1	16-bit Capture
1	X	1	Baud Rate Generator
X	X	0	(Off)

Sumber : Data Sheet Mikrokontroller ATMEL AT89S8252

RCLK = Receive clock enable. Jika diset menyebabkan serial port menggunakan pulsa overflow Timer 2 sebagai detak penerimaan pada serial port. Jika RCLK = 0 Timer 1 yang digunakan.

TCLK = Transmit clock enable. Jika diset menyebabkan serial port menggunakan pulsa overflow Timer 2 sebagai detak pengiriman. Jika TCLK = 0 pulsa overflow timer 1 yang digunakan.

CP/RL2 = Pemilihan capture/Reload. Jika diset maka proses capture yang terjadi sedangkan jika bit ini diclear maka proses reload.

TR2 = Bit untuk mengatur start/stop untuk timer 2 jika TR2 = 1 Timer akan aktif.

2.1.8. Idle Mode

Saat *Idle Mode* mikrokontroler tidak melakukan apa-apa. Tetapi peralatan lain yang terhubung tetap aktif. Kondisi ini dapat dihentikan dengan sebuah *interrupt* atau dengan *re-set* sistem.

2.1.9. Sistem Interupt

Mikrokontroler AT89S8252 mempunyai 6 buah sumber interrupt yang dapat membangkitkan permintaan interrupt, yaitu INTO, INT1, T0, T1, T2 dan port serial.

Saat terjadinya interupt, mikrokontroler secara otomatis akan menuju ke *sub rutin* pada alamat tersebut. Setelah interrupt service selesai dikerjakan, mikrokontroler akan mengerjakan program semula. Dua sumber interrupt *external* adalah INTO dan INT1, dimana kedua interupsi eksternal akan aktif atau aktif transisi tergantung isi dari IT0 dan IT1 pada register TCON. Interupsi T0, T1, T2 aktif pada saat timer yang sesuai mengalami *roll over*, interupsi serial dibangkitkan dengan melakukan operasi OR pada RI dan TI. Tiap-tiap sumber interupsi dapat *enable* atau *disable* secara otomatis.

Tingkat prioritas semua sumber interupsi dapat diprogram sendiri-sendiri dengan *set* atau *clear bit* pada SFRS IP (*interrupt Priority*).

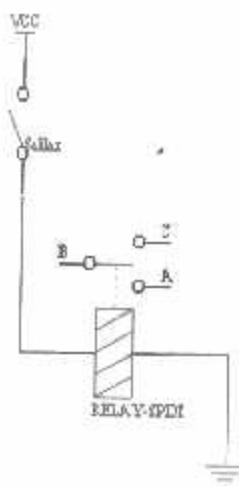
Tabel 2-5 Alamat Sumber Interupsi

(MSB(LSB)							
EA	-	ET2	ES	ET1	EX1	ET0	EX0
Enable Bit = 1 enables the interrupt.							
Enable Bit = 0 disables the interrupt.							
Symbol						Position	Function
EA						IE.7	Disables all interrupts. If EA = 0, no interrupt is acknowledged. If EA = 1, each interrupt source is individually enabled or disabled by setting or clearing its enable bit.
-						IE.6	Reserved.
ET2						IE.5	Timer 2 interrupt enable bit.
ES						IE.4	SPI and UART interrupt enable bit.
ET1						IE.3	Timer 1 interrupt enable bit.
EX1						IE.2	External interrupt 1 enable bit.
ET0						IE.1	Timer 0 interrupt enable bit.
EX0						IE.0	External interrupt 0 enable bit.

Sumber : WWW.datasheet.atmel.com

2.2. Relay

Relay adalah salah satu dari komponen elektronika yang terdiri dari sebuah lilitan kawat(kumparan/koil) yang terlilit pada sebuah besi lunak. Jika kumparan dialiri arus listrik maka inti besi akan menjadi magnet dan dapat menarik pegas sehingga kontak BA terhubung dan kontak BC terputus.



Gambar 2-7 Rangkaian Relay

Sumber : WWW.relay.com

Relay berfungsi untuk menghubungkan atau memutuskan aliran listrik yang dikontrol dengan memberikan tegangan dan arus tertentu pada koilnya. Dalam memutus atau menghubungkan kontak digerakkan oleh *fluksi* yang ditimbulkan dari adanya medan magnet listrik yang dihasilkan oleh kumparan yang melilit pada besi lunak.

2.3. Op-Amp.

2.3.1. Penguat Inverting

Rangkaian penguat inverting merupakan rangkaian Op-Amp yang paling luas digunakan, rangkaian ini merupakan penguat yang gainnya dari v_i ke v_o ditentukan oleh R_f dan R_i yang dapat memperkecil isyarat ac atau dc.

Tabel 2-6 Fungsi-fungsi pin-pin LM741

PIN	NAMA	KET
1.	BAL	OFFSET BALANCE
2.	- IN	INVERTING INPUT
3.	+ IN	NON INVERTING INPUT
4.	- V	GROUND
5.	BAL	OFFSET BALANCE
6.	OUT	OUTPUT SIGNAL
7.	+ V	POSITIF POWER SUPPLY
8.	NC	NO CONNECTION

Sumber : WWW.Datasheet.LM741.com

2.5. Switch Elektronik

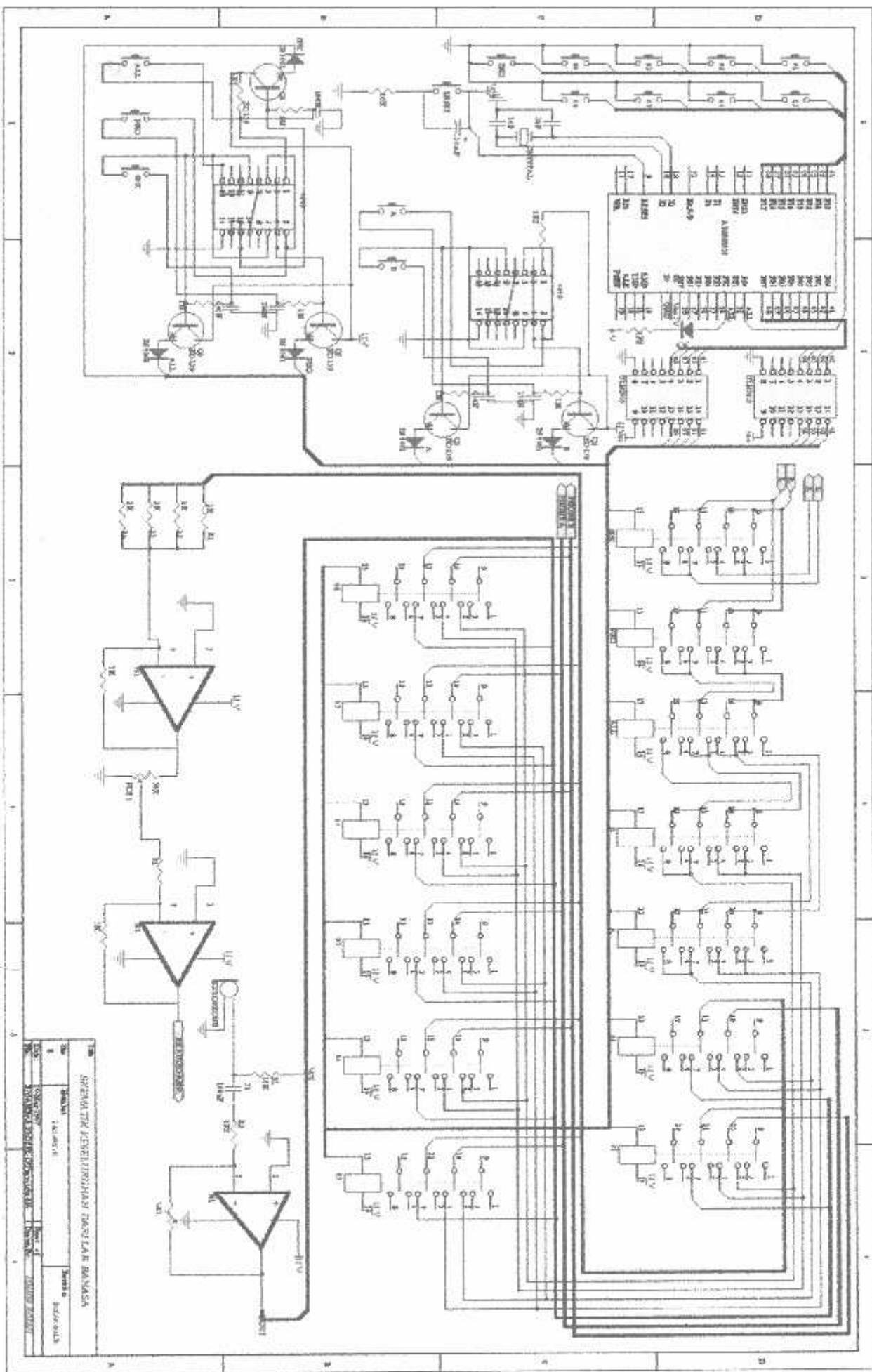
Rangkaian switch elektronik yang digunakan adalah jenis ic cmos dengan tipe 4049, dimana ic ini akan aktif jika mendapat logic high dan jika logic low maka tidak aktif

Gambar 2-10 Konfigurasi Pin-pin 4049

PIN ASSIGNMENT		
VCO	1	16 NC
CUTA	2	15 OUTF
INA	3	14 INE
OUTB	4	13 NC
NE	6	12 OUTE
CUTC	6	11 INE
NC	7	10 OUTD
VSS	8	9 ND

NC = NO CONNECTION

Sumber : WWW.datasheet.4049.com



```
goto awal;
```

```
(C == 1)
```

```
(P1 == 251)
```

```
X = 0;  
MURID3 = 1;  
goto awal;
```

```
X == 0)
```

```
(P1 == 251)
```

```
X = 1;  
MURID3 = 0;  
goto awal;
```

```
X == 1)
```

```
(P1 == 247)
```

```
X = 0;  
MURID4 = 1;  
goto awal;
```

```
X == 0)
```

```
(P1 == 247)
```

```
X = 1;  
MURID4 = 0;  
goto awal;
```

```
X == 1)
```

```
(P1 == 239)
```

```
{  
X = 0;  
MURID5 = 1;  
goto awal;  
}
```

```
(X == 0)
```

```
f (P1 == 239)
```

```
{  
    X = 1;  
    MURID5 = 0;  
    goto awal;  
}  
}
```

```
: (X == 1)  
{  
    if (P1 == 223)  
    {  
        X = 0;  
        MURID6 = 1;  
        goto awal;  
    }
```

```
: (X == 0)  
if (P1 == 223)  
{  
    X = 1;  
    MURID6 = 0;  
    goto awal;  
}
```

```
: (X == 1)  
f (P1 == 191)  
{  
    X = 0;  
    MURID7 = 1;  
    goto awal;  
}
```

```
: (X == 0)  
F (P1 == 191)  
{  
    X = 1;  
    MURID7 = 0;  
    goto awal;  
}
```

```
: X == 1)  
(P1 == 127)  
X = 0;  
MURID8 = 1;  
goto awal;
```

```
: X == 0)
```

7.C01.cpp

```
if (P1 == 127)
{
    X = 1;
    MURID8 = 0;
    goto awal;
}
(P2 == 254)
{
    goto awal1;
}
oto awal;

:al1:
{
    X = 2;
    P2 = 251;
    if (Y == 1)
    {
        if (P1 == 254)
        {
            Y = 0;
            P2 = 254;
            goto awal1;
        }
    }

    if (Y == 0)
    {
        if (P1 == 254)
        {
            Y = 1;
            P2 = 255;
            goto awal1;
        }
    }

    if (Y == 1)
    {
        if (P1 == 253)
        {
            Y = 0;
            P2 = 255;
            P2 = 253;
            goto awal1;
        }
    }

    if (Y == 0)
    {
        if (P1 == 253)
        {
            Y = 1;
            P2 = 255;
            goto awal1;
        }
    }
}
```

```
if (Y == 1)
{
    if (P1 == 251)
    {
        Y = 0;
        P2 = 255;
        P2 = 251;
        goto awal1;
    }
}
```

```
if (Y == 0)
{
    if (P1 == 251)
    {
        Y = 1;
        P2 = 255;
        goto awal1;
    }
}
```

```
if (Y == 1)
{
    if (P1 == 247)
    {
        Y = 0;
        P2 = 255;
        P2 = 247;
        goto awal1;
    }
}
```

```
if (Y == 0)
{
    if (P1 == 247)
    {
        Y = 1;
        P2 = 255;
        goto awal1;
    }
}
```

```
if (Y == 1)
{
    if (P1 == 239)
    {
        Y = 0;
        P2 = 255;
        P2 = 239;
        goto awal1;
    }
}
```

7.C01.cpp

```
if (Y == 0)
{
    if (P1 == 239)
    {
        Y = 1;
    P2 = 255;
        goto awal1;
    }
}

if (Y == 1)
{
    if (P1 == 223)
    {
        Y = 0;
    P2 = 255;
    P2 = 223;
        goto awal1;
    }
}

if (Y == 0)
{
    if (P1 == 223)
    {
        Y = 1;
    P2 = 255;
        goto awal1;
    }
}

if (Y == 1)
{
    if (P1 == 191)
    {
        Y = 0;
    P2 = 255;
    P2 = 191;
        goto awal1;
    }
}

if (Y == 0)
{
    if (P1 == 191)
    {
        Y = 1;
    P2 = 255;
        goto awal1;
    }
}

if (Y == 1)
{
    if (P1 == 127)
    {
        Y = 0;
```



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JURUSAN TEKNIK ELEKTRO S-1
KONSENTRASI ELEKTRONIKA

LEMBAR PERSETUJUAN PERBAIKAN SKRIPSI

Dari hasil ujian skripsi Jurusan Teknik Elektro jenjang Strata Satu (S-1), yang diselenggarakan pada :

Hari : Jum'at

Tanggal : 23 maret 2007

Telah dilakukan perbaikan oleh :

Nama : Isnaini Rozafi

NIM : 01.17.044

Jurusan : Teknik Elektro S-1

Konsentrasi : Elektronika

Judul Skripsi : **Perencanaan dan Pembuatan Master Kontrol Lab.Bahasa Delapan Channel Berbasis Mikrokontroller AT89S8252**

Perbaikan meliputi :

No.	Materi Perbaikan	Keterangan	Paraf Pengaji I
01.	Gambar Skema Lengkap	Harap disempurnakan	
02.	Gambar hal.36 dan 39	Diperbaiki	
03.	Hasil Pengujian hal.37	Diperbaiki	

Pengaji I

Ir. Widodo Pudji M, MT.
NIP.Y. 1028700171



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KONSENTRASI ELEKTRONIKA

LEMBAR PERSETUJUAN PERBAIKAN SKRIPSI

Dari hasil ujian skripsi Jurusan Teknik Elektro jenjang Strata Satu (S-1), yang diselenggarakan pada :

Hari : Jum'at

Tanggal : 23 maret 2007

Telah dilakukan perbaikan oleh :

Nama : Isnaini Rozafi

NIM : 01.17.044

Jurusan : Teknik Elektro S-1

Konsentrasi : Elektronika

Judul Skripsi : Perencanaan dan Pembuatan Master Kontrol Lab.Bahasa

Delapan Channel Berbasis Mikrokontroller AT89S8252

Perbaikan meliputi :

No.	Materi Perbaikan	Keterangan	Paraf Pengaji II
01.	Gambar Skema Lengkap	Harap disempurnakan	

Pengaji II



Dr. Cahyo Crysdiyan, MSC.
NIP. 1030400412

INSTITUT TEKNOLOGI NASIONAL
FAKULTAS TEKNOLOGI INDUSTRI
JURUSAN TEKNIK ELEKTRO

Formulir Perbaikan Ujian Skripsi

Untuk pelaksanaan Ujian Skripsi Janjang Strata 1 Jurusan Teknik Elektro Konsentrasi T. Energi
s / T. Elektronika, maka perlu adanya perbaikan skripsi untuk mahasiswa :

IA : Yudaini R
I :
aikan meliputi :

Gambar Skema layar display
Akhir penulisan .

Gambar hal 36 2 yg lengkap
Akhir penulisan .

) Gambar hal 37 yg lengkap
Akhir penulisan .

Malang,

W.R.
(Lilik Herlina)



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FAKULTAS TEKNOLOGI INDUSTRI
JURUSAN TEKNIK ELEKTRO

Formulir Perbaikan Ujian Skripsi

Dalam pelaksanaan Ujian Skripsi Janjang Strata 1 Jurusan Teknik Elektro Konsentasi T. Energi Listrik / T. Elektronika, maka perlu adanya perbaikan skripsi untuk mahasiswa :

NAMA : ...

N I M : ...

Perbaikan meliputi : ...

(1) *Pelajaran Agama* *mpa*
(Dua detahnya)

Malang,

(*[Signature]*)

FORMULIR BIMBINGAN SKRIPSI

Bimbingan Skripsi : ISNAINI ROZAFI
Bimbingan Skripsi : 01.17.044
Bimbingan Skripsi : 10 Februari 2007 s/d 10 Agustus 2007
Bimbingan Skripsi : Perencanaan dan Pembuatan Master Kontrol Lab. Bahasa 8 Channel
Berbasis Mikrokontroller AT89S8252

Tanggal	Uraian	Paraf Pembimbing
10 - 2 - 07	Bab I & II	X
10 - 2 - 07	Revisi Bab III	X
18 - 02 - 07	Revisi Bab IV	X
21 - 02 - 07	Bab III & IV	X
25 - 02 - 07	Revisi Bab V	X
27 - 02 - 07	Revisi Bab VI	X
08 - 03 - 07	Acc Seminar Engl	X
11 - 03 - 2007	Acc Lengkap	X

Malang, 12 - 3 - 2007
Dosen Pembimbing

Joseph Dedy Irawan, ST., MT.
NIP. 132315178

Form S-4

features

Compatible with MCS®51 Products

8K Bytes of In-System Reprogrammable Downloadable Flash Memory

- SPI Serial Interface for Program Downloading
- Endurance: 1,000 Write/Erase Cycles

1K Bytes EEPROM

- Endurance: 100,000 Write/Erase Cycles

3V to 6V Operating Range

Fully Static Operation: 0 Hz to 24 MHz

Three-level Program Memory Lock

156 x 8-bit Internal RAM

32 Programmable I/O Lines

Three 16-bit Timer/Counters

Nine Interrupt Sources

Programmable UART Serial Channel

SPI Serial Interface

Low-power Idle and Power-down Modes

Interrupt Recovery from Power-down

Programmable Watchdog Timer

Dual Data Pointer

Power-off Flag

description

The AT89S8252 is a low-power, high-performance CMOS 8-bit microcontroller with 8K bytes of downloadable Flash programmable and erasable read-only memory and 2K bytes of EEPROM. The device is manufactured using Atmel's high-density nonvolatile memory technology and is compatible with the industry-standard 80C51 instruction set and pinout. The on-chip downloadable Flash allows the program memory to be programmed In-System through an SPI serial interface or by a conventional nonvolatile memory programmer. By combining a versatile 8-bit CPU with downloadable Flash on a monolithic chip, the Atmel AT89S8252 is a powerful microcontroller, which provides a highly-flexible and cost-effective solution to many embedded control applications.

The AT89S8252 provides the following standard features: 8K bytes of downloadable Flash, 2K bytes of EEPROM, 256 bytes of RAM, 32 I/O lines, programmable watchdog timer, two data pointers, three 16-bit timer/counters, a six-vector two-level interrupt architecture, a full duplex serial port, on-chip oscillator, and clock circuitry. In addition, the AT89S8252 is designed with static logic for operation down to zero frequency and supports two software selectable power saving modes. The Idle Mode stops the CPU while allowing the RAM, timer/counters, serial port, and interrupt system to continue functioning. The Power-down mode saves the RAM contents but freezes the oscillator, enabling all other chip functions until the next external interrupt or hardware reset.

The downloadable Flash can be changed a single byte at a time and is accessible through the SPI serial interface. Holding RESET active forces the SPI bus into a serial programming interface and allows the program memory to be written to or read from unless lock bits have been activated.



8-bit Microcontroller with 8K Bytes Flash

AT89S8252

0401F-MICRO-11/03



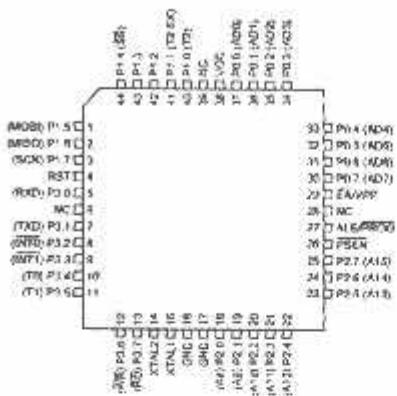


in Configurations

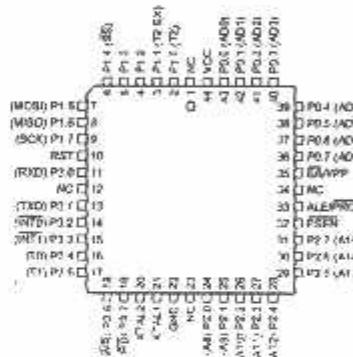
PDIP



TQFP



PLCC



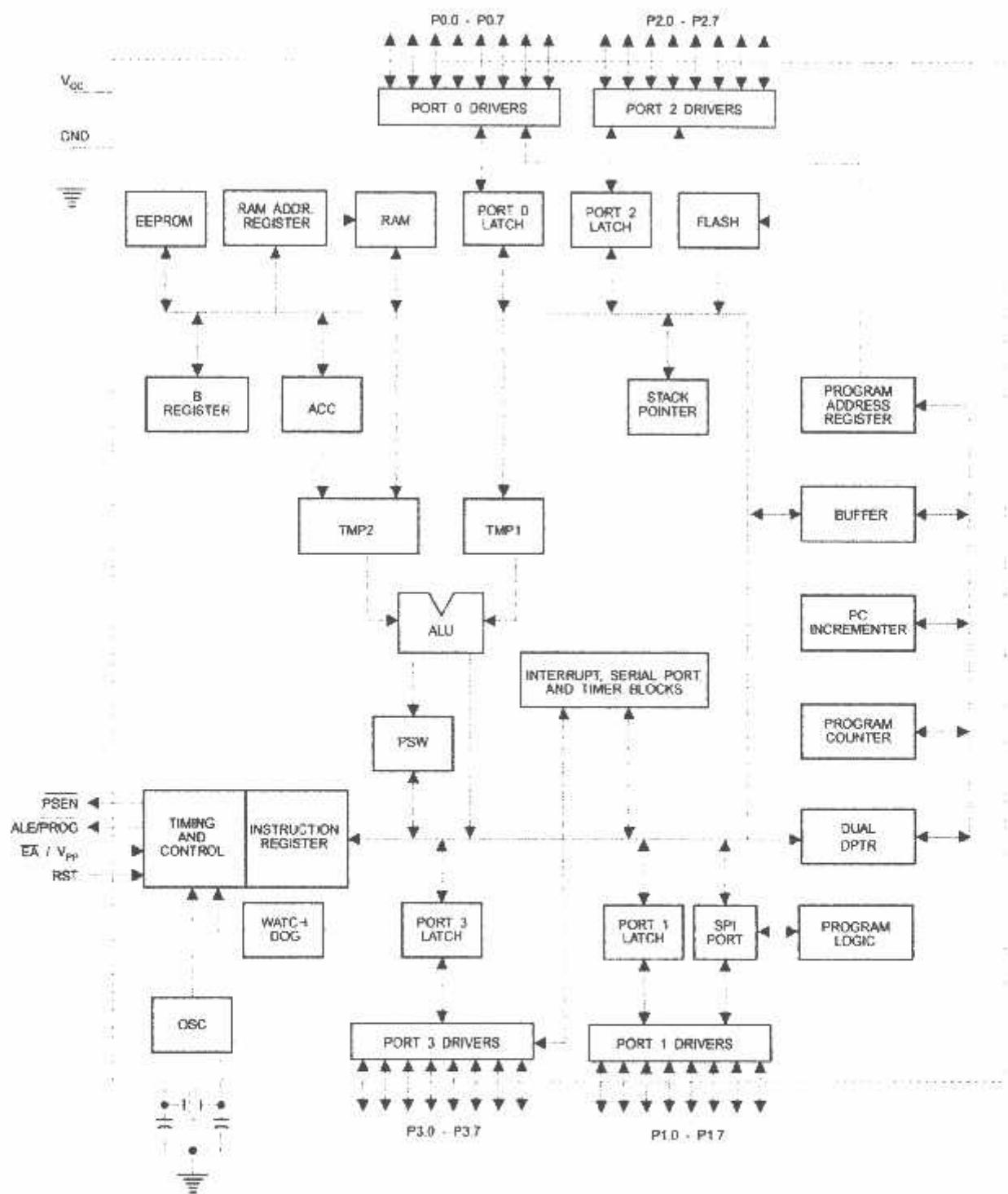
Description

- :C Supply voltage.
- :G Ground.
- :P0 Port 0 is an 8-bit open drain bi-directional I/O port. As an output port, each pin can sink eight TTL inputs. When 1s are written to Port 0 pins, the pins can be used as high-impedance inputs. Port 0 can also be configured to be the multiplexed low-order address/data bus during accesses to external program and data memory. In this mode, P0 has internal pull-ups. Port 0 also receives the code bytes during Flash programming and outputs the code bytes during program verification. External pull-ups are required during program verification.
- :P1 Port 1 is an 8-bit bi-directional I/O port with internal pull-ups. The Port 1 output buffers can sink/source four TTL inputs. When 1s are written to Port 1 pins, they are pulled high by the internal pull-ups and can be used as inputs. As inputs, Port 1 pins that are externally being pulled low will source current (I_{IL}) because of the internal pull-ups.

AT89S8252

0401F-MICRO-11/03

Block Diagram





Some Port 1 pins provide additional functions. P1.0 and P1.1 can be configured to be the timer/counter 2 external count input (P1.0/T2) and the timer/counter 2 trigger input (P1.1/T2EX), respectively.

Furthermore, P1.4, P1.5, P1.6, and P1.7 can be configured as the SPI slave port select, data input/output and shift clock input/output pins as shown in the following table.

Port Pin	Alternate Functions
P1.0	T2 (external count input to Timer/Counter 2), clock-out
P1.1	T2EX (Timer/Counter 2 capture/reload trigger and direction control)
P1.4	SS (Slave port select input)
P1.5	MOSI (Master data output, slave data input pin for SPI channel)
P1.6	MISO (Master data input, slave data output pin for SPI channel)
P1.7	SCK (Master clock output, slave clock input pin for SPI channel)

Port 1 also receives the low-order address bytes during Flash programming and verification.

Port 2

Port 2 is an 8-bit bi-directional I/O port with internal pull-ups. The Port 2 output buffers can sink/source four TTL inputs. When 1s are written to Port 2 pins, they are pulled high by the internal pull-ups and can be used as inputs. As inputs, Port 2 pins that are externally being pulled low will source current (I_{IL}) because of the internal pull-ups.

Port 2 emits the high-order address byte during fetches from external program memory and during accesses to external data memory that use 16-bit addresses (MOVX @ DPTR). In this application, Port 2 uses strong internal pull-ups when emitting 1s. During accesses to external data memory that use 8-bit addresses (MOVX @ R1), Port 2 emits the contents of the P2 Special Function Register.

Port 2 also receives the high-order address bits and some control signals during Flash programming and verification.

Port 3

Port 3 is an 8-bit bi-directional I/O port with internal pull-ups. The Port 3 output buffers can sink/source four TTL inputs. When 1s are written to Port 3 pins, they are pulled high by the internal pull-ups and can be used as inputs. As inputs, Port 3 pins that are externally being pulled low will source current (I_{IL}) because of the pull-ups.

Port 3 receives some control signals for Flash programming and verification.

Port 3 also serves the functions of various special features of the AT89S8252, as shown in the following table.

Port Pin	Alternate Functions
P3.0	RXD (serial input port)
P3.1	TXD (serial output port)
P3.2	INT0 (external interrupt 0)
P3.3	INT1 (external interrupt 1)
P3.4	T0 (timer 0 external input)
P3.5	T1 (timer 1 external input)
P3.6	WR (external data memory write strobe)
P3.7	RD (external data memory read strobe)

ST

Reset Input. A high on this pin for two machine cycles while the oscillator is running resets the device.

LE/PROG

Address Latch Enable is an output pulse for latching the low byte of the address during accesses to external memory. This pin is also the program pulse input (PROG) during Flash programming.

In normal operation, ALE is emitted at a constant rate of 1/6 the oscillator frequency and may be used for external timing or clocking purposes. Note, however, that one ALE pulse is skipped during each access to external data memory.

If desired, ALE operation can be disabled by setting bit 0 of SFR location 8EH. With the bit set, ALE is active only during a MOVX or MOVC instruction. Otherwise, the pin is weakly pulled high. Setting the ALE-disable bit has no effect if the microcontroller is in external execution mode.

SEN

Program Store Enable is the read strobe to external program memory.

When the AT89S8252 is executing code from external program memory, PSEN is activated twice each machine cycle, except that two PSEN activations are skipped during each access to external data memory.

A/VPP

External Access Enable. EA must be strapped to GND in order to enable the device to fetch code from external program memory locations starting at 0000H up to FFFFH. Note, however, that if lock bit 1 is programmed, EA will be internally latched on reset.

EA should be strapped to V_{CC} for internal program executions. This pin also receives the 12-volt programming enable voltage (V_{pp}) during Flash programming when 12-volt programming is selected.

TAL1

Input to the inverting oscillator amplifier and input to the internal clock operating circuit.

TAL2

Output from the inverting oscillator amplifier.



Special Function Registers

A map of the on-chip memory area called the Special Function Register (SFR) space is shown in Table 1.

Note that not all of the addresses are occupied, and unoccupied addresses may not be implemented on the chip. Read accesses to these addresses will in general return random data, and write accesses will have an indeterminate effect.

User software should not write 1s to these unlisted locations, since they may be used in future products to invoke new features. In that case, the reset or inactive values of the new bits will always be 0.

Timer 2 Registers Control and status bits are contained in registers T2CON (shown in Table 2) and T2MOD (shown in Table 9) for Timer 2. The register pair (RCAP2H, RCAP2L) are the Capture/Reload registers for Timer 2 in 16-bit capture mode or 16-bit auto-reload mode.

Table 1. AT89S8252 SFR Map and Reset Values

1FBH								0FFH
1F0H	B 00000000							0F7H
0E8H								0EFH
0E0H	ACC 00000000							0E7H
0DBH								0DFH
0DDH	PSW 00000000					SPCR 000001XX		0D7H
0C8H	T2CON 00000000	T2MOD XXXXXX00	RCAP2L 00000000	RCAP2H 00000000	TL2 00000000	TH2 00000000		0CFH
1C0H								0C7H
0B8H	P XXXX0000							0BFH
1B0H	P3 11111111							0B7H
1A8H	IE 0X000000		SPSR 00XXXXXX					0AFH
1A0H	P2 11111111							0A7H
98H	SCON 00000000	SBJF XXXXXXXX						9FH
90H	P1 11111111						WMCON 00000010	87H
88H	TCON 00000000	TMOD 00000000	TL0 00000000	TL1 00000000	TH0 00000000	TH1 00000000		8FH
80H	P0 11111111	SP 00000111	DPOL 00000000	DP0H 00000000	DP1L 00000000	DP1H 00000000	SPDR XXXXXXX	PCON 00000000
								87H

2. T2CON – Timer/Counter 2 Control Register

I/O Address = 0C8H

Reset Value = 0000 0000B

Addressable

TF2	EXF2	RCLK	TCLK	EXEN2	TR2	C/T2	CP/RL2
7	6	5	4	3	2	1	0

Symbol	Function
TF2	Timer 2 overflow flag set by a Timer 2 overflow and must be cleared by software. TF2 will not be set when either RCLK = 1 or TCLK = 1.
EXF2	Timer 2 external flag set when either a capture or reload is caused by a negative transition on T2EX and EXEN2 = 1. When Timer 2 interrupt is enabled, EXF2 = 1 will cause the CPU to vector to the Timer 2 interrupt routine. EXF2 must be cleared by software. EXF2 does not cause an interrupt in up/down counter mode (DCEN = 1).
CLK	Receive clock enable. When set, causes the serial port to use Timer 2 overflow pulses for its receive clock in serial port Modes 1 and 3. RCLK = 0 causes Timer 1 overflows to be used for the receive clock.
CLK	Transmit clock enable. When set, causes the serial port to use Timer 2 overflow pulses for its transmit clock in serial port Modes 1 and 3. TCLK = 0 causes Timer 1 overflows to be used for the transmit clock.
EXEN2	Timer 2 external enable. When set, allows a capture or reload to occur as a result of a negative transition on T2EX if Timer 2 is not being used to clock the serial port. EXEN2 = 0 causes Timer 2 to ignore events at T2EX.
TR2	Start/Stop control for Timer 2. TR2 = 1 starts the timer.
C/T2	Timer or counter select for Timer 2. C/T2 = 0 for timer function. C/T2 = 1 for external event counter (falling edge triggered).
CP/RL2	Capture/Reload select. CP/RL2 = 1 causes captures to occur on negative transitions at T2EX if EXEN2 = 1. CP/RL2 = 0 causes automatic reloads to occur when Timer 2 overflows or negative transitions occur at T2EX when EXEN2 = 1. When either RCLK or TCLK = 1, this bit is ignored and the timer is forced to auto-reload on Timer 2 overflow.



Watchdog and Memory Control Register The WMCON register contains control bits for the Watchdog Timer (shown in Figure 3). The EEMEN and EEMWE bits are used to select the 2K bytes on-chip EEPROM, and to enable byte-write. The DPS bit selects one of two DPTR registers available.

3. WMCON—Watchdog and Memory Control Register

WMCON Address = 96H Reset Value = 0000 0010B

PS2	PS1	PS0	EEMWE	EEMEN	DPS	WDTRST	WDTEN
7	6	5	4	3	2	1	0

Symbol	Function
PS2 PS1 PS0	Prescaler Bits for the Watchdog Timer. When all three bits are set to "0", the watchdog timer has a nominal period of 16 ms. When all three bits are set to "1", the nominal period is 2048 ms.
EEMWE	EEPROM Data Memory Write Enable Bit. Set this bit to "1" before initiating byte write to on-chip EEPROM with the MOVX instruction. User software should set this bit to "0" after EEPROM write is completed.
EEMEN	Internal EEPROM Access Enable. When EEMEN = 1, the MOVX instruction with DPTR will access on-chip EEPROM instead of external data memory. When EEMEN = 0, MOVX with DPTR accesses external data memory.
DPS	Data Pointer Register Select. DPS = 0 selects the first bank of Data Pointer Register, DP0, and DPS = 1 selects the second bank, DP1
WDTRST RDY/BSY	Watchdog Timer Reset and EEPROM Ready/Busy Flag. Each time this bit is set to "1" by user software, a pulse is generated to reset the watchdog timer. The WDTRST bit is then automatically reset to "0" in the next instruction cycle. The WDTRST bit is Write-Only. This bit also serves as the RDY/BSY flag in a Read-Only mode during EEPROM write. RDY/BSY = 1 means that the EEPROM is ready to be programmed. While programming operations are being executed, the RDY/BSY bit equals "0" and is automatically reset to "1" when programming is completed.
WDTEN	Watchdog Timer Enable Bit. WDTEN = 1 enables the watchdog timer and WDTEN = 0 disables the watchdog timer.

SPI Registers Control and status bits for the Serial Peripheral Interface are contained in registers SPCR (shown in Table 4) and SPSR (shown in Table 5). The SPI data bits are contained in the SPDR register. Writing the SPI data register during serial data transfer sets the Write Collision bit, WCOL, in the SPSR register. The SPDR is double buffered for writing and the values in SPDR are not changed by Reset.

Interrupt Registers The global interrupt enable bit and the individual interrupt enable bits are in the IE register. In addition, the individual interrupt enable bit for the SPI is in the SPCR register. Two priorities can be set for each of the six interrupt sources in the IP register.

Dual Data Pointer Registers To facilitate accessing both internal EEPROM and external data memory, two banks of 16-bit Data Pointer Registers are provided: DP0 at SFR address locations 82H-83H and DP1 at 84H-85H. Bit DPS = 0 in SFR WMCON selects DP0 and DPS = 1 selects DP1. The user should **ALWAYS** initialize the DPS bit to the appropriate value before accessing the respective Data Pointer Register.

Power Off Flag The Power Off Flag (POF) is located at bit_4 (PCON.4) in the PCON SFR. POF is set to "1" during power up. It can be set and reset under software control and is not affected by RESET.

Table 4. SPCR – SPI Control Register

SPCR Address = D5H								Reset Value = 0000 01XXB
Bit	SPIE	SPE	DORD	MSTR	CPOL	CPHA	SPR1	SPR0
	7	6	5	4	3	2	1	0
Symbol Function								
PIE	SPI Interrupt Enable. This bit, in conjunction with the ES bit in the IE register, enables SPI interrupts: SPIE = 1 and ES = 1 enable SPI interrupts. SPIE = 0 disables SPI interrupts.							
PE	SPI Enable. SPI = 1 enables the SPI channel and connects SS, MOSI, MISO and SCK to pins P1.4, P1.5, P1.6, and P1.7. SPI = 0 disables the SPI channel.							
ORD	Data Order. DORD = 1 selects LSB first data transmission. DORD = 0 selects MSB first data transmission.							
ISTR	Master/Slave Select. MSTR = 1 selects Master SPI mode. MSTR = 0 selects Slave SPI mode.							
POL	Clock Polarity. When CPOL = 1, SCK is high when idle. When CPOL = 0, SCK of the master device is low when not transmitting. Please refer to figure on SPI Clock Phase and Polarity Control.							
PHA	Clock Phase. The CPHA bit together with the CPOL bit controls the clock and data relationship between master and slave. Please refer to figure on SPI Clock Phase and Polarity Control.							
PR0	SPI Clock Rate Select. These two bits control the SCK rate of the device configured as master. SPR1 and SPR0 have no effect on the slave. The relationship between SCK and the oscillator frequency, F_{osc} , is as follows:							
PR1	SPR1 SPR0 SCK = F_{osc} divided by							
	0	0	4					
	0	1	16					
	1	0	64					
	1	1	128					





5. SPSR – SPI Status Register

DR Address = AAH

Reset Value = 00XX XXXXB

SPIF	WCOL	-	-	-	-	-	-
7	8	5	4	3	2	1	0

Bit	Function
F	SPI Interrupt Flag. When a serial transfer is complete, the SPIF bit is set and an interrupt is generated if SPIE = 1 and ES = 1. The SPIF bit is cleared by reading the SPI status register with SPIF and WCOL bits set, and then reading/writing the SPI data register.
COL	Write Collision Flag. The WCOL bit is set if the SPI data register is written during a data transfer. During data transfer, the result of reading the SPDR register may be incorrect, and writing to it has no effect. The WCOL bit (and the SPIF bit) are cleared by reading the SPI status register with SPIF and WCOL set, and then accessing the SPI data register.

6. SPDR – SPI Data Register

DR Address = 86H

Reset Value = unchanged

Bit	SPD7	SPD6	SPD5	SPD4	SPD3	SPD2	SPD1	SPD0
	7	6	5	4	3	2	1	0

Data Memory – EEPROM and RAM

The AT89S8252 implements 2K bytes of on-chip EEPROM for data storage and 256 bytes of RAM. The upper 128 bytes of RAM occupy a parallel space to the Special Function Registers. That means the upper 128 bytes have the same addresses as the SFR space but are physically separate from SFR space.

When an instruction accesses an internal location above address 7FH, the address mode used in the instruction specifies whether the CPU accesses the upper 128 bytes of RAM or the SFR space. Instructions that use direct addressing access SFR space.

For example, the following direct addressing instruction accesses the SFR at location 0A0H (which is P2).

```
MOV 0A0H, #data
```

Instructions that use indirect addressing access the upper 128 bytes of RAM. For example, the following indirect addressing instruction, where R0 contains 0A0H, accesses the data byte at address 0A0H, rather than P2 (whose address is 0A0H).

```
MOV @R0, #data
```

Note that stack operations are examples of indirect addressing, so the upper 128 bytes of data RAM are available as stack space.

The on-chip EEPROM data memory is selected by setting the EEMEN bit in the WMCON register at SFR address location 96H. The EEPROM address range is from 000H to 7FFH. The MOVX instructions are used to access the EEPROM. To access off-chip data memory with the MOVX instructions, the EEMEN bit needs to be set to "0".

The EEMWE bit in the WMCON register needs to be set to "1" before any byte location in the EEPROM can be written. User software should reset EEMWE bit to "0" if no further EEPROM write is required. EEPROM write cycles in the serial programming mode are self-timed and typically take 2.5 ms. The progress of EEPROM write can be monitored by reading the RDY/BSY bit (read-only) in SFR WMCON. RDY/BSY = 0 means

programming is still in progress and RDY/BSY = 1 means EEPROM write cycle is completed and another write cycle can be initiated.

In addition, during EEPROM programming, an attempted read from the EEPROM will fetch the byte being written with the MSB complemented. Once the write cycle is completed, true data are valid at all bit locations.

Programmable Watchdog Timer

The programmable Watchdog Timer (WDT) operates from an independent internal oscillator. The prescaler bits, PS0, PS1 and PS2 in SFR WMCON are used to set the period of the Watchdog Timer from 16 ms to 2048 ms. The available timer periods are shown in the following table and the actual timer periods (at $V_{CC} = 5V$) are within $\pm 30\%$ of the nominal.

The WDT is disabled by Power-on Reset and during Power-down. It is enabled by setting the WDTE bit in SFR WMCON (address = 96H). The WDT is reset by setting the WDTRST bit in WMCON. When the WDT times out without being reset or disabled, an internal RST pulse is generated to reset the CPU.

Table 7. Watchdog Timer Period Selection

WDT Prescaler Bits			Period (nominal)
PS2	PS1	PS0	
0	0	0	16 ms
0	0	1	32 ms
0	1	0	64 ms
0	1	1	128 ms
1	0	0	256 ms
1	0	1	512 ms
1	1	0	1024 ms
1	1	1	2048 ms

Timer 0 and 1

Timer 0 and Timer 1 in the AT89S8252 operate the same way as Timer 0 and Timer 1 in the AT89C51 and AT89C52. For further information on the timers' operation, refer to the Atmel web site (<http://www.atmel.com>). From the home page, select "Products", then "Microcontrollers", then "8051-Architecture". Click on "Documentation", then on "Other Documents". Open the document "AT89 Series Hardware Description".

Timer 2

Timer 2 is a 16-bit Timer/Counter that can operate as either a timer or an event counter. The type of operation is selected by bit C/T2 in the SFR T2CON (shown in Table 2). Timer 2 has three operating modes: capture, auto-reload (up or down counting), and baud rate generator. The modes are selected by bits in T2CON, as shown in Table 8.

Timer 2 consists of two 8-bit registers, TH2 and TL2. In the Timer function, the TL2 register is incremented every machine cycle. Since a machine cycle consists of 12 oscillator periods, the count rate is 1/12 of the oscillator frequency.

In the Counter function, the register is incremented in response to a 1-to-0 transition at its corresponding external input pin, T2. In this function, the external input is sampled during S5P2 of every machine cycle. When the samples show a high in one cycle and a low in the next cycle, the count is incremented. The new count value appears in the register during S3P1 of the cycle following the one in which the transition was detected.

Since two machine cycles (24 oscillator periods) are required to recognize a 1-to-0 transition, the maximum count rate is 1/24 of the oscillator frequency. To ensure that a given level is sampled at least once before it changes, the level should be held for at least one full machine cycle.

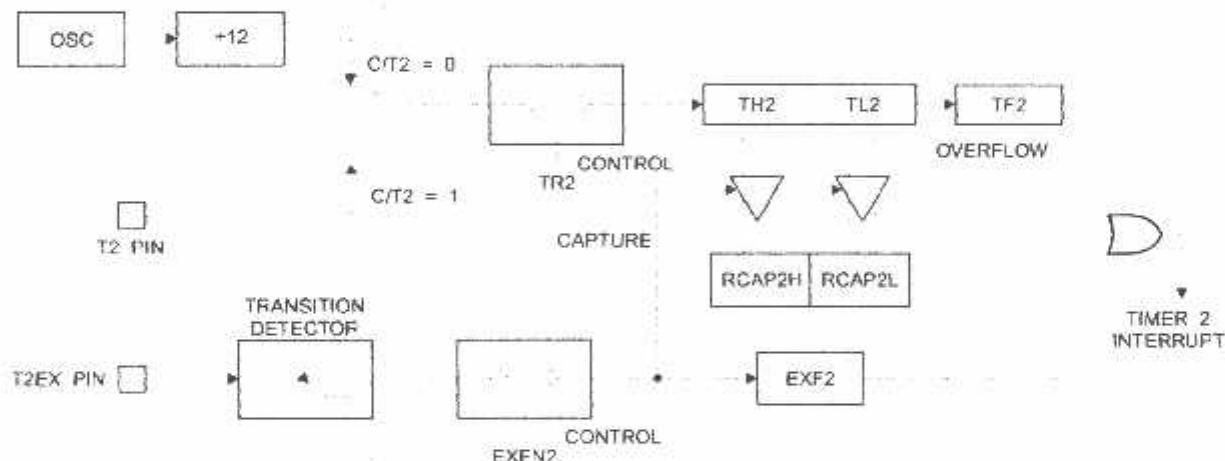
Table 8. Timer 2 Operating Modes

RCLK + TCLK	CP/RRL2	TR2	MODE
0	0	1	16-bit Auto-reload
0	1	1	16-bit Capture
1	X	1	Baud Rate Generator
X	X	0	(Off)

Capture Mode

In the capture mode, two options are selected by bit EXEN2 in T2CON. If EXEN2 = 0, Timer 2 is a 16-bit timer or counter which upon overflow sets bit TF2 in T2CON. This bit can then be used to generate an interrupt. If EXEN2 = 1, Timer 2 performs the same operation, but a 1-to-0 transition at external input T2EX also causes the current value in TH2 and TL2 to be captured into RCAP2H and RCAP2L, respectively. In addition, the transition at T2EX causes bit EXF2 in T2CON to be set. The EXF2 bit, like TF2, can generate an interrupt. The capture mode is illustrated in Figure 1.

Figure 1. Timer 2 in Capture Mode



to-reload (Up or Down counter)

Timer 2 can be programmed to count up or down when configured in its 16-bit auto-reload mode. This feature is invoked by the DCEN (Down Counter Enable) bit located in the SFR T2MOD (see Table 9). Upon reset, the DCEN bit is set to 0 so that timer 2 will default to count up. When DCEN is set, Timer 2 can count up or down, depending on the value of the T2EX pin.

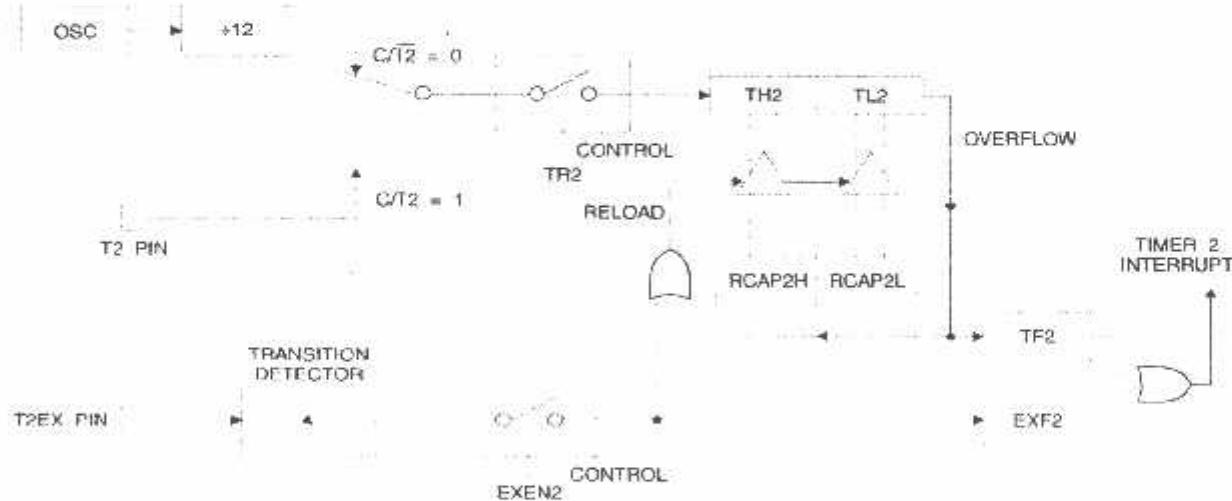
Figure 2 shows Timer 2 automatically counting up when DCEN = 0. In this mode, two options are selected by bit EXEN2 in T2CON. If EXEN2 = 0, Timer 2 counts up to OFFFFH and then sets the TF2 bit upon overflow. The overflow also causes the timer registers to be reloaded with the 16-bit value in RCAP2H and RCAP2L. The values in RCAP2H and RCAP2L are preset by software. If EXEN2 = 1, a 16-bit reload can be triggered either by an overflow or by a 1-to-0 transition at external input T2EX. This transition also sets the EXF2 bit. Both the TF2 and EXF2 bits can generate an interrupt if enabled.

Setting the DCEN bit enables Timer 2 to count up or down, as shown in Figure 3. In this mode, the T2EX pin controls the direction of the count. A logic 1 at T2EX makes Timer 2 count up. The timer will overflow at OFFFFH and set the TF2 bit. This overflow also causes the 16-bit value in RCAP2H and RCAP2L to be reloaded into the timer registers, TH2 and TL2, respectively.

A logic 0 at T2EX makes Timer 2 count down. The timer underflows when TH2 and TL2 equal the values stored in RCAP2H and RCAP2L. The underflow sets the TF2 bit and causes OFFFFH to be reloaded into the timer registers.

The EXF2 bit toggles whenever Timer 2 overflows or underflows and can be used as a 17th bit of resolution. In this operating mode, EXF2 does not flag an interrupt.

Figure 2. Timer 2 in Auto Reload Mode (DCEN = 0)



Lampiran



PIC16F87XA Data Sheet

28/40/44-Pin Enhanced Flash
Microcontrollers

28/40/44-Pin Enhanced Flash Microcontrollers

Devices Included in this Data Sheet:

- PIC16F873A
- PIC16F876A
- PIC16F874A
- PIC16F877A

High-Performance RISC CPU:

- Only 35 single-word instructions to learn
- All single-cycle instructions except for program branches, which are two-cycle
- Operating speed: DC – 20 MHz clock input
DC – 200 ns instruction cycle
- Up to 8K x 14 words of Flash Program Memory,
Up to 368 x 8 bytes of Data Memory (RAM),
Up to 256 x 8 bytes of EEPROM Data Memory
- Pinout compatible to other 28-pin or 40/44-pin
PIC16CXXX and PIC16FXXX microcontrollers

Peripheral Features:

- Timer0: 8-bit timer/counter with 8-bit prescaler
- Timer1: 16-bit timer/counter with prescaler, can be incremented during Sleep via external crystal/clock
- Timer2: 8-bit timer/counter with 8-bit period register, prescaler and postscaler
- Two Capture, Compare, PWM modules
 - Capture is 16-bit, max. resolution is 12.5 ns
 - Compare is 16-bit, max. resolution is 200 ns
 - PWM max. resolution is 10-bit
- Synchronous Serial Port (SSP) with SPI™ (Master mode) and I²C™ (Master/Slave)
- Universal Synchronous Asynchronous Receiver Transmitter (USART/SCI) with 9-bit address detection
- Parallel Slave Port (PSP) – 8 bits wide with external RD, WR and CS controls (40/44-pin only)
- Brown-out detection circuitry for Brown-out Reset (BOR)

Analog Features:

- 10-bit, up to 8-channel Analog-to-Digital Converter (A/D)
- Brown-out Reset (BOR)
- Analog Comparator module with:
 - Two analog comparators
 - Programmable on-chip voltage reference (VREF) module
 - Programmable input multiplexing from device inputs and internal voltage reference
 - Comparator outputs are externally accessible

Special Microcontroller Features:

- 100,000 erase/write cycle Enhanced Flash program memory typical
- 1,000,000 erase/write cycle Data EEPROM memory typical
- Data EEPROM Retention > 40 years
- Self-reprogrammable under software control
- In-Circuit Serial Programming™ (ICSP™) via two pins
- Single-supply 5V In-Circuit Serial Programming
- Watchdog Timer (WDT) with its own on-chip RC oscillator for reliable operation
- Programmable code protection
- Power saving Sleep mode
- Selectable oscillator options
- In-Circuit Debug (ICD) via two pins

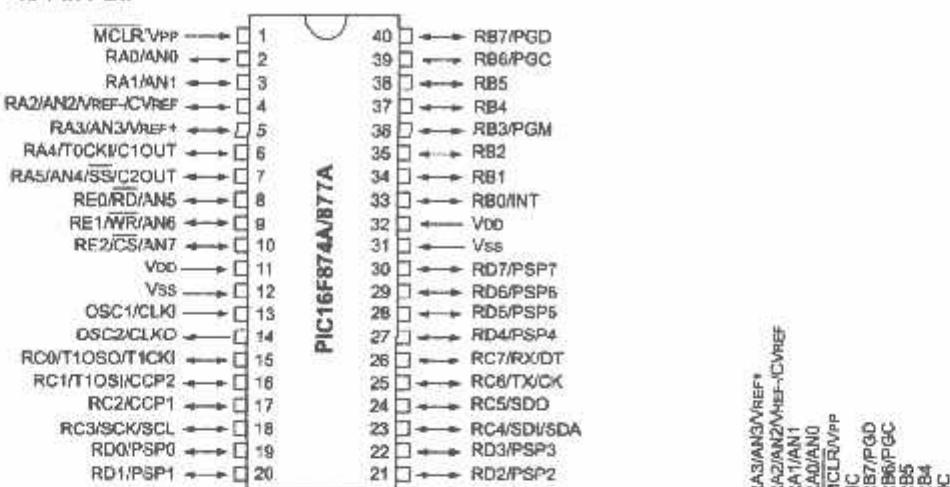
CMOS Technology:

- Low-power, high-speed Flash/EEPROM technology
- Fully static design
- Wide operating voltage range (2.0V to 5.5V)
- Commercial and Industrial temperature ranges
- Low-power consumption

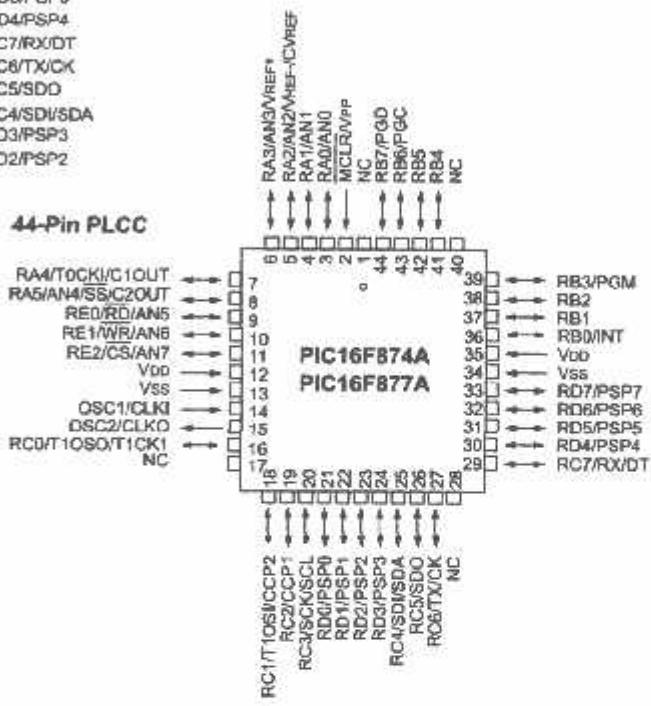
Device	Program Memory		Data SRAM (Bytes)	EEPROM (Bytes)	I/O	10-bit A/D (ch)	CCP (PWM)	MSSP		USART	Timers 8/16-bit	Comparators
	Bytes	# Single Word Instructions						SPI	Master I²C			
PIC16F873A	7.2K	4096	192	128	22	5	2	Yes	Yes	Yes	2/1	2
PIC16F874A	7.2K	4096	192	128	33	8	2	Yes	Yes	Yes	2/1	2
PIC16F876A	14.3K	8192	368	256	22	5	2	Yes	Yes	Yes	2/1	2
PIC16F877A	14.3K	8192	368	256	33	8	2	Yes	Yes	Yes	2/1	2

Pin Diagrams (Continued)

40-Pin PDIP



44-Pin PLCC



44-Pin TQFP

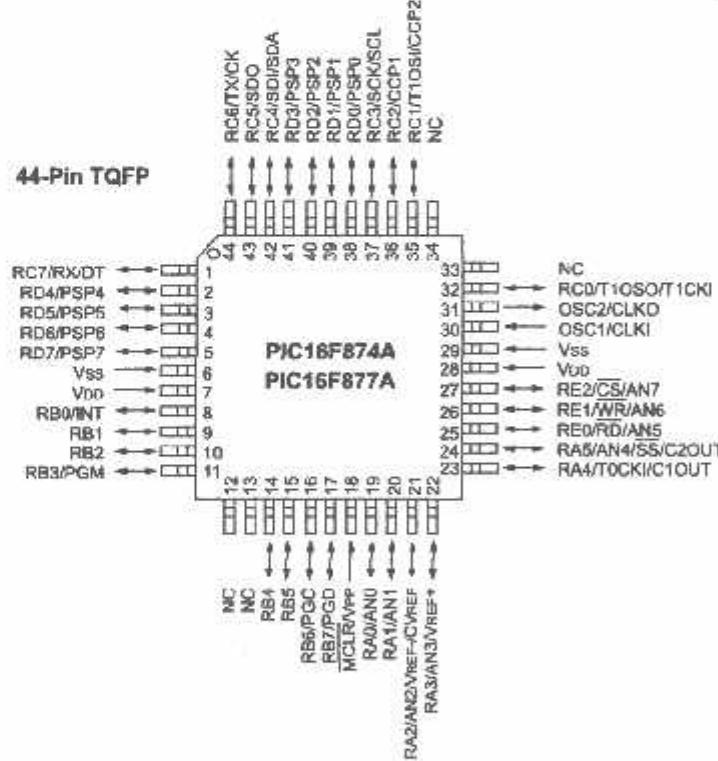
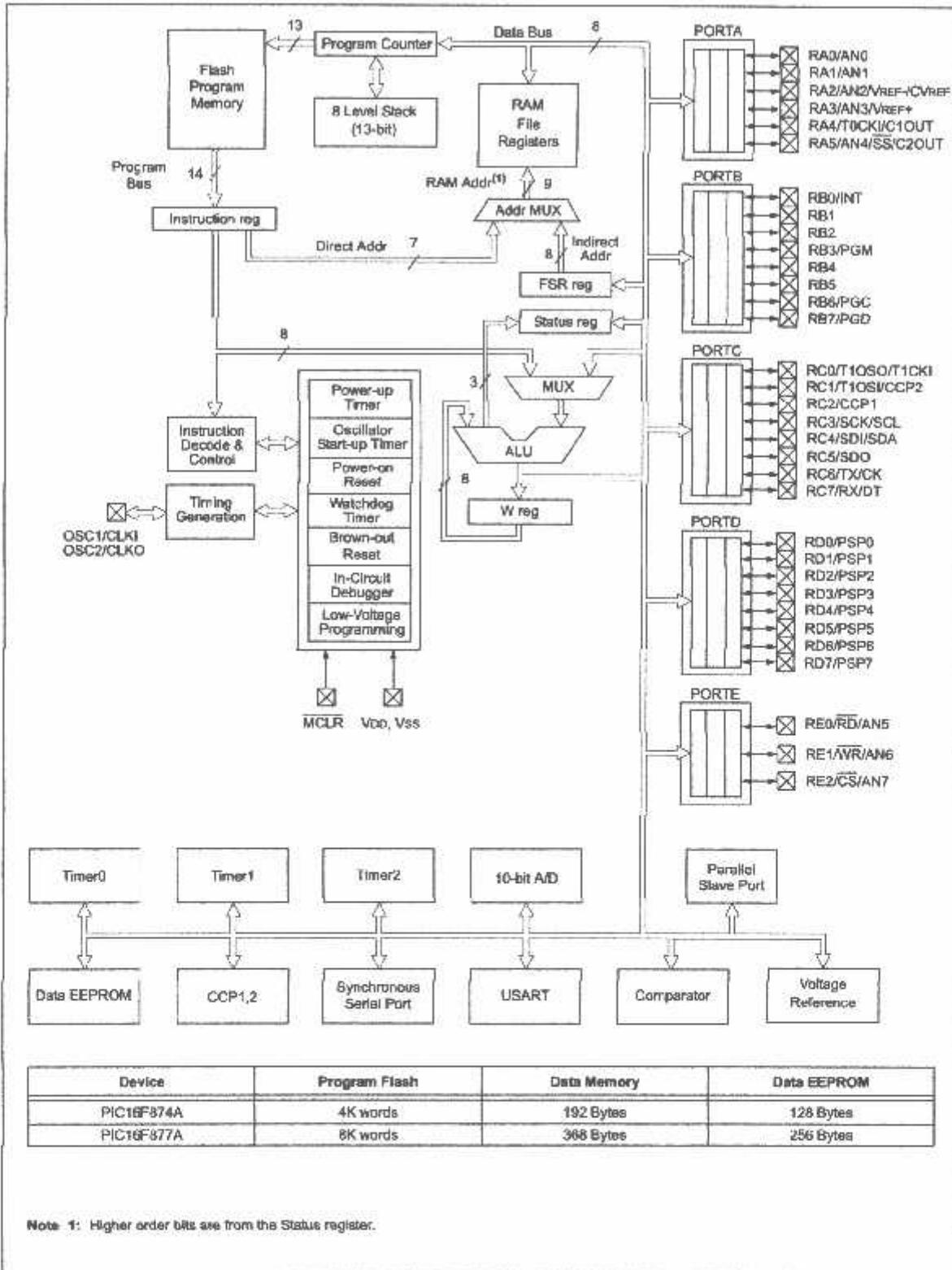


FIGURE 1-2: PIC16F874A/877A BLOCK DIAGRAM



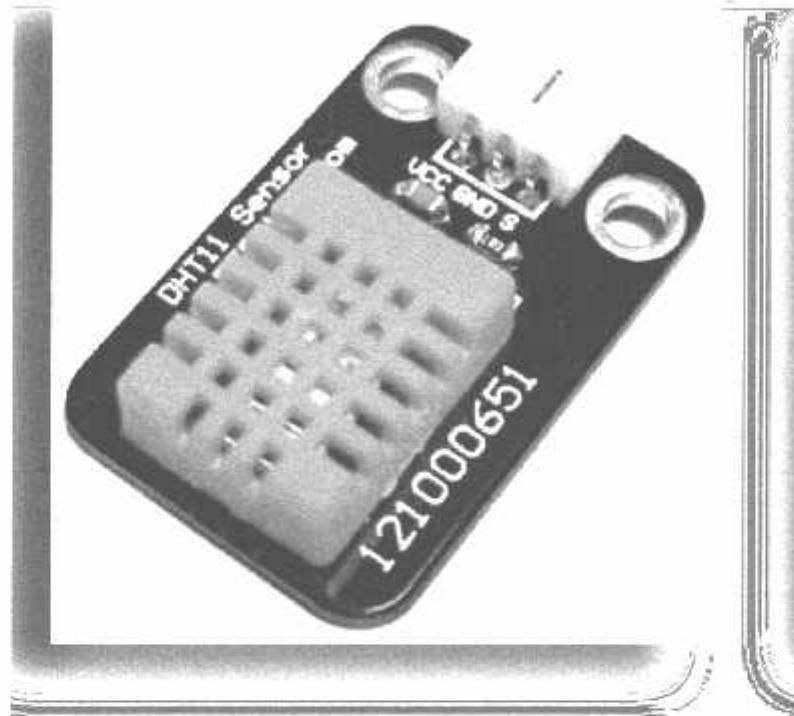
DHT11 Humidity & Temperature Sensor

D-Robotics UK (www.droboticsonline.com)

DHT11 Temperature & Humidity Sensor features a temperature & humidity sensor complex with a calibrated digital signal output.

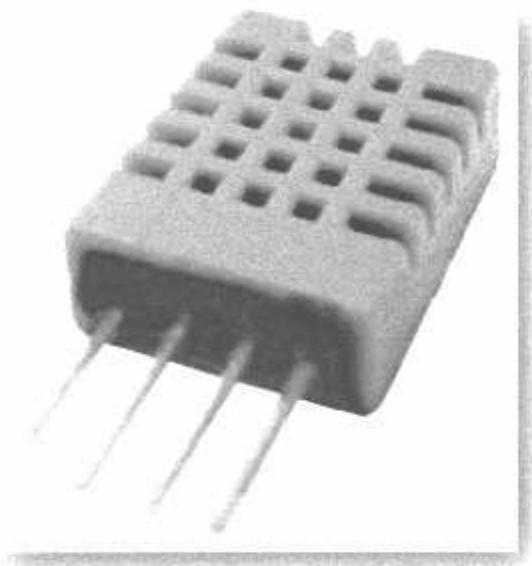
D-Robotics
7/30/2010

DHT 11 Humidity & Temperature Sensor



1. Introduction

This DFRobot DHT11 Temperature & Humidity Sensor features a temperature & humidity sensor complex with a calibrated digital signal output. By using the exclusive digital-signal-acquisition technique and temperature & humidity sensing technology, it ensures high reliability and excellent long-term stability. This sensor includes a resistive-type humidity measurement component and an NTC temperature measurement component, and connects to a high-performance 8-bit microcontroller, offering excellent quality, fast response, anti-interference ability and cost-effectiveness.



Each DHT11 element is strictly calibrated in the laboratory that is extremely accurate on humidity calibration. The calibration coefficients are stored as programmes in the OTP memory, which are used by the sensor's internal signal detecting process. The single-wire serial interface makes system integration quick and easy. Its small size, low power consumption and up-to-20 meter signal transmission making it the best choice for various applications, including those most demanding ones. The component is 4-pin single row pin package. It is convenient to connect and special packages can be provided according to users' request.

2. Technical Specifications:

Overview:

Item	Measurement Range	Humidity Accuracy	Temperature Accuracy	Resolution	Package
DHT11	20-90%RH 0-50 °C	±5%RH	±2°C	1	4 Pin Single Row

Detailed Specifications:

Parameters	Conditions	Minimum	Typical	Maximum
Humidity				
Resolution		1%RH	1%RH	1%RH
			8 Bit	
Repeatability			±1%RH	
Accuracy	25°C		±4%RH	
	0-50°C			±5%RH
Interchangeability	Fully Interchangeable			
Measurement Range	0°C	30%RH		90%RH
	25°C	20%RH		90%RH
	50°C	20%RH		80%RH
Response Time (Seconds)	1/e(63%) 25°C, 1m/s Air	6 S	10 S	15 S
Hysteresis			±1%RH	
Long-Term Stability	Typical		±1%RH/year	
Temperature				
Resolution		1°C	1°C	1°C
		8 Bit	8 Bit	8 Bit
Repeatability			±1°C	
Accuracy		±1°C		±2°C
Measurement Range		0°C		50°C
Response Time (Seconds)	1/e(63%)	6 S		30 S

3. Typical Application (Figure 1)

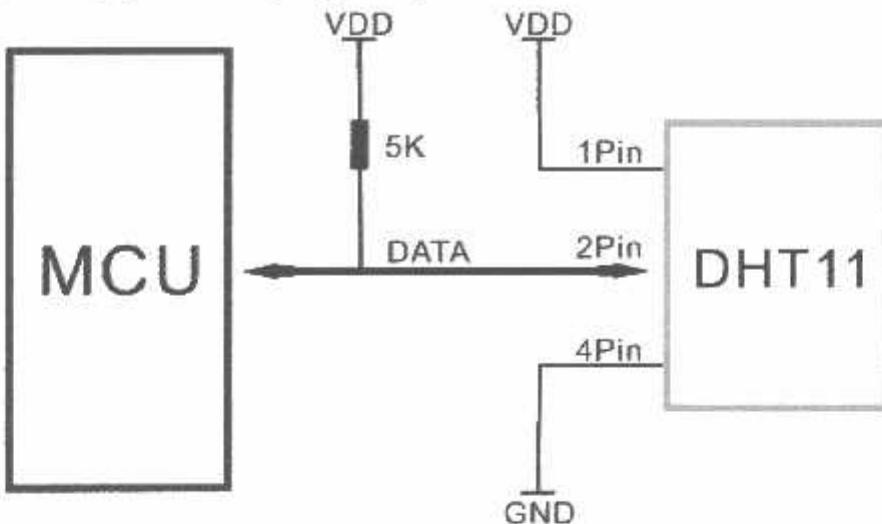


Figure 1 Typical Application

Note: 3Pin – Null; MCU = Micro-computer Unite or single chip Computer

When the connecting cable is shorter than 20 metres, a 5K pull-up resistor is recommended; when the connecting cable is longer than 20 metres, choose an appropriate pull-up resistor as needed.

4. Power and Pin

DHT11's power supply is 3-5.5V DC. When power is supplied to the sensor, do not send any instruction to the sensor in within one second in order to pass the unstable status. One capacitor valued 100nF can be added between VDD and GND for power filtering.

5. Communication Process: Serial Interface (Single-Wire Two-Way)

Single-bus data format is used for communication and synchronization between MCU and DHT11 sensor. One communication process is about 4ms.

Data consists of decimal and integral parts. A complete data transmission is 40bit, and the sensor sends higher data bit first.

Data format: 8bit integral RH data + 8bit decimal RH data + 8bit integral T data + 8bit decimal T data + 8bit check sum. If the data transmission is right, the check-sum should be the last 8bit of "8bit integral RH data + 8bit decimal RH data + 8bit integral T data + 8bit decimal T data".

5.1 Overall Communication Process (Figure 2, below)

When MCU sends a start signal, DHT11 changes from the low-power-consumption mode to the running-mode, waiting for MCU completing the start signal. Once it is completed, DHT11 sends a response signal of 40-bit data that include the relative humidity and temperature information to MCU. Users can choose to collect (read) some data. Without the start signal from MCU, DHT11 will not give the response signal to MCU. Once data is collected, DHT11 will change to the low-power-consumption mode until it receives a start signal from MCU again.

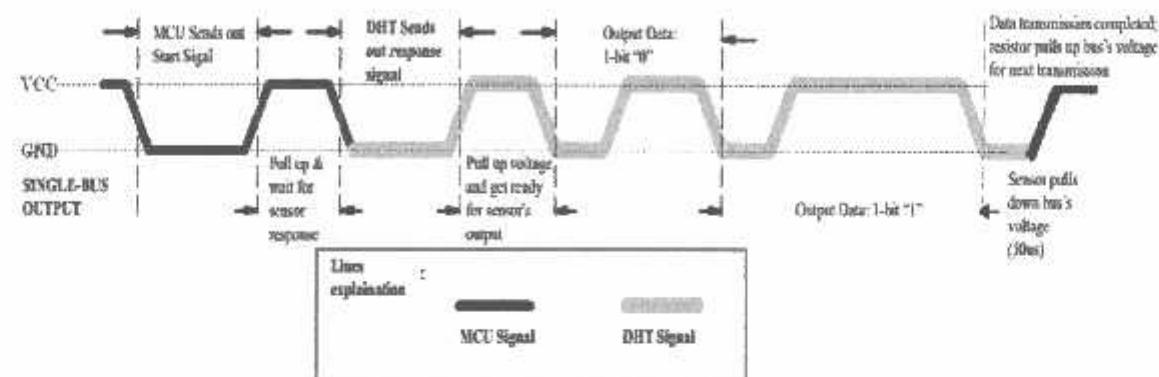


Figure 2 Overall Communication Process

5.2 MCU Sends out Start Signal to DHT (Figure 3, below)

Data Single-bus free status is at high voltage level. When the communication between MCU and DHT11 begins, the programme of MCU will set Data Single-bus voltage level from high to low and this process must take at least 18ms to ensure DHT's detection of MCU's signal, then MCU will pull up voltage and wait 20-40us for DHT's response.

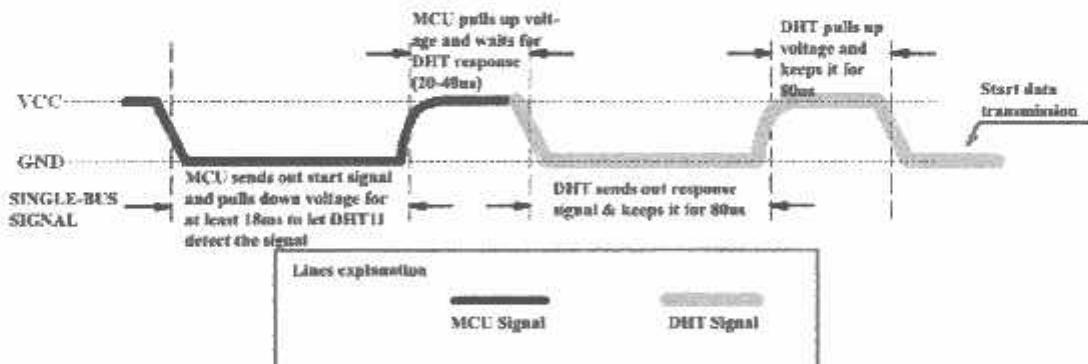


Figure 3 MCU Sends out Start Signal & DHT Responses

5.3 DHT Responses to MCU (Figure 3, above)

Once DHT detects the start signal, it will send out a low-voltage-level response signal, which lasts 80us. Then the programme of DHT sets Data Single-bus voltage level from low to high and keeps it for 80us for DHT's preparation for sending data.

When DATA Single-Bus is at the low voltage level, this means that DHT is sending the response signal. Once DHT sent out the response signal, it pulls up voltage and keeps it for 80us and prepares for data transmission.

When DHT is sending data to MCU, every bit of data begins with the 50us low-voltage-level and the length of the following high-voltage-level signal determines whether data bit is "0" or "1" (see Figures 4 and 5 below).

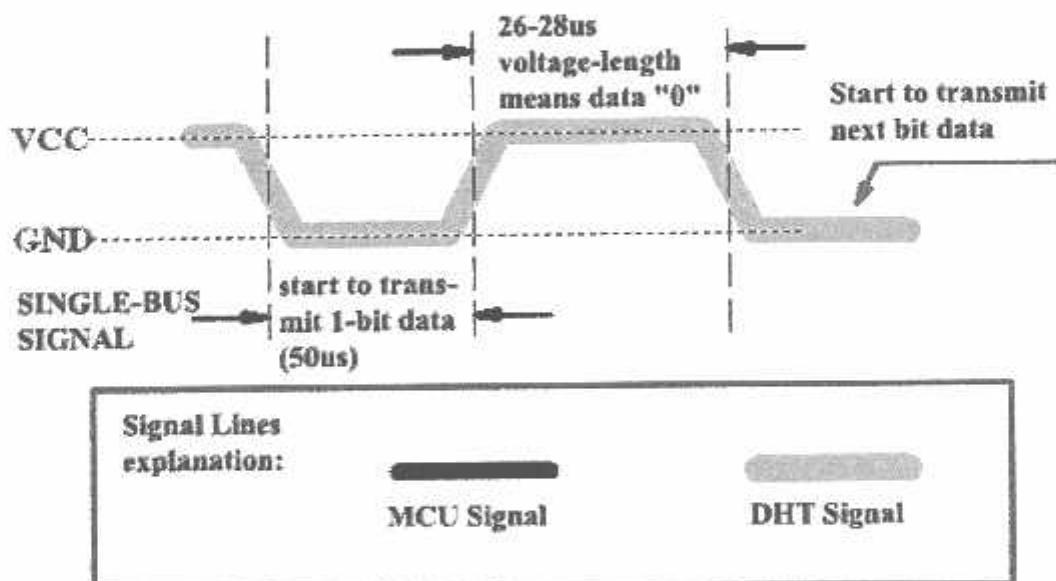


Figure 4 Data "0" Indication

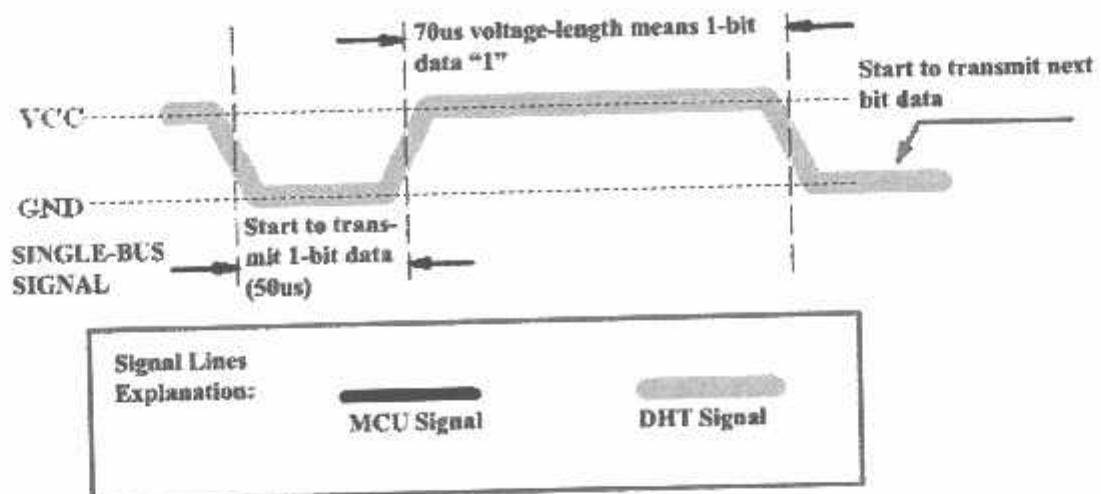


Figure 5 Data "1" Indication

If the response signal from DHT is always at high-voltage-level, it suggests that DHT is not responding properly and please check the connection. When the last bit data is transmitted, DHT11 pulls down the voltage level and keeps it for 50us. Then the Single-Bus voltage will be pulled up by the resistor to set it back to the free status.

6. Electrical Characteristics

VDD=5V, T = 25°C (unless otherwise stated)

	Conditions	Minimum	Typical	Maximum
Power Supply	DC	3V	5V	5.5V
Current Supply	Measuring	0.5mA		2.5mA
	Average	0.2mA		1mA
	Standby	100uA		150uA
Sampling period	Second	1		

Note: Sampling period at intervals should be no less than 1 second.

7. Attentions of application

(1) Operating conditions

Applying the DHT11 sensor beyond its working range stated in this datasheet can result in 3%RH signal shift/discrepancy. The DHT11 sensor can recover to the calibrated status gradually when it gets back to the normal operating condition and works within its range. Please refer to (3) of

this section to accelerate its recovery. Please be aware that operating the DHT11 sensor in the non-normal working conditions will accelerate sensor's aging process.

(2) Attention to chemical materials

Vapor from chemical materials may interfere with DHT's sensitive-elements and debase its sensitivity. A high degree of chemical contamination can permanently damage the sensor.

(3) Restoration process when (1) & (2) happen

Step one: Keep the DHT sensor at the condition of Temperature 50~60Celsius, humidity <10%RH for 2 hours;

Step two: Keep the DHT sensor at the condition of Temperature 20~30Celsius, humidity >70%RH for 5 hours.

(4) Temperature Affect

Relative humidity largely depends on temperature. Although temperature compensation technology is used to ensure accurate measurement of RH, it is still strongly advised to keep the humidity and temperature sensors working under the same temperature. DHT11 should be mounted at the place as far as possible from parts that may generate heat.

(5) Light Affect

Long time exposure to strong sunlight and ultraviolet may debase DHT's performance.

(6) Connection wires

The quality of connection wires will affect the quality and distance of communication and high quality shielding-wire is recommended.

(7) Other attentions

* Welding temperature should be below 260Celsius and contact should take less than 10 seconds.

* Avoid using the sensor under dew condition.

* Do not use this product in safety or emergency stop devices or any other occasion that failure of DHT11 may cause personal injury.

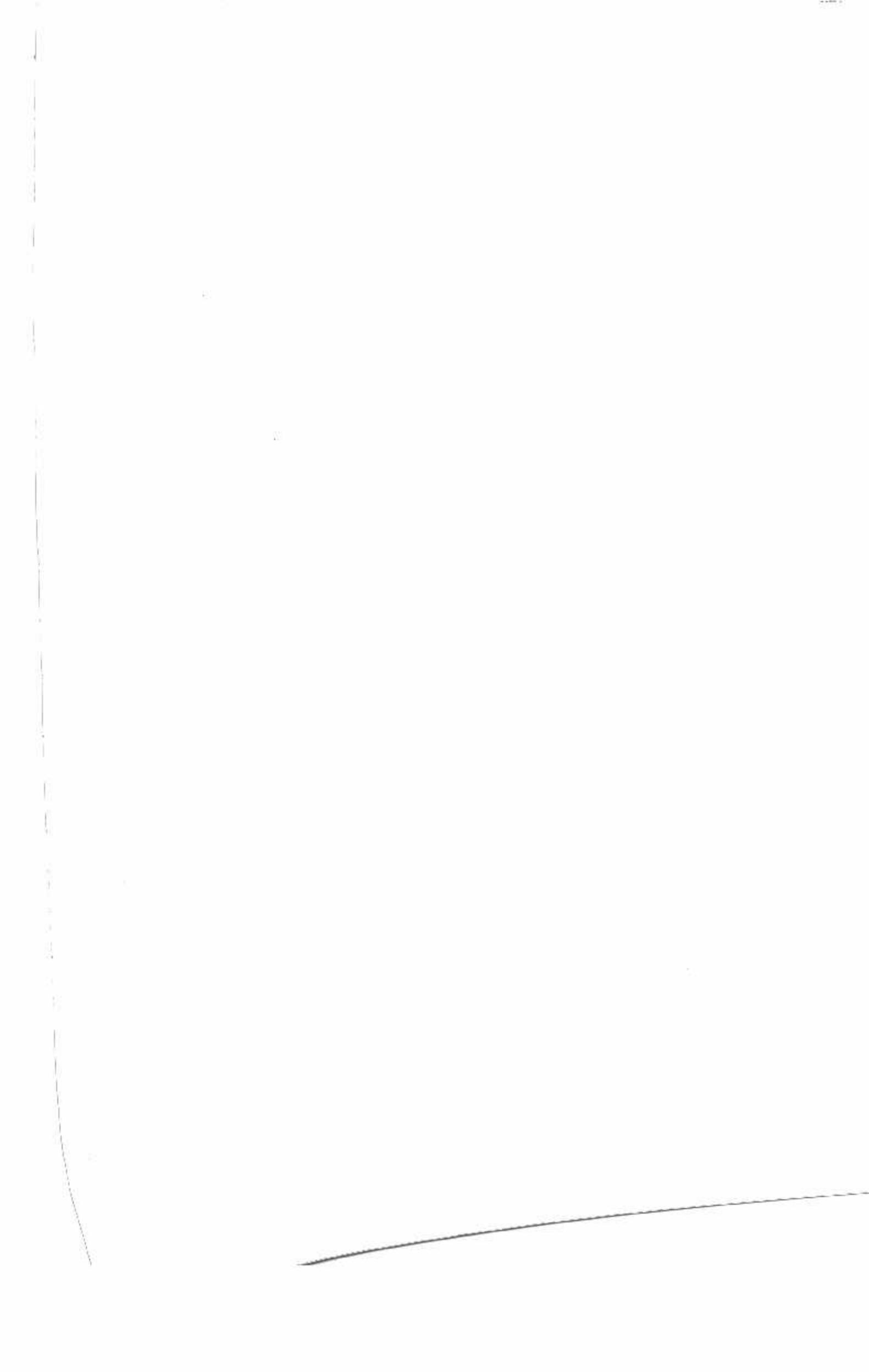
* Storage: Keep the sensor at temperature 10~40°C, humidity <60%RH.

Disclaimer:

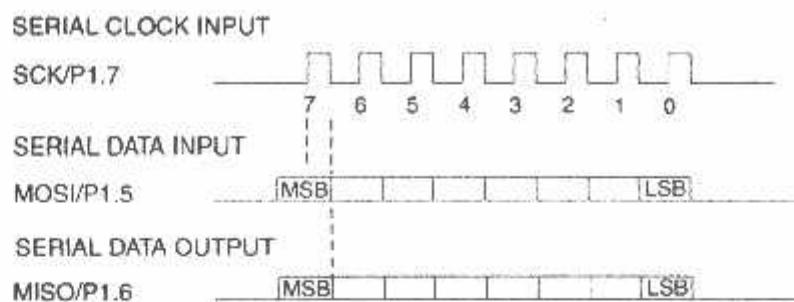
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Serial Downloading Waveforms



Serial Programming Characteristics

Figure 16. Serial Programming Timing

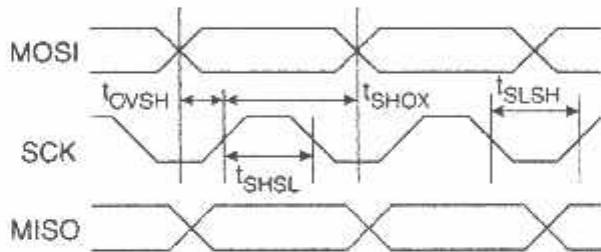


Table 11. Serial Programming Characteristics, $T_A = -40^\circ\text{C}$ to 85°C , $V_{CC} = 4.0$ - 6.0V (Unless Otherwise Noted)

Symbol	Parameter	Min	Typ	Max	Units
f_{CLCL}	Oscillator Frequency	0		24	MHz
t_{CLCL}	Oscillator Period	41.6			ns
t_{SHSL}	SCK Pulse Width High	24 t_{CLCL}			ns
t_{SLSH}	SCK Pulse Width Low	24 t_{CLCL}			ns
t_{OVSH}	MOSI Setup to SCK High	t_{CLCL}			ns
t_{SHOX}	MOSI Hold after SCK High	2 t_{CLCL}			ns

Absolute Maximum Ratings*

Operating Temperature	-55°C to +125°C
Storage Temperature	-65°C to +150°C
Voltage on Any Pin with Respect to Ground	-1.0V to +7.0V
Maximum Operating Voltage	6.6V
DC Output Current	15.0 mA

*NOTICE: Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

DC Characteristics

The values shown in this table are valid for $T_A = -40^\circ\text{C}$ to 85°C and $V_{CC} = 5.0\text{V} \pm 20\%$, unless otherwise noted.

Symbol	Parameter	Condition	Min	Max	Units
V_{IL}	Input Low-voltage	(Except EA)	-0.5	0.2 V_{CC} - 0.1	V
V_{IL1}	Input Low-voltage (EA)		0.5	0.2 V_{CC} - 0.3	V
V_{IH}	Input High-voltage	(Except XTAL1, RST)	0.2 V_{CC} + 0.9	V_{CC} + 0.5	V
V_{IH1}	Input High-voltage	(XTAL1, RST)	0.7 V_{CC}	V_{CC} + 0.5	V
V_{OL}	Output Low-voltage ⁽¹⁾ (Ports 1,2,3)	$I_{OL} = 1.6\text{ mA}$		0.5	V
V_{OL0}	Output Low-voltage ⁽¹⁾ (Port 0, ALE, PSEN)	$I_{OL} = 3.2\text{ mA}$		0.5	V
V_{OH}	Output High-voltage (Ports 1,2,3, ALE, PSEN)	$I_{OH} = -60\text{ }\mu\text{A}, V_{CC} = 5\text{V} \pm 10\%$ $I_{OH} = -25\text{ }\mu\text{A}$ $I_{OH} = -10\text{ }\mu\text{A}$	2.4 0.75 V_{CC} 0.9 V_{CC}		V
V_{OHI}	Output High-voltage (Port 0 In External Bus Mode)	$I_{OH} = -800\text{ }\mu\text{A}, V_{CC} = 5\text{V} \pm 10\%$ $I_{OH} = -300\text{ }\mu\text{A}$ $I_{OH} = -80\text{ }\mu\text{A}$	2.4 0.75 V_{CC} 0.9 V_{CC}		V
I_L	Logical 0 Input Current (Ports 1,2,3)	$V_{IN} = 0.45\text{V}$		-50	μA
I_T	Logical 1 to 0 Transition Current (Ports 1,2,3)	$V_{IN} = 2\text{V}, V_{CC} = 5\text{V} \pm 10\%$		-650	μA
I_{L0}	Input Leakage Current (Port 0, EA)	$0.45 < V_{IN} < V_{CC}$		+10	μA
RRST	Reset Pull-down Resistor		50	300	K Ω
C_{IO}	Pin Capacitance	Test Freq. = 1 MHz, $T_A = 25^\circ\text{C}$		10	pF
I_{CC}	Power Supply Current	Active Mode, 12 MHz		25	mA
		Idle Mode, 12 MHz		8.5	mA
	Power-down Mode ⁽²⁾	$V_{CC} = 6\text{V}$		100	μA
		$V_{CC} = 3\text{V}$		40	μA

- Notes:
- Under steady state (non-transient) conditions, I_{OL} must be externally limited as follows:
Maximum I_{OL} per port pin: 10 mA
Maximum I_{OL} per 8-bit port, Port 0: 26 mA; Ports 1, 2, 3: 15 mA
Maximum total I_{OL} for all output pins: 71 mA
If I_{OL} exceeds the test condition, V_{OL} may exceed the related specification. Pins are not guaranteed to sink current greater than the listed test conditions.
 - Minimum V_{CC} for Power-down is 2V



AC Characteristics

Under operating conditions, load capacitance for Port 0, ALE/PROG, and PSEN = 100 pF; load capacitance for all other outputs = 80 pF.

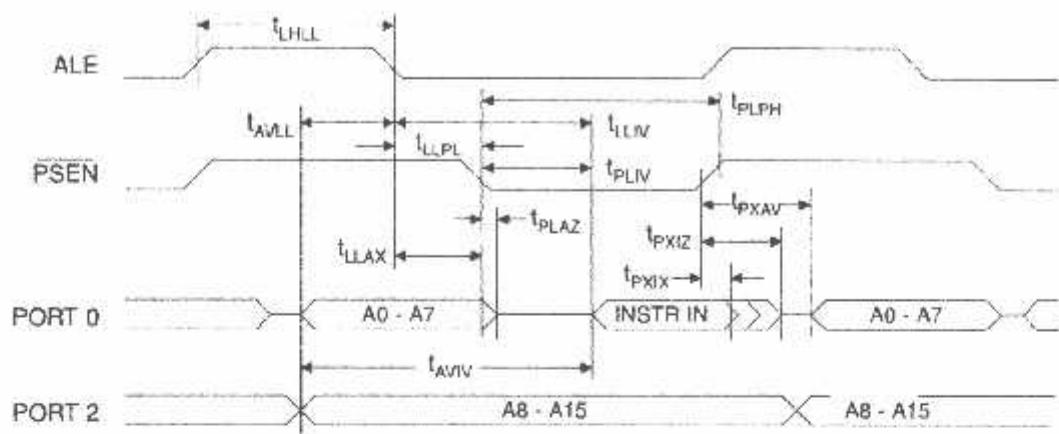
External Program and Data Memory Characteristics

Symbol	Parameter	Variable Oscillator		
		Min	Max	Units
t_{CLCL}	Oscillator Frequency	0	24	MHz
t_{UHL}	ALE Pulse Width	$2t_{CLCL} + 40$		ns
t_{AVLL}	Address Valid to ALE Low	$t_{CLCL} - 13$		ns
t_{LAX}	Address Hold after ALE Low	$t_{CLCL} - 20$		ns
t_{LLV}	ALE Low to Valid Instruction In		$4t_{CLCL} - 65$	ns
t_{LPL}	ALE Low to PSEN Low	$t_{CLCL} - 13$		ns
t_{PLPH}	PSEN Pulse Width	$3t_{CLCL} - 20$		ns
t_{PLV}	PSEN Low to Valid Instruction In		$3t_{CLCL} - 45$	ns
t_{PXIX}	Input Instruction Hold after PSEN	0		ns
t_{PXIZ}	Input Instruction Float after PSEN		$t_{CLCL} - 10$	ns
t_{PXAV}	PSEN to Address Valid	$t_{CLCL} - 8$		ns
t_{AVV}	Address to Valid Instruction In		$5t_{CLCL} - 55$	ns
t_{PLAZ}	PSEN Low to Address Float		10	ns
t_{PLRH}	RD Pulse Width	$6t_{CLCL} - 100$		ns
t_{WLWH}	WR Pulse Width	$6t_{CLCL} - 100$		ns
t_{RLDV}	RD Low to Valid Data In		$5t_{CLCL} - 90$	ns
t_{RHDX}	Data Hold after RD	0		ns
t_{RHDZ}	Data Float after RD		$2t_{CLCL} - 28$	ns
t_{LLDV}	ALE Low to Valid Data In		$8t_{CLCL} - 150$	ns
t_{AVDV}	Address to Valid Data In		$9t_{CLCL} - 165$	ns
t_{LWML}	ALE Low to RD or WR Low	$3t_{CLCL} - 50$	$3t_{CLCL} + 50$	ns
t_{AVWL}	Address to RD or WR Low	$4t_{CLCL} - 75$		ns
t_{CVWX}	Data Valid to WR Transition	$t_{CLCL} - 20$		ns
t_{CVWH}	Data Valid to WR High	$7t_{CLCL} - 120$		ns
t_{WHQX}	Data Hold after WR	$t_{CLCL} - 20$		ns
t_{PLAZ}	RD Low to Address Float		0	ns
t_{WHJH}	RD or WR High to ALE High	$t_{CLCL} - 20$	$t_{CLCL} + 25$	ns

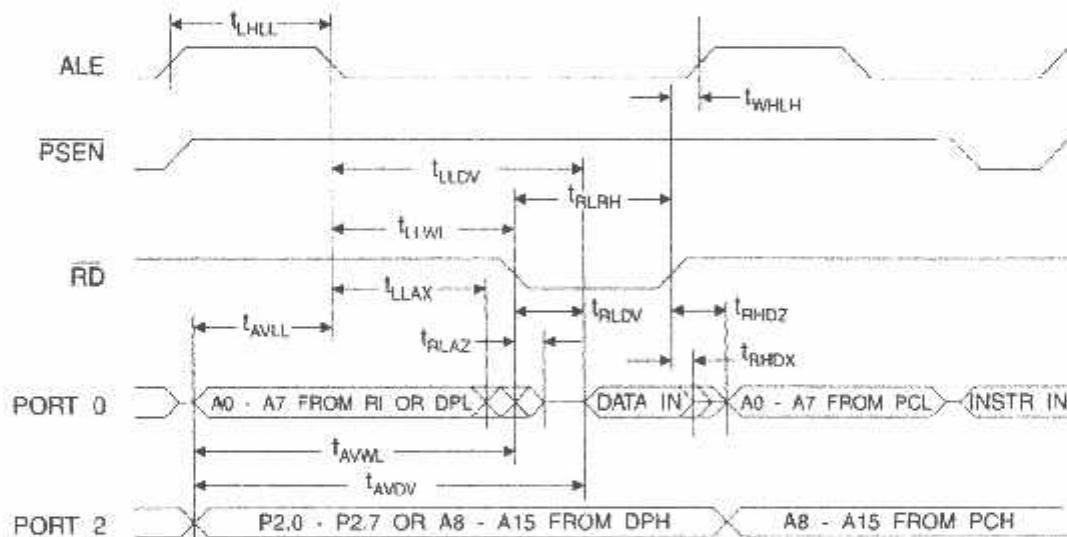
AT89S8252

D401G-MICRO-3/06

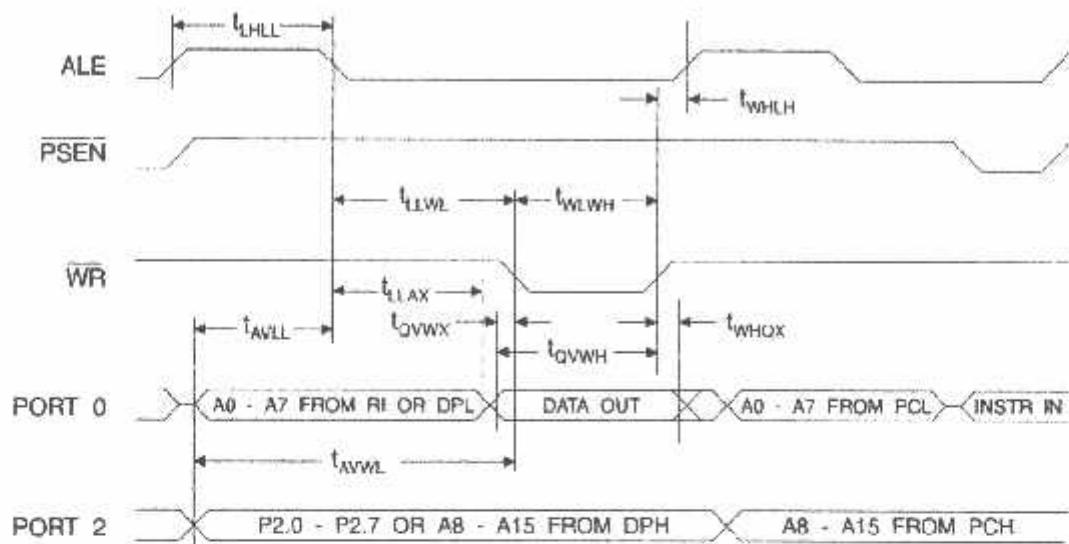
External Program Memory Read Cycle



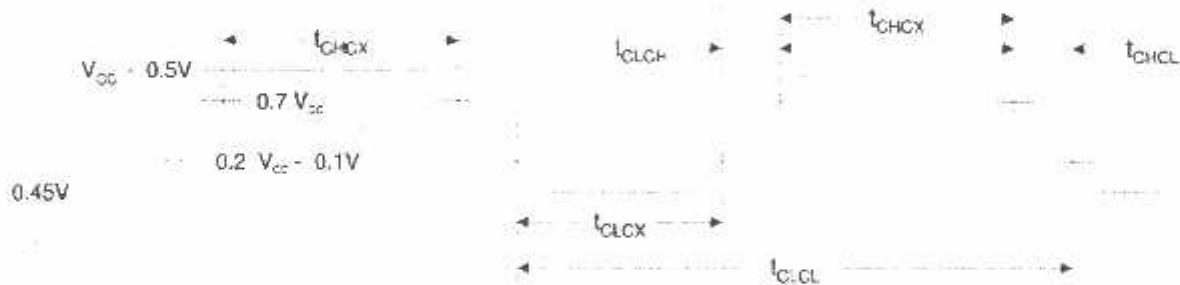
External Data Memory Read Cycle



External Data Memory Write Cycle



External Clock Drive Waveforms



External Clock Drive

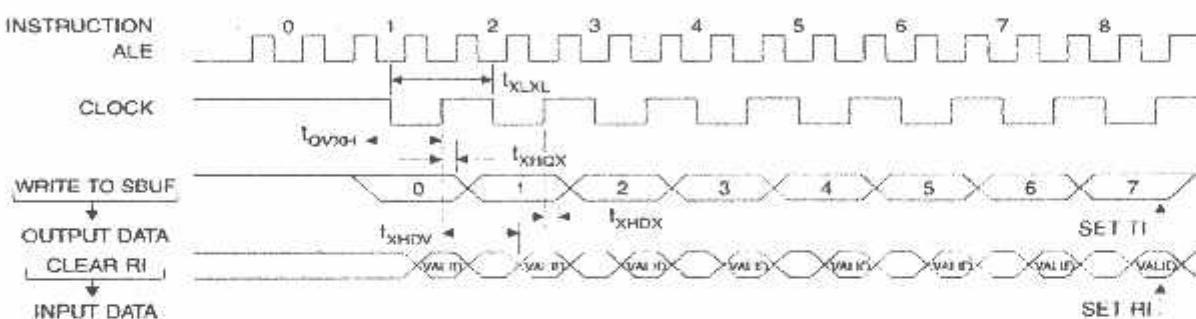
Symbol	Parameter	$V_{CC} = 4.0V \text{ to } 6.0V$		
		Min	Max	Units
$1/t_{CCLK}$	Oscillator Frequency	0	24	MHz
t_{CLCH}	Clock Period	41.6		ns
t_{CHCX}	High Time	15		ns
t_{CCLK}	Low Time	15		ns
t_{CHL}	Rise Time		20	ns
t_{CCL}	Fall Time		20	ns

Serial Port Timing: Shift Register Mode Test Conditions

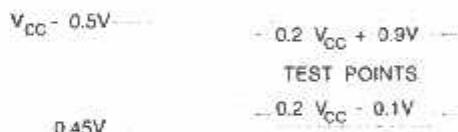
The values in this table are valid for $V_{CC} = 4.0V$ to $6V$ and Load Capacitance = 80 pF .

Symbol	Parameter	Variable Oscillator		Units
		Min	Max	
t_{XLXL}	Serial Port Clock Cycle Time	$12t_{CLCL}$		μs
t_{OVXH}	Output Data Setup to Clock Rising Edge	$10t_{CLCL} - 133$		ns
t_{XHOX}	Output Data Hold after Clock Rising Edge	$2t_{CLCL} - 117$		ns
t_{XHDX}	Input Data Hold after Clock Rising Edge	0		ns
t_{XHOV}	Clock Rising Edge to Input Data Valid		$10t_{CLCL} - 133$	ns

Shift Register Mode Timing Waveforms

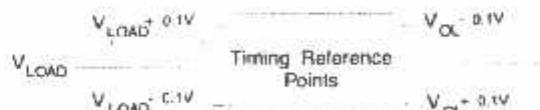


AC Testing Input/Output Waveforms⁽¹⁾



Note: 1. AC Inputs during testing are driven at $V_{CC} - 0.5V$ for a logic 1 and $0.45V$ for a logic 0. Timing measurements are made at V_{IH} min. for a logic 1 and V_{IL} max. for a logic 0.

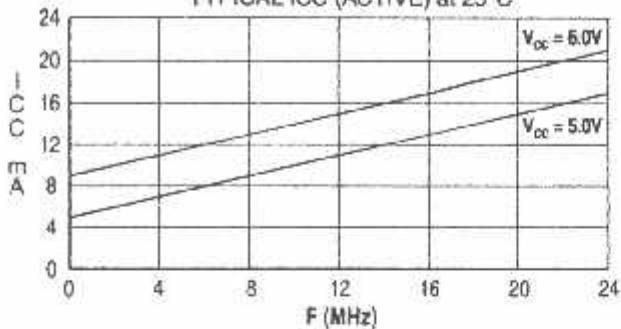
Float Waveforms⁽¹⁾



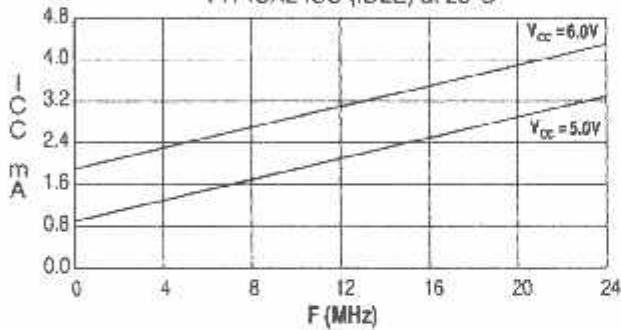
Note: 1. For timing purposes, a port pin is no longer floating when a 100 mV change from load voltage occurs. A port pin begins to float when a 100 mV change from the loaded V_{OL}/V_{OL} level occurs.

AT89S8252

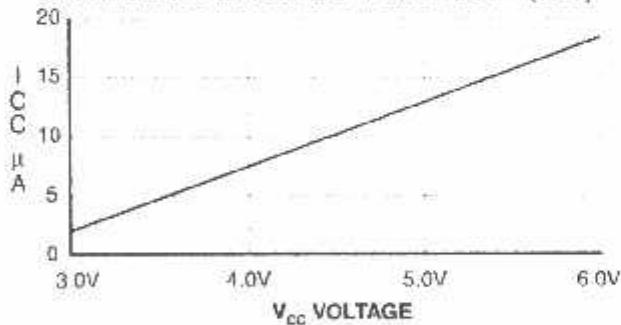
TYPICAL ICC (ACTIVE) at 25°C

**AT89S8252**

TYPICAL ICC (IDLE) at 25°C

**AT89S8252**

TYPICAL ICC vs. VOLTAGE - POWER DOWN (85°C)



Notes:

1. XTAL1 tied to GND for Icc (power-down)
2. Lock bits programmed

Ordering Information

Speed (MHz)	Power Supply	Ordering Code	Package	Operation Range
24	4.0V to 6.0V	AT89S8252-24AC	44A	Commercial (0°C to 70°C)
		AT89S8252-24JC	44J	
		AT89S8252-24PC	40P6	
	4.0V to 6.0V	AT89S8252-24AI	44A	Industrial (-40°C to 85°C)
		AT89S8252-24JI	44J	
		AT89S8252-24PI	40P6	

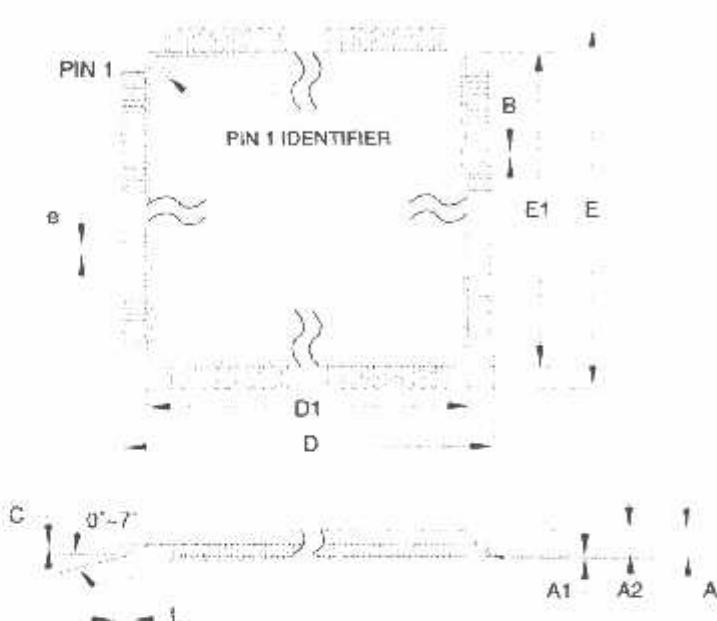
Package Type

4A	44-lead, Thin Plastic Gull Wing Quad Flatpack (TQFP)
J	44-lead, Plastic J-leaded Chip Carrier (PLCC)
P6	40-lead, 0.600" Wide, Plastic Dual Inline Package (PDIP)



Packaging Information

44A – TQFP



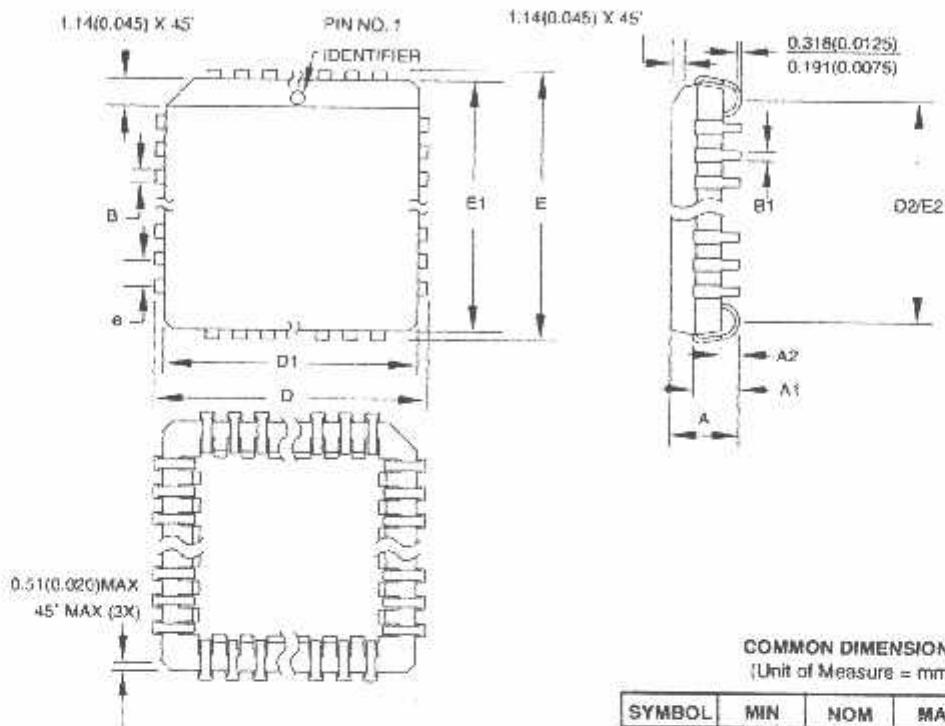
COMMON DIMENSIONS
(Unit of Measure = mm)

SYMBOL	MIN	NOM	MAX	NOTE
A	—	—	1.20	
A1	0.05	—	0.15	
A2	0.95	1.00	1.05	
D	11.75	12.00	12.25	
D1	9.90	10.00	10.10	Note 2
E	11.75	12.00	12.25	
E1	9.90	10.00	10.10	Note 2
B	0.30	—	0.45	
C	0.09	—	0.20	
L	0.45	—	0.75	
e		0.80 TYP		

10/5/2001

2325 Orchard Parkway San Jose, CA 95131	TITLE 44A, 44-lead, 10 x 10 mm Body Size, 1.0 mm Body Thickness, 0.8 mm Lead Pitch, Thin Profile Plastic Quad Flat Package (TQFP)	DRAWING NO.	REV.
		44A	B

44J - PLCC



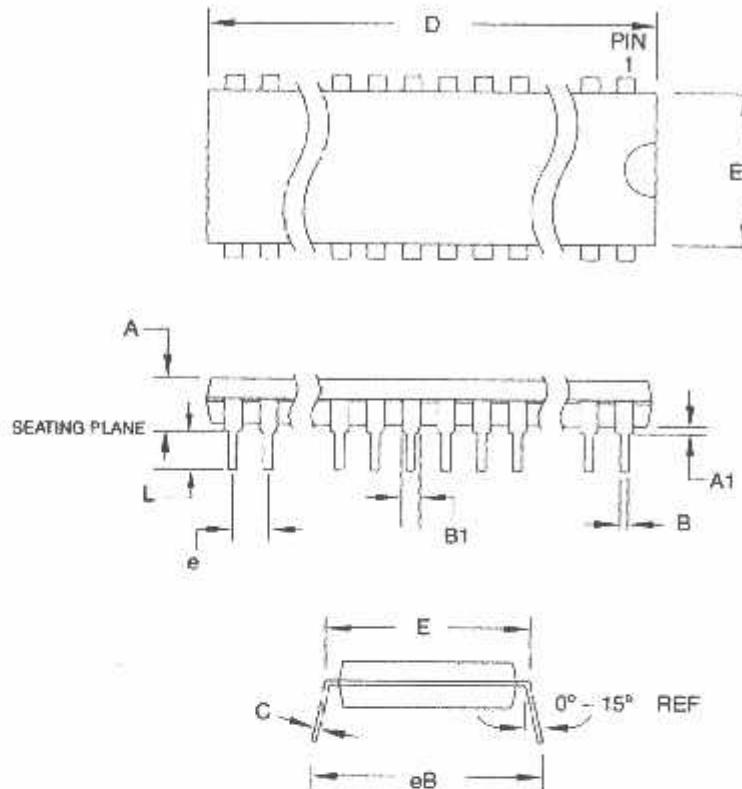
- Notes:
1. This package conforms to JEDEC reference MS-018, Variation AC.
 2. Dimensions D1 and E1 do not include mold protrusion.
Allowable protrusion is 0.10" (0.254 mm) per side. Dimension D1 and E1 include mold mismatch and are measured at the extreme material condition at the upper or lower parting line.
 3. Lead coplanarity is 0.004" (0.102 mm) maximum.

COMMON DIMENSIONS
(Unit of Measure = mm)

SYMBOL	MIN	NOM	MAX	NOTE
A	4.191	—	4.572	
A1	2.286	—	3.048	
A2	0.508	—	—	
D	17.399	—	17.653	
D1	16.510	—	16.662	Note 2
E	17.399	—	17.653	
E1	16.510	—	16.662	Note 2
D2/E2	14.986	—	16.002	
B	0.660	—	0.813	
B1	0.330	—	0.533	
B	1.270 TYP			

10/04/01

ATMEL 2325 Orchard Parkway San Jose, CA 95131	TITLE 44J, 44-lead, Plastic J-leaded Chip Carrier (PLCC)	DRAWING NO. 44J	REV. B
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40P6 – PDIP


- Notes:**
1. This package conforms to JEDEC reference MS-011, Variation AC.
 2. Dimensions D and E1 do not include mold Flash or Protrusion.
Mold Flash or Protrusion shall not exceed 0.25 mm (0.010").

COMMON DIMENSIONS
(Unit of Measure = mm)

SYMBOL	MIN	NOM	MAX	NOTE
A	—	—	4.826	
A1	0.381	—	—	
D	52.070	—	52.578	Note 2
E	15.240	—	15.875	
E1	13.462	—	13.970	Note 2
B	0.356	—	0.559	
B1	1.041	—	1.651	
L	3.048	—	3.556	
C	0.203	—	0.381	
eB	15.494	—	17.526	
e	2.540 TYP			

09/28/01

 2325 Orchard Parkway San Jose, CA 95131	TITLE 40P6, 40-lead (0.600"/15.24 mm Wide) Plastic Dual Inline Package (PDIP)	DRAWING NO. 40P6	REV. B
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Fax: (33) 2-40-18-19-60

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Scottish Enterprise Technology Park
Maxwell Building
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Tel: (44) 1355-803-000
Fax: (44) 1355-242-743

RF/Automotive

Theresienstrasse 2
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74026 Heilbronn, Germany
Tel: (49) 71-31-67-0
Fax: (49) 71-31-67-2340

1150 East Cheyenne Mtn. Blvd.
Colorado Springs, CO 80906, USA
Tel: (719) 576-3300
Fax: (719) 540-1759

Biometrics/Imaging/Hi-Rel MPU/

High-Speed Converters/RF Datacom
Avenue de Rochepleine
BP 123
38521 Saint-Egreve Cedex, France
Tel: (33) 4-76-58-30-00
Fax: (33) 4-76-58-34-80

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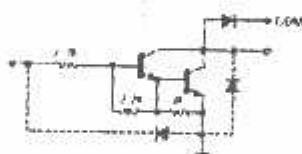
de KITS Application Note

AN1 - Relay Board

Oleh: Tim IE

Hardware

Relay merupakan salah satu komponen output yang paling sering digunakan baik pada industri, otomotif, ataupun peralatan elektronika lainnya. Relay berfungsi untuk menghubungkan atau memulus aliran arus listrik yang dikontrol dengan memberikan tegangan dan arus tertentu pada koilnya. Ada 2 macam relay berdasarkan tegangan untuk menggerakkan koilnya yaitu AC dan DC. Pada relay board ini digunakan relay DC dengan tegangan koil 12V DC, arus yang diperlukan sekitar 20-30mA. Karena itu pada umumnya kita tidak bisa langsung menghubungkan output suatu IC logic (TTL/CMOS) atau peripheral lain seperti µC 89C51, PPI 82C55 dengan relay karena I_{Omax} (arus maximum yang dikeluarkan pada saat logic '1') atau I_{OLmax} (arus maximum yang mampu dibenamkan pada saat logic '0') tidak cukup besar. Karena itu perlu digunakan driver untuk penguatan arus yang biasanya berupa transistor, di sini digunakan "Darlington Array" ULN 2803A yang merupakan sekumpulan transistor dengan konfigurasi Darlington sehingga mempunyai β (penguatan arus) yang besar (lihat gambar 1).



Gambar 1. Konfigurasi Output ULN2803A

Setiap output pada ULN 2803A dapat dibebani sampai 0mA, serta dilengkapi dengan 'supression diode'. Iode ini berfungsi untuk mencegah 'kickback' yaitu transient yang terjadi pada koil relay (beban induktif) saat relay matikan. Fenomena ini bisa dianalisa dari rumus berikut :

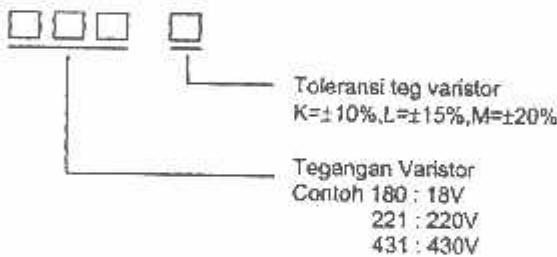
$$V = -L \frac{di}{dt}$$

yang akan terjadi perubahan arus yang cukup besar dalam setuan waktu yang sangat cepat $dt = 0$ maka tegangan balik ini menjadi sangat besar, dan dapat menyebabkan kerusakan pada transistor. Problem lain yang sering terjadi pada kontak relay adalah ioncatan api listrik yang dapat memperpendek umur kontak. Api ini terutama terjadi pada beban induktif seperti motor, solenoid, dll. Untuk mencegah hal ini digunakan MOV (Metallic Oxide Varistor) yang dipasang secara paralel dengan kontak.

Varistor bersifat seperti resistor dengan nilai resistansinya tergantung pada tegangan. Ketika kontak terbuka, beban induktif menghasilkan tegangan balik yang cukup besar akibat perubahan medan magnet. Pada saat ini ('protective state') nilai resistansi varistor menjadi sangat kecil dan arus akan mengalir melalui MOV, sehingga transient dapat direndah. Pada saat keadaan normal resistansi MOV sangat besar dan hanya menarik arus yang sangat kecil, bunga api juga menyebabkan sinyal Radio Frequency Interference (RFI) yang dapat mengganggu peralatan – peralatan sensitif. Karena itu komponen perekam transient seperti MOV sangat diperlukan terutama pada beban induktif. Parameter-parameter penting Varistor :

- Tegangan Varistor (tegangan breakdown) :
Tegangan pada varistor yang diukur pada arus tertentu (0,1 mA atau 1 mA) selama waktu tertentu
- Tegangan maksimum yang diperbolehkan :
Tegangan maksimum pada varistor agar tetap pada keadaan normal ('rest state')

Cara membaca kode varistor pada umumnya



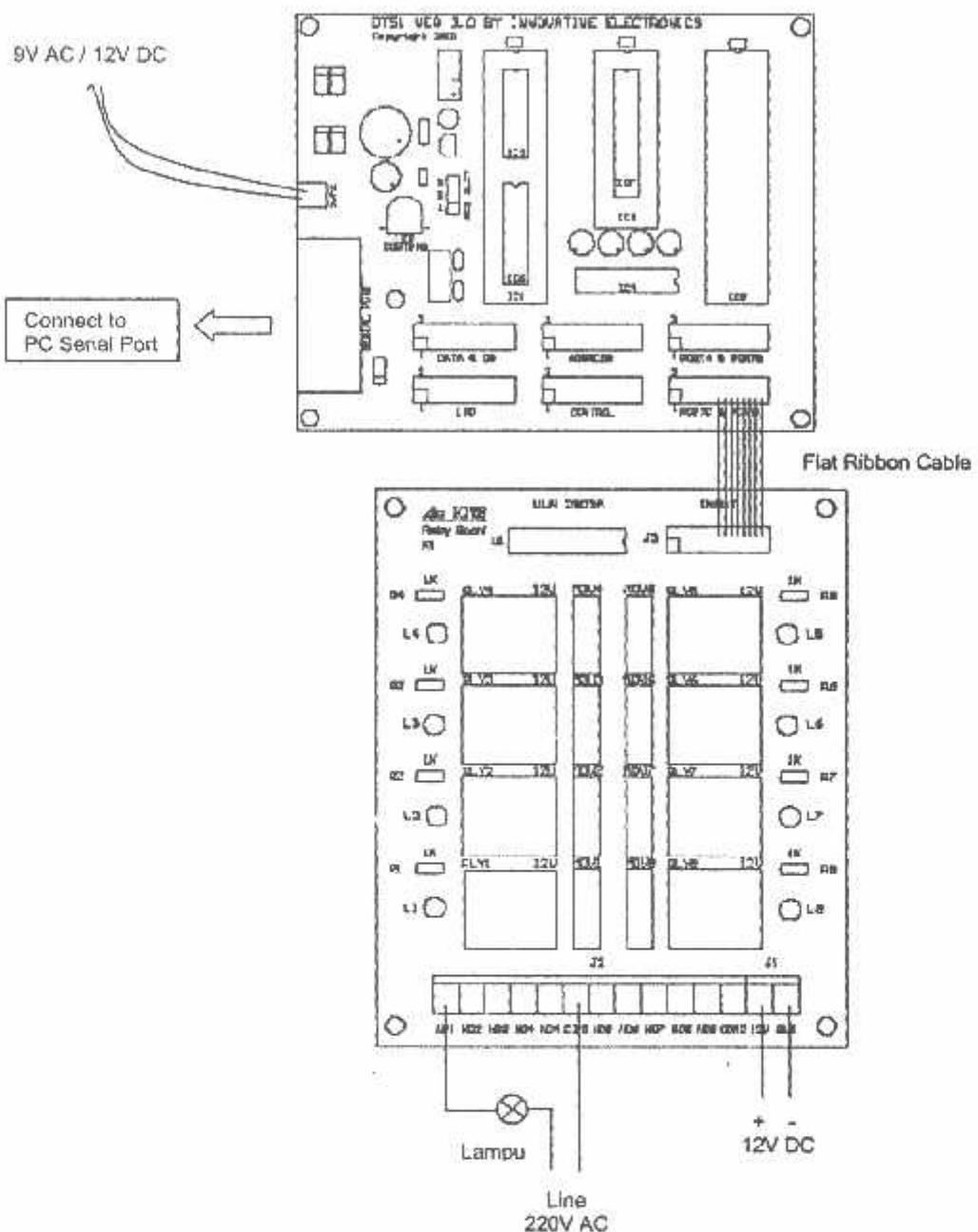
Pemilihan tegangan varistor tergantung pada tegangan beban yang dihubungkan pada relay.

Contoh :

Tegangan Supply	Tipe Varistor
12V DC	220K
24V DC	390K
110V AC	201K, 221K
220V AC	431K, 471K

Software

Pemrograman untuk mengendalikan relay board ini sangat sederhana. Bila menggunakan DT-51® Development Tools kita bisa memilih apakah menggunakan Port 1 (P1.0 ~ P1.7) atau Port A, B, dan C dari PPI (Programmable Peripheral Interface). Pada gambar 2 ditunjukkan salah satu contoh aplikasi relay board, yaitu untuk mematikan dan menghidupkan lampu bohlam 220V AC. Input Relay Board (IN1-IN8) dihubungkan dengan Port 1 DT-51® Development Tools (P1.0-P1.7).



Gambar 2. Hubungan DT-51 MinSys Ver 3.0 dengan Relay Board

Listing Program

Berikut ini listing program (dengan Assembler ASM51) untuk menyalaikan lampu bohlam selama ± 15 detik, kemudian padam ± 15 detik secara periodik.

```
$TITLE(Demo RELAY BOARD dengan lampu bohlam)
$MOD51
```

```
CSEG
```

```
ORG      4000H
LJMP    Start
ORG      4100H

Delay:    PUSH    06H
          PUSH    07H
          PUSH    05H
          PUSH    04H

          MOV     R4,#01H
DelayOutsLoop: MOV     R5,#0AAH
DelayOutLoop:  MOV     R6,#0BBH
DelayOuterLoop: MOV    R7,#0CCH
DelayInnerLoop: DJNZ   R7,$
                  DJNZ   R6,DelayOuterLoop
                  DJNZ   R5,DelayOutLoop
                  DJNZ   R4,DelayOutsLoop

          POP     04H
          POP     05H
          POP     07H
          POP     06H
RET

Start:    SETB   P1.0
ACALL   Delay
CLR     P1.0
ACALL   Delay
AJMP   Start
END
```

MC14049UB

Hex Buffers

The MC14049UB hex inverter/buffer is constructed with MOS P-channel and N-channel enhancement mode devices in a single monolithic structure. This complementary MOS device finds primary use where low power dissipation and/or high noise immunity is desired. This device provides logic-level conversion using only one supply voltage, V_{DD}. The input-signal high level (V_{IH}) can exceed the V_{DD} supply voltage for logic-level inversions. Two TTL/DTL Loads can be driven when the device is used as OS-to-TTL/DTL converters (V_{DD} = 5.0 V, V_{OL} ≤ 0.4 V, I_{OL} ≥ 3.2 mA). Note that pins 13 and 16 are not connected internally on this device; consequently connections to these terminals will not affect circuit operation.

High Source and Sink Currents

High-to-Low Level Converter

Supply Voltage Range = 3.0 V to 18 V

Meets JEDEC UB Specifications

V_{IN} can exceed V_{DD}

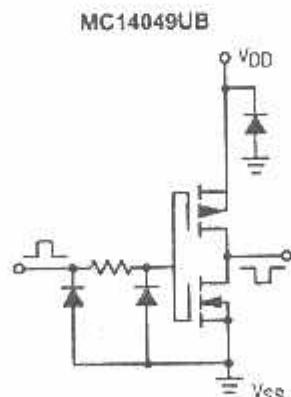
Improved ESD Protection on All Inputs

MAXIMUM RATINGS* (Voltages Referenced to V_{SS})

Rating	Symbol	Value	Unit
DC Supply Voltage	V _{DD}	–0.5 to + 18	V
Input Voltage (DC or Transient)	V _{IN}	–0.5 to + 18	V
Output Voltage (DC or Transient)	V _{OUT}	–0.5 to V _{DD} + 0.5	V
Input Current (DC or Transient), per Pin	I _{IN}	± 10	mA
Output Current (DC or Transient), per Pin	I _{OUT}	+ 45	mA
Power Dissipation, per Package†	P _D		mW
Plastic/Ceramic		825	
SOIC		740	
Storage Temperature	T _{STG}	–65 to + 150	°C
Lead Temperature (3-Second Soldering)	T _L	260	°C

Maximum Ratings are those values beyond which damage to the device may occur.
Temperature Derating: All Packages: See Figure 4.

CIRCUIT SCHEMATIC (1/6 OF CIRCUIT SHOWN)



L SUFFIX
CERAMIC
CASE 620



P SUFFIX
PLASTIC
CASE 848



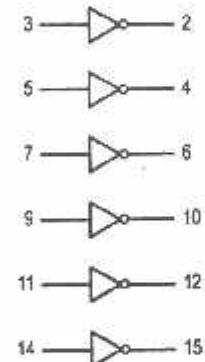
D SUFFIX
SOIC
CASE 751B

ORDERING INFORMATION

MC14XXXBCP	Plastic
MC14XXXBCL	Ceramic
MC14XXXBD	SOIC

T_A = –55° to 125°C for all packages.

LOGIC DIAGRAM MC14049UB



ELECTRICAL CHARACTERISTICS (Voltages Referenced to V_{SS})

Characteristic	Symbol	V _{DD} Vdc	-55°C		25°C			125°C		Unit
			Min	Max	Min	Typ	Max	Min	Max	
Input Voltage V _{in} = V _{DD} or 0	V _{OL}	5.0	—	0.05	—	0	0.05	—	0.05	Vdc
		10	—	0.05	—	0	0.05	—	0.05	Vdc
		15	—	0.05	—	0	0.05	—	0.05	Vdc
	V _{OH}	5.0	4.95	—	4.95	5.0	—	4.95	—	Vdc
		10	9.95	—	9.95	10	—	9.95	—	Vdc
		15	14.95	—	14.95	15	—	14.95	—	Vdc
Output Voltage (V _O = 4.5 Vdc) (V _O = 9.0 Vdc) (V _O = 13.5 Vdc)	V _{IL}	5.0	—	1.0	—	2.25	1.0	—	1.0	Vdc
		10	—	2.0	—	4.50	2.0	—	2.0	Vdc
		15	—	2.5	—	6.75	2.5	—	2.5	Vdc
	V _{IH}	5.0	4.0	—	4.0	2.75	—	4.0	—	Vdc
		10	9.0	—	8.0	5.50	—	8.0	—	Vdc
		15	12.5	—	12.5	8.25	—	12.5	—	Vdc
Output Drive Current (V _{OH} = 2.5 Vdc) (V _{OH} = 8.5 Vdc) (V _{OH} = 13.5 Vdc)	I _{OH}	5.0	-1.6	—	-1.25	-2.5	—	-1.0	—	mAdc
		10	-1.6	—	-1.3	-2.6	—	-1.0	—	mAdc
		15	-4.7	—	-3.75	-10	—	-3.0	—	mAdc
	I _{OL}	5.0	3.75	—	3.2	6.0	—	2.6	—	mAdc
		10	10	—	8.0	16	—	8.6	—	mAdc
		15	30	—	24	40	—	19	—	mAdc
Input Current	I _{in}	15	—	±0.1	—	±0.00001	±0.1	—	±1.0	μAdc
Input Capacitance (V _{in} = 0)	C _{in}	--	—	—	—	10	20	—	—	pF
Quiescent Current (Per Package)	I _{DD}	5.0	—	1.0	—	0.002	1.0	—	30	μAdc
Total Supply Current**†	I _T	5.0	I _T = (1.6 μA/kHz) f + I _{DD} I _T = (3.5 μA/kHz) f + I _{DD} I _T = (5.3 μA/kHz) f + I _{DD}						μAdc	
(Dynamic plus Quiescent, Per Package) (C _L = 50 pF on all outputs, all buffers switching)		10								
		15								

Data labelled "Typ" is not to be used for design purposes but is intended as an indication of the IC's potential performance.

* The formulas given are for the typical characteristics only at 25°C.

† To calculate total supply current at loads other than 50 pF:

$$I_T(C_L) = I_T(50 \text{ pF}) + (C_L - 50) Vfk$$

where: I_T is in μA (per package), C_L in pF, V = (V_{DD} - V_{SS}) in volts, f in kHz is input frequency, and k = 0.002.

ITCHING CHARACTERISTICS* ($C_L = 50 \text{ pF}$, $T_A = 25^\circ\text{C}$)

Characteristic	Symbol	V_{DD} Vdc	Min	Typ #	Max	Unit
Input Rise Time $t_{TLH} = (0.8 \text{ ns/pF}) C_L + 60 \text{ ns}$	t_{TLH}	5.0	—	100	160	ns
$t_{TLH} = (0.3 \text{ ns/pF}) C_L + 35 \text{ ns}$		10	—	50	100	
$t_{TLH} = (0.27 \text{ ns/pF}) C_L + 26.5 \text{ ns}$		15	—	40	60	
Input Fall Time $t_{THL} = (0.3 \text{ ns/pF}) C_L + 25 \text{ ns}$	t_{THL}	5.0	—	40	60	ns
$t_{THL} = (0.12 \text{ ns/pF}) C_L + 14 \text{ ns}$		10	—	20	40	
$t_{THL} = (0.1 \text{ ns/pF}) C_L + 10 \text{ ns}$		15	—	15	30	
Propagation Delay Time $t_{PLH} = (0.38 \text{ ns/pF}) C_L + 61 \text{ ns}$	t_{PLH}	5.0	—	80	120	ns
$t_{PLH} = (0.20 \text{ ns/pF}) C_L + 30 \text{ ns}$		10	—	40	65	
$t_{PLH} = (0.11 \text{ ns/pF}) C_L + 24.5 \text{ ns}$		15	—	30	50	
Propagation Delay Time $t_{PHL} = (0.38 \text{ ns/pF}) C_L + 11 \text{ ns}$	t_{PHL}	5.0	—	30	80	ns
$t_{PHL} = (0.12 \text{ ns/pF}) C_L + 9 \text{ ns}$		10	—	15	30	
$t_{PHL} = (0.11 \text{ ns/pF}) C_L + 4.5 \text{ ns}$		15	—	10	20	

The formulas given are for the typical characteristics only at 25°C .

The labelled "Typ" is not to be used for design purposes but is intended as an indication of the IC's potential performance.

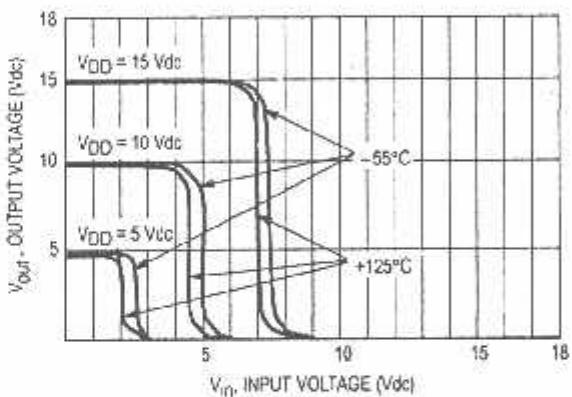
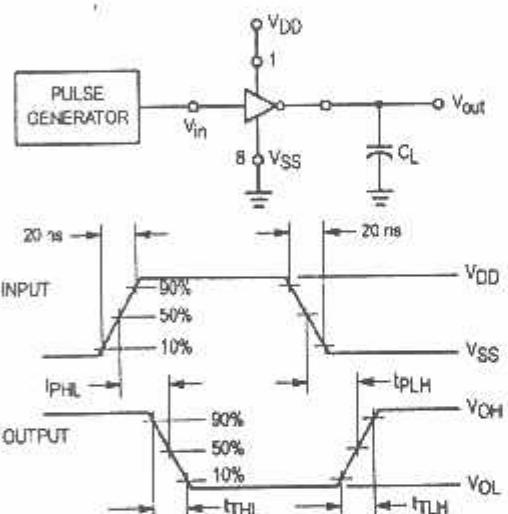
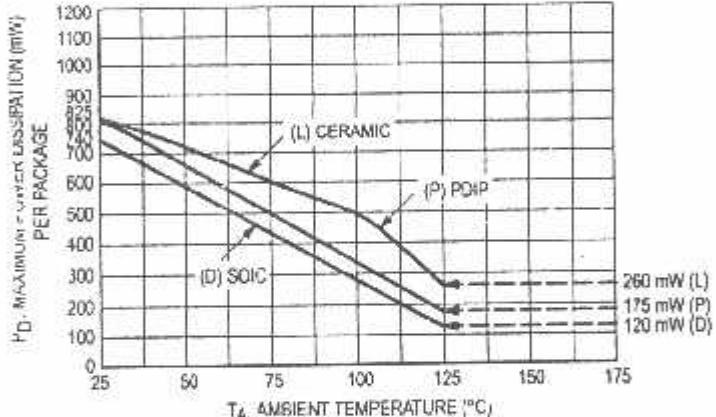
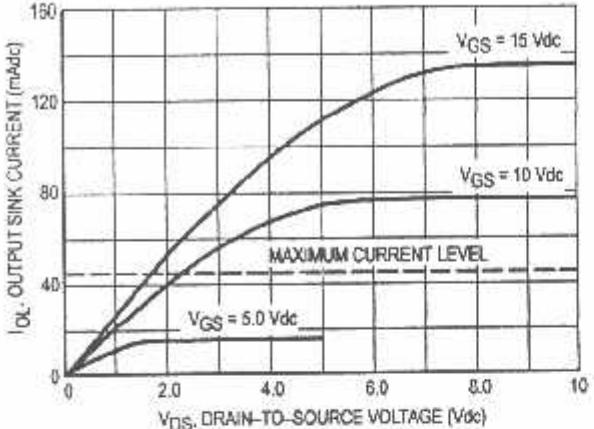
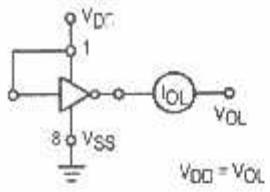
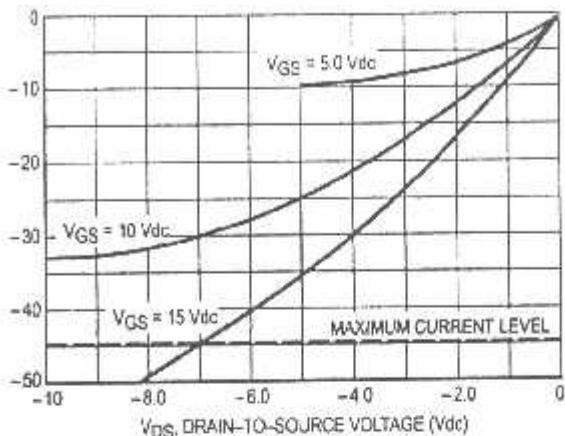
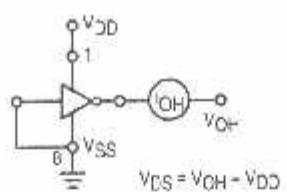


Figure 1. Typical Voltage Transfer Characteristics versus Temperature



This device contains circuitry to protect the inputs against damage due to high static voltages or electric fields referenced to the V_{SS} pin, only. Extra precautions must be taken to avoid applications of any voltage higher than the maximum rated voltages to this high-impedance circuit. For proper operation, the ranges V_{SS} ≤ V_{in} ≤ 18 V and V_{SS} ≤ V_{out} ≤ V_{DD} are recommended.

Unused inputs must always be tied to an appropriate logic voltage level (e.g., either V_{SS} or V_{DD}). Unused outputs must be left open.

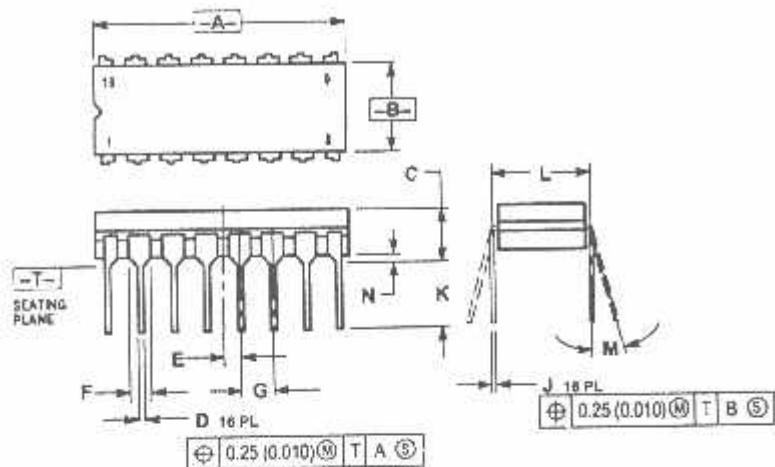
PIN ASSIGNMENT

V _{DD}	1 *	16	NC
OUTA	2	15	OUTF
INA	3	14	INP
OUTB	4	13	NC
INB	5	12	OUTE
OUTC	6	11	INE
INC	7	10	OUTD
V _{SS}	8	9	IND

NC = NO CONNECTION

OUTLINE DIMENSIONS

L SUFFIX
CERAMIC DIP PACKAGE
CASE 620-10
ISSUE V

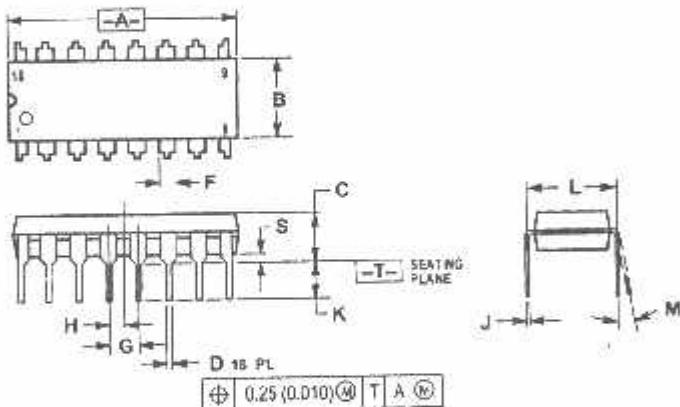


NOTES:

1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: INCH.
3. DIMENSION L TO CENTER OF LEAD WHEN FORMED PARALLEL.
4. DIMENSION F MAY NARROW TO 0.76 (0.030) WHERE THE LEAD ENTERS THE CERAMIC BODY.

DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	0.730	0.765	19.05	19.93
B	0.240	0.295	6.10	7.48
C	—	0.200	—	5.08
D	0.015	0.020	0.39	0.50
E	0.063 BSC	—	1.27 BSC	—
F	0.065	0.065	1.40	1.65
G	0.100 BSC	—	2.54 BSC	—
H	0.058	0.015	0.21	0.38
K	0.125	0.170	3.18	4.31
L	0.300 BSC	—	7.62 BSC	—
M	0°	15°	0°	15°
N	0.020	0.040	0.51	1.01

P SUFFIX
PLASTIC DIP PACKAGE
CASE 648-08
ISSUE R



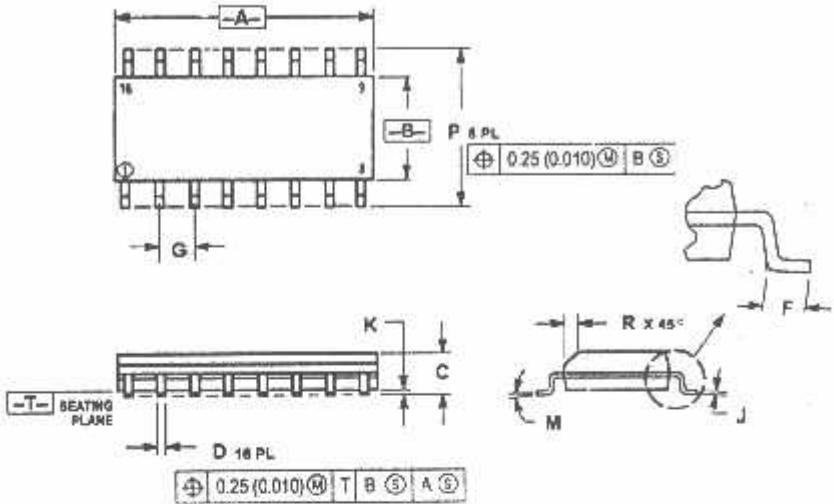
NOTES:

1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: INCH.
3. DIMENSION L TO CENTER OF LEADS WHEN FORMED PARALLEL.
4. DIMENSION B DOES NOT INCLUDE WOLD FLASH.
5. ROUNDED CORNERS OPTIONAL.

DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	0.740	0.770	18.82	19.55
B	0.250	0.270	6.35	6.85
C	0.145	0.175	3.69	4.44
D	0.015	0.021	0.38	0.53
E	0.040	0.70	1.02	1.77
G	0.100 BSC	—	2.54 BSC	—
H	0.016 BSC	—	0.27 BSC	—
J	0.028	0.035	0.21	0.38
K	0.110	0.130	2.81	3.30
L	0.285	0.305	7.20	7.74
M	0°	10°	0°	12°
N	0.020	0.040	0.51	1.01

OUTLINE DIMENSIONS

D SUFFIX
PLASTIC SOIC PACKAGE
CASE 751B-05
ISSUE J



NOTES:

1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: WELDMENT.
3. DIMENSIONS A AND B DO NOT INCLUDE MOLD PROTRUSION.
4. MAXIMUM MOLD PROTRUSION 0.15 (0.006) PER SIDE.
5. DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.127 (0.005) TOTAL IN EXCESS OF THE D DIMENSION AT MAXIMUM MATERIAL CONDITION.

DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	9.80	10.00	0.386	0.393
B	3.92	4.00	0.154	0.157
C	1.35	1.75	0.054	0.066
D	0.35	0.45	0.014	0.019
E	0.40	1.25	0.016	0.049
G	1.27 (55)	—	0.050 (55)	—
J	0.19	0.25	0.008	0.009
K	0.10	0.25	0.004	0.009
M	1.50	7.00	0.060	0.276
P	5.10	5.20	0.229	0.244
R	0.25	1.50	0.010	0.059

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ASIA/PACIFIC: Motorola Semiconductors H.K. Ltd.; 8B Tai Ping Industrial Park, 51 Ting Kok Road, Tai Po, N.T., Hong Kong. 852-26629298



MC14049UB/D



LM741

Operational Amplifier

General Description

The LM741 series are general purpose operational amplifiers which feature improved performance over industry standards like the LM709. They are direct, plug-in replacements for the 709C, LM201, MC1439 and 748 in most applications.

The amplifiers offer many features which make their application nearly foolproof: overload protection on the input and

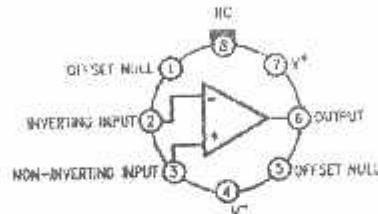
output, no latch-up when the common mode range is exceeded, as well as freedom from oscillations.

The LM741C is identical to the LM741/LM741A except that the LM741C has their performance guaranteed over a 0°C to +70°C temperature range, instead of -55°C to +125°C.

Features

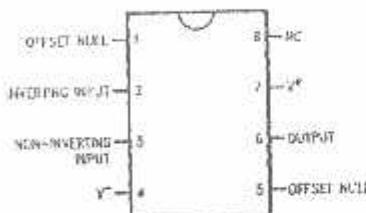
Connection Diagrams

Metal Can Package



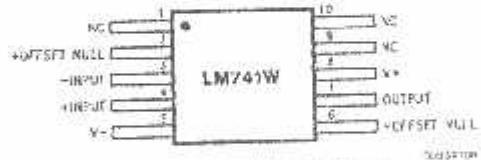
Note 1: LM741H is available per JM38510V10101
 Order Number LM741H, LM741H/883 (Note 1),
 LM741AH/883 or LM741CH
 See NS Package Number H08C

Dual-In-Line or S.O. Package



Order Number LM741J, LM741J/883, LM741CN
 See NS Package Number J08A, M08A or NC8E

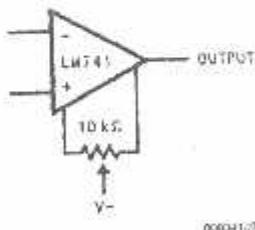
Ceramic Flatpak



Order Number LM741W/883
 See NS Package Number W10A

Typical Application

Offset Nulling Circuit



Absolute Maximum Ratings (Note 2)

If Military/Aerospace specified devices are required,
please contact the National Semiconductor Sales Office/
Distributors for availability and specifications.

(Note 7)

	LM741A	LM741	LM741C
Supply Voltage	±22V	±22V	±18V
Power Dissipation (Note 3)	500 mW	500 mW	500 mW
Differential Input Voltage	±30V	±30V	±30V
Input Voltage (Note 4)	±15V	±15V	±15V
Output Short Circuit Duration	Continuous	Continuous	Continuous
Operating Temperature Range	-55°C to +125°C	-55°C to +125°C	0°C to +70°C
Storage Temperature Range	-65°C to +150°C	-65°C to +150°C	-65°C to +150°C
Junction Temperature	150°C	150°C	100°C
Soldering Information			
N-Package (10 seconds)	260°C	260°C	260°C
J- or H-Package (10 seconds)	300°C	300°C	300°C
M-Package			
Vapor Phase (60 seconds)	215°C	215°C	215°C
Infrared (15 seconds)	215°C	215°C	215°C
See AN-450 "Surface Mounting Methods and Their Effect on Product Reliability" for other methods of soldering			
surface mount devices:			
ESD Tolerance (Note 8)	400V	400V	400V

Electrical Characteristics (Note 5)

Parameter	Conditions	LM741A			LM741			LM741C			Units
		Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	
Input Offset Voltage	$T_A = 25^\circ\text{C}$ $R_S \leq 10 \text{ k}\Omega$ $R_B \leq 50\Omega$					1.0	5.0		2.0	6.0	mV mV
	$T_{A\text{MIN}} \leq T_A \leq T_{A\text{MAX}}$ $R_S \leq 50\Omega$ $R_B \leq 10 \text{ k}\Omega$		0.8	3.0		4.0		6.0		7.5	mV mV
Average Input Offset Voltage Drift				15							µV/°C
											mV
Input Offset Voltage Adjustment Range	$T_A = 25^\circ\text{C}, V_S = \pm 20\text{V}$	±10			±15			±15			
Input Offset Current	$T_A = 25^\circ\text{C}$		3.0	30		20	200		20	200	nA
	$T_{A\text{MIN}} \leq T_A \leq T_{A\text{MAX}}$			70		85	500			300	nA
Average Input Offset Current Drift				0.5							nA/°C
Input Bias Current	$T_A = 25^\circ\text{C}$		30	80		80	500		80	500	nA
	$T_{A\text{MIN}} \leq T_A \leq T_{A\text{MAX}}$			0.210			1.5			0.8	µA
Input Resistance	$T_A = 25^\circ\text{C}, V_S = \pm 20\text{V}$	1.0	6.0		0.3	2.0		0.3	2.0		MΩ
	$T_{A\text{MIN}} \leq T_A \leq T_{A\text{MAX}}$ $V_S = \pm 20\text{V}$		0.5								MΩ
Input Voltage Range	$T_A = 25^\circ\text{C}$							±12	±13		V
	$T_{A\text{MIN}} \leq T_A \leq T_{A\text{MAX}}$				±12	±13					V

Electrical Characteristics (Note 5) (Continued)

Parameter	Conditions	LM741A			LM741			LM741C			Units
		Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	
Input Signal Voltage Gain	$T_A = 25^\circ\text{C}$, $R_L \geq 2 \text{ k}\Omega$ $V_S = \pm 20\text{V}$, $V_O = \pm 15\text{V}$ $V_S = \pm 15\text{V}$, $V_O = \pm 10\text{V}$	50			50	200		20	200		V/mV V/mV
	$T_{A\text{MIN}} \leq T_A \leq T_{A\text{MAX}}$ $R_L \geq 2 \text{ k}\Omega$ $V_S = \pm 20\text{V}$, $V_O = \pm 15\text{V}$ $V_S = \pm 15\text{V}$, $V_O = \pm 10\text{V}$ $V_S = \pm 5\text{V}$, $V_O = \pm 2\text{V}$		32			25			15		
Output Voltage Swing	$V_S = \pm 20\text{V}$ $R_L \geq 10 \text{ k}\Omega$ $R_L \geq 2 \text{ k}\Omega$	± 16			± 12	± 14		± 12	± 14		V V
	$V_S = \pm 15\text{V}$ $R_L \geq 10 \text{ k}\Omega$ $R_L \geq 2 \text{ k}\Omega$		± 15			± 10	± 13		± 10	± 13	
Output Short Circuit Current	$T_A = 25^\circ\text{C}$	10	25	35	25			25			mA mA
	$T_{A\text{MIN}} \leq T_A \leq T_{A\text{MAX}}$	10		43							
Common-Mode Rejection Ratio	$T_{A\text{MIN}} \leq T_A \leq T_{A\text{MAX}}$ $R_S \leq 10 \text{ k}\Omega$, $V_{CM} = \pm 12\text{V}$ $R_S \leq 50\Omega$, $V_{CM} = \pm 12\text{V}$	80	95		70	90		70	90		dB dB
Supply Voltage Rejection Ratio	$T_{A\text{MIN}} \leq T_A \leq T_{A\text{MAX}}$ $V_S = \pm 20\text{V}$ to $V_S = \pm 5\text{V}$ $R_E \leq 50\Omega$ $R_E \leq 10 \text{ k}\Omega$	86	96		77	96		77	96		dB dB
Transient Response	$T_A = 25^\circ\text{C}$, Unity Gain		0.25	0.8		0.3			0.3		μs %
	Rise Time		6.0	20		5			5		
Overshoot											
Bandwidth (Note 6)	$T_A = 25^\circ\text{C}$	0.437	1.5								MHz
Slew Rate	$T_A = 25^\circ\text{C}$, Unity Gain	0.3	0.7			0.5		0.5			$\text{V}/\mu\text{s}$
Supply Current	$T_A = 25^\circ\text{C}$					1.7	2.8		1.7	2.8	mA
Power Consumption	$T_A = 25^\circ\text{C}$ $V_S = \pm 20\text{V}$ $V_S = \pm 15\text{V}$		80	150		50	85		50	85	mW mW
LM741A	$V_S = \pm 20\text{V}$ $T_A = T_{A\text{MIN}}$ $T_A = T_{A\text{MAX}}$			165							mW nW
				135							
LM741	$V_S = \pm 15\text{V}$ $T_A = T_{A\text{MIN}}$ $T_A = T_{A\text{MAX}}$					60	100				mW nW
						45	75				

Note 2: "Absolute Maximum Ratings" indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional, but do not guarantee specific performance limits.

Electrical Characteristics (Note 5) (Continued)

Note 3: For operation at elevated temperatures, these devices must be derated based on thermal resistance, and T_j max. (listed under "Absolute Maximum Ratings"). $T_j = T_A + (D_{JA} P_0)$.

Thermal Resistance	Cerdip (J)	DIP (N)	HO8 (H)	SO-8 (M)
θ_{JA} (Junction to Ambient)	100°C/W	100°C/W	170°C/W	195°C/W
θ_{JC} (Junction to Case)	N/A	N/A	25°C/W	N/A

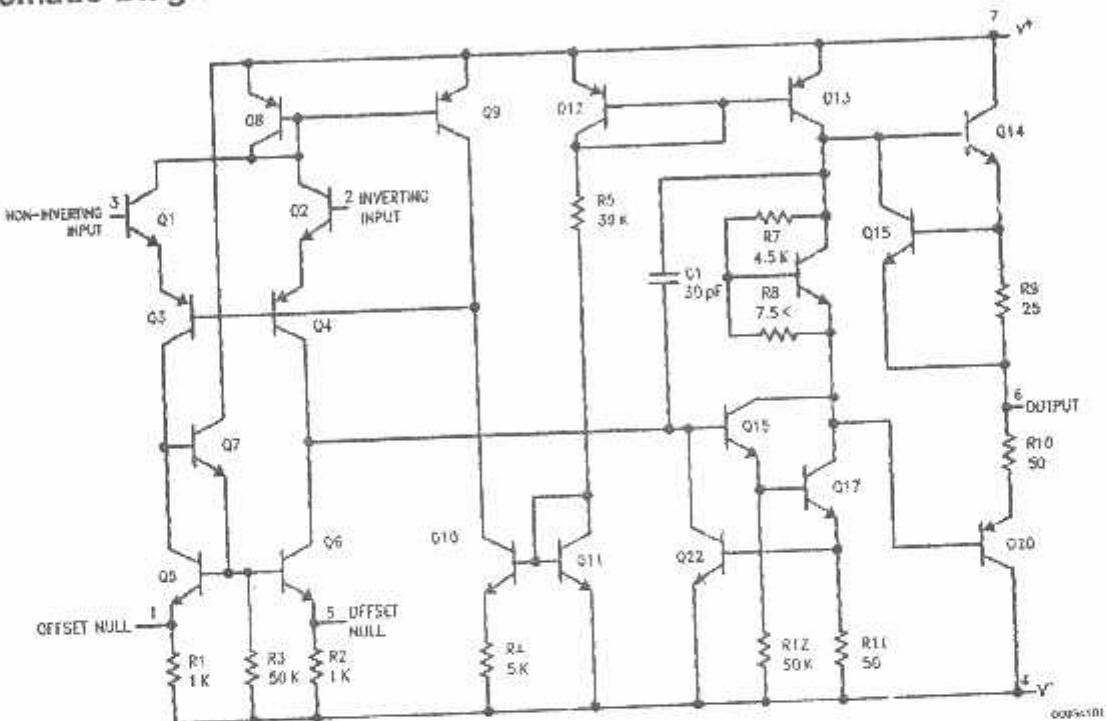
Note 4: For supply voltages less than $\pm 15V$, the absolute maximum input voltage is equal to the supply voltage.

Note 5: Unless otherwise specified, these specifications apply for $V_S = \pm 15V$, $-55^\circ C \leq T_A \leq +125^\circ C$ (LM741/LM741A). For the LM741C/LM741E, these specifications are limited to $0^\circ C \leq T_A \leq +70^\circ C$.

Note 6: Calculated value from BW (MHz) = $0.3/\text{Rise Time}(\mu\text{s})$.

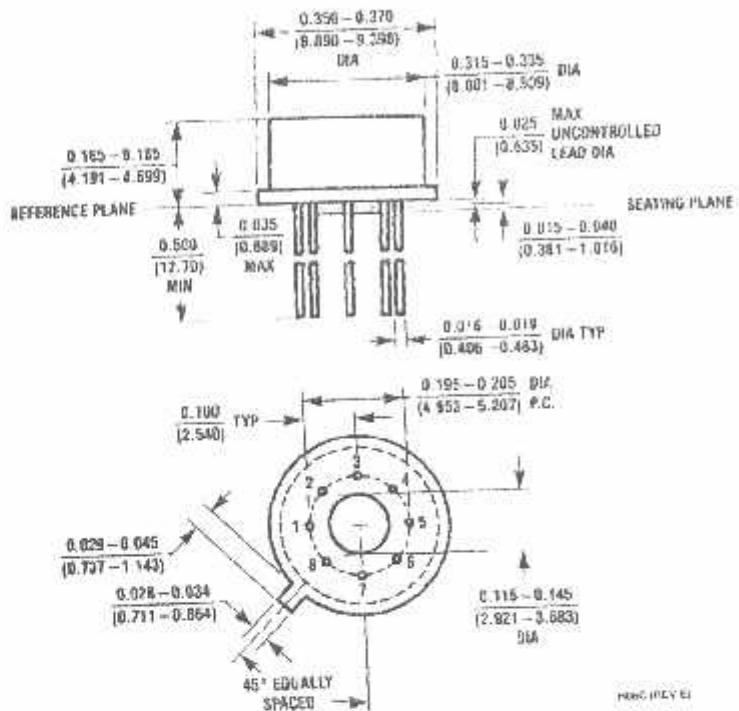
Note 7: For military specifications see MIL-S-741X for LM741 and MIL-S-741AX for LM741A.

Note 8: Human body model, 1.5 kΩ in series with 100 pF.

Schematic Diagram

Physical Dimensions inches (millimeters)

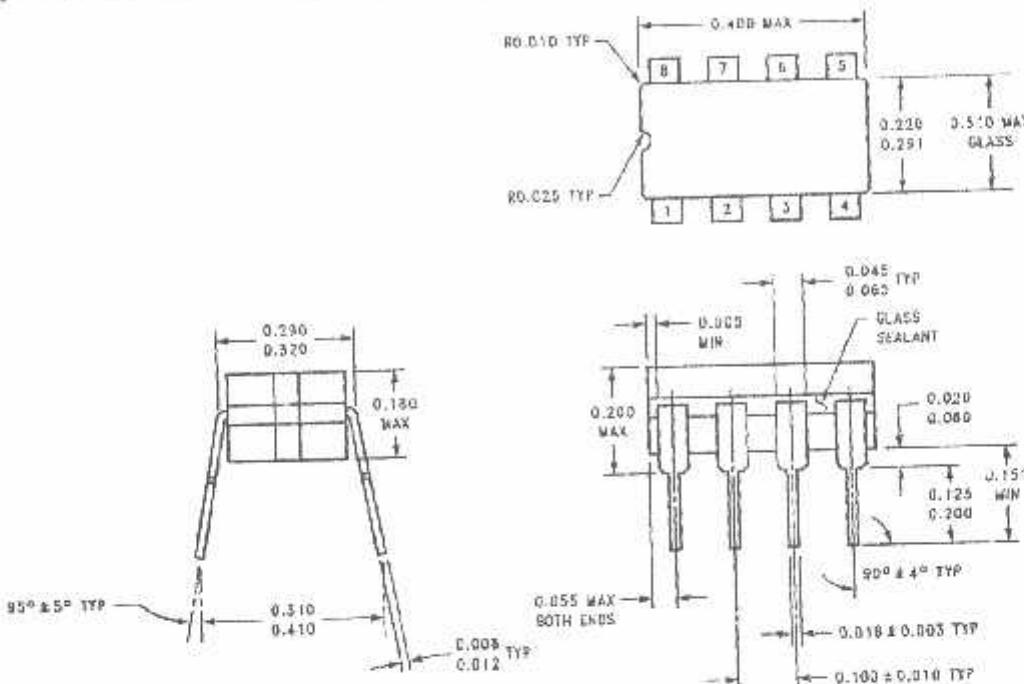
Unless otherwise noted



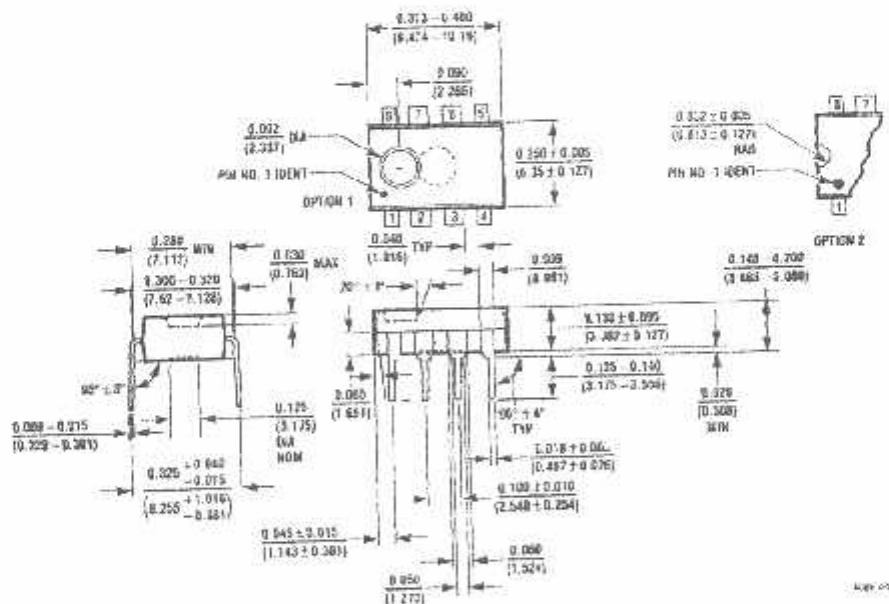
Metal Can Package (H)

Order Number LM741H, LM741H/883, LM741AH/883, LM741AH-MIL or LM741CH
NS Package Number H08C

Physical Dimensions Inches (millimeters) unless otherwise noted (Continued)

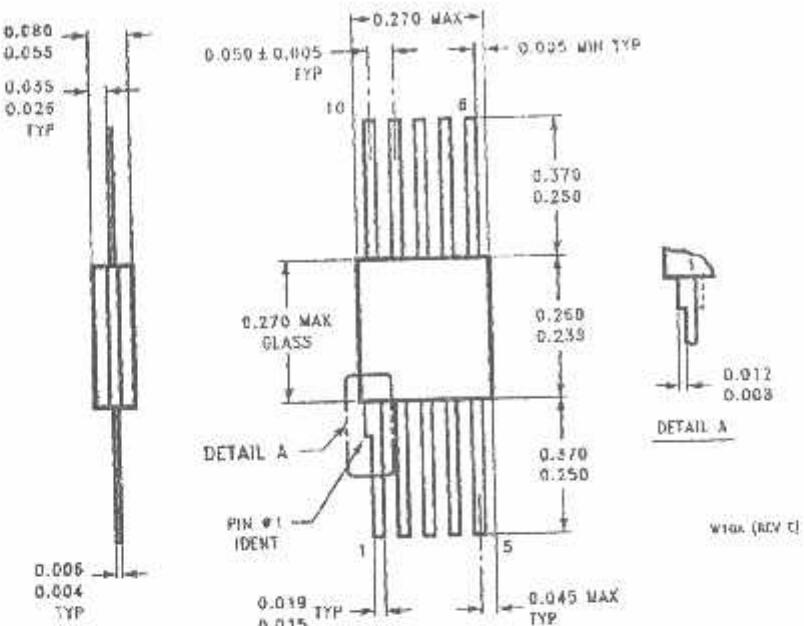


Ceramic Dual-In-Line Package (J)
Order Number LM741J/683
NS Package Number J08A



Dual-In-Line Package (N)
Order Number LM741CN
NS Package Number N08E

Physical Dimensions inches (millimeters) unless otherwise noted (Continued)



10-Lead Ceramic Flatpak (W)
Order Number LM741W/883, LM741WG-MPR or LM741W/G/883
NS Package Number W10A

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FORMULIR BIMBINGAN SKRIPSI

na : ISNAINI ROZAFI
A : 01.17.044
sa Bimbingan : 10 Februari 2007 s/d 10 Agustus 2007
ul Skripsi : Perencanaan dan Pembuatan Master Kontrol Lab. Bahasa 8 Channel
Berbasis Mikrokontroller AT89S8252

Tanggal	Uraian	Paraf Pembimbing
10-2-07	Isbab I & II	✓
10-2-07	Revisi Bab III	✓
18-02-07	Revisi Bab IV	✓
21-02-07	Bab III & IV	✓
25-02-07	Revisi Bab V	✓
28-02-07	Revisi Bab VI	✓
08-3-07	Ace Seminar Inkl	✓
12-3-2007	An. wmpkt	✓

Malang, 12 - 3 - 2007
Dosen Pembimbing

Joseph Dedy Irawan, ST., MT.
NIP. 132315178

Form S-4

e 9. T2MOD – Timer 2 Mode Control Register

Reset Value = XXXX XX00B

I/O Address = 0C9H

Bit Addressable

	7	6	5	4	3	2	1	T2OE	DCEN
1	-	-	-	-	-	-	-	-	0

Symbol	Function
	Not implemented, reserved for future use.
OE	Timer 2 Output Enable bit.
EN	When set, this bit allows Timer 2 to be configured as an up/down counter.

Figure 3. Timer 2 Auto Reload Mode (DCEN = 1)

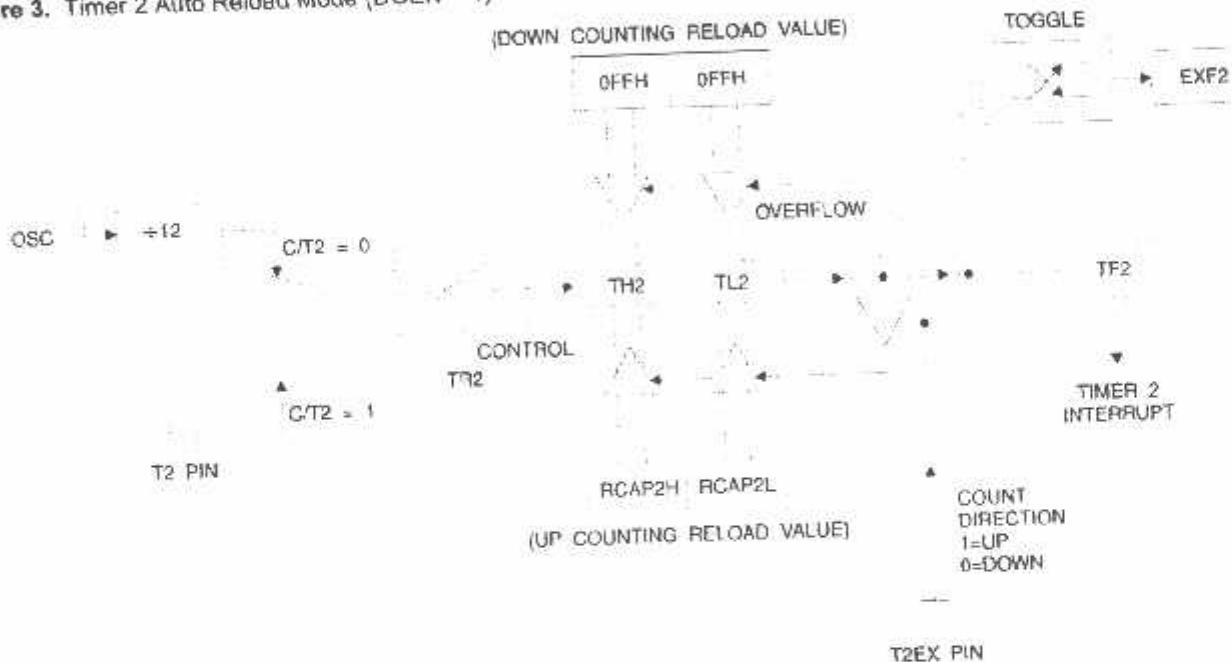
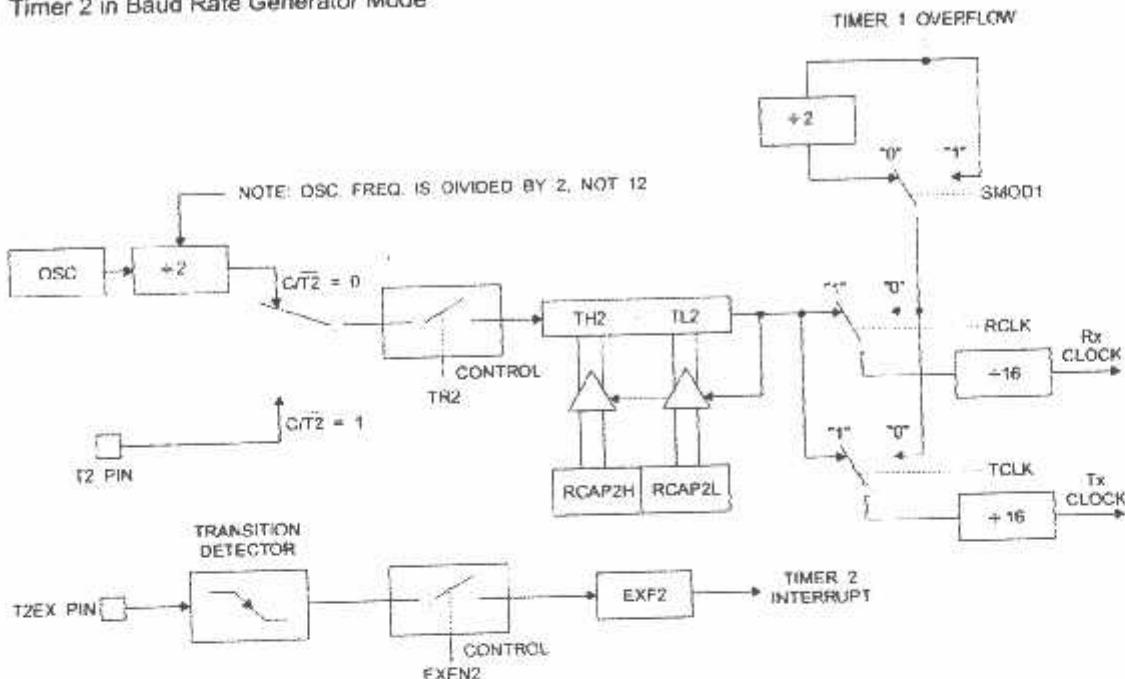


Figure 4. Timer 2 in Baud Rate Generator Mode



Baud Rate Generator

Timer 2 is selected as the baud rate generator by setting TCLK and/or RCLK in T2CON (Table 2). Note that the baud rates for transmit and receive can be different if Timer 2 is used for the receiver or transmitter and Timer 1 is used for the other function. Setting RCLK and/or TCLK puts Timer 2 into its baud rate generator mode, as shown in Figure 4.

The baud rate generator mode is similar to the auto-reload mode, in that a rollover in TH2 causes the Timer 2 registers to be reloaded with the 16-bit value in registers RCAP2H and RCAP2L, which are preset by software.

The baud rates in Modes 1 and 3 are determined by Timer 2's overflow rate according to the following equation.

$$\text{Modes 1 and 3 Baud Rates} = \frac{\text{Timer 2 Overflow Rate}}{16}$$

The Timer can be configured for either timer or counter operation. In most applications, it is configured for timer operation ($CP/T2 = 0$). The timer operation is different for Timer 2 when it is used as a baud rate generator. Normally, as a timer, it increments every machine cycle (at 1/12 the oscillator frequency). As a baud rate generator, however, it increments every state time (at 1/2 the oscillator frequency). The baud rate formula is given below.

$$\frac{\text{Modes 1 and 3}}{\text{Baud Rate}} = \frac{\text{Oscillator Frequency}}{32 \times [65536 - (\text{RCAP2H}, \text{RCAP2L})]}$$

where $(\text{RCAP2H}, \text{RCAP2L})$ is the content of RCAP2H and RCAP2L taken as a 16-bit unsigned integer.





Timer 2 as a baud rate generator is shown in Figure 4. This figure is valid only if RCLK = 1 or TCLK = 1 in T2CON. Note that a rollover in TH2 does not set TF2 and will not generate an interrupt. Note too, that if EXEN2 is set, a 1-to-0 transition in T2EX will set EXF2 but will not cause a reload from (RCAP2H, RCAP2L) to (TH2, TL2). Thus when Timer 2 is in use as a baud rate generator, T2EX can be used as an extra external interrupt.

Note that when Timer 2 is running (TR2 = 1) as a timer in the baud rate generator mode, TH2 or TL2 should not be read from or written to. Under these conditions, the Timer is incremented every state time, and the results of a read or write may not be accurate. The RCAP2 registers may be read but should not be written to, because a write might overlap a reload and cause write and/or reload errors. The timer should be turned off (clear TR2) before accessing the Timer 2 or RCAP2 registers.

Programmable Clock Out

A 50% duty cycle clock can be programmed to come out on P1.0, as shown in Figure 5. This pin, besides being a regular I/O pin, has two alternate functions. It can be programmed to input the external clock for Timer/Counter 2 or to output a 50% duty cycle clock ranging from 61 Hz to 4 MHz (for a 16-MHz operating frequency).

To configure the Timer/Counter 2 as a clock generator, bit C/T2 (T2CON.1) must be cleared and bit T2OE (T2MOD.1) must be set. Bit TR2 (T2CON.2) starts and stops the timer.

The clock-out frequency depends on the oscillator frequency and the reload value of Timer 2 capture registers (RCAP2H, RCAP2L), as shown in the following equation:

$$\text{Clock Out Frequency} = \frac{\text{Oscillator Frequency}}{4 \times [65536 - (\text{RCAP2H}, \text{RCAP2L})]}$$

In the clock-out mode, Timer 2 rollovers will not generate an interrupt. This behavior is similar to when Timer 2 is used as a baud-rate generator. It is possible to use Timer 2 as a baud-rate generator and a clock generator simultaneously. Note, however, that the baud-rate and clock-out frequencies cannot be determined independently from one another since they both use RCAP2H and RCAP2L.

Figure 5. Timer 2 in Clock-out Mode

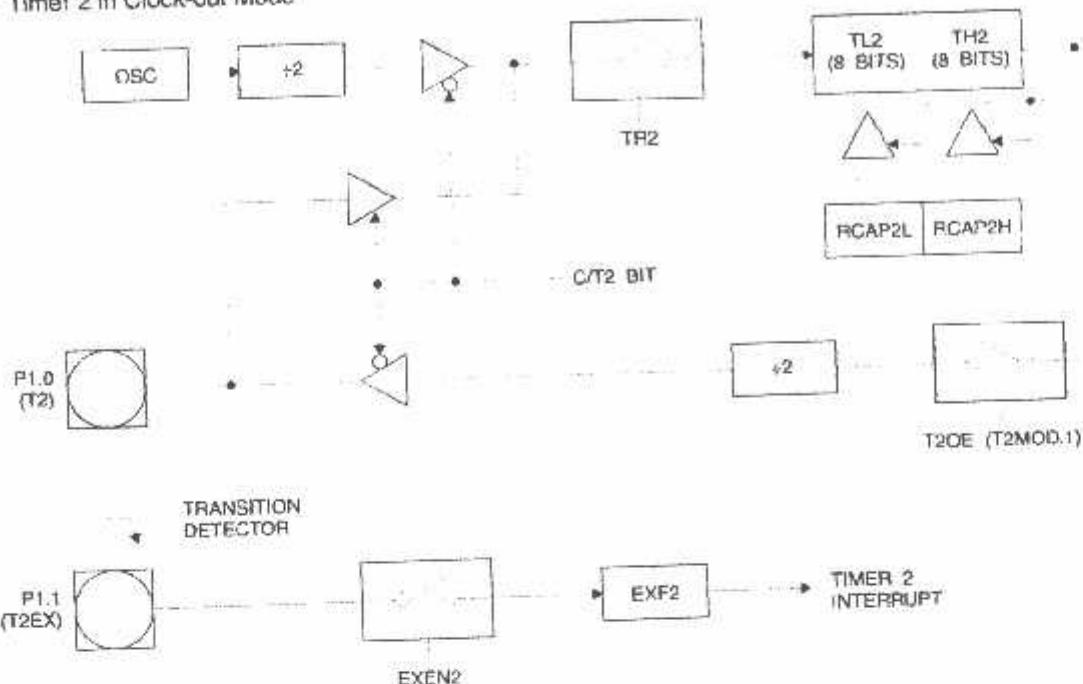
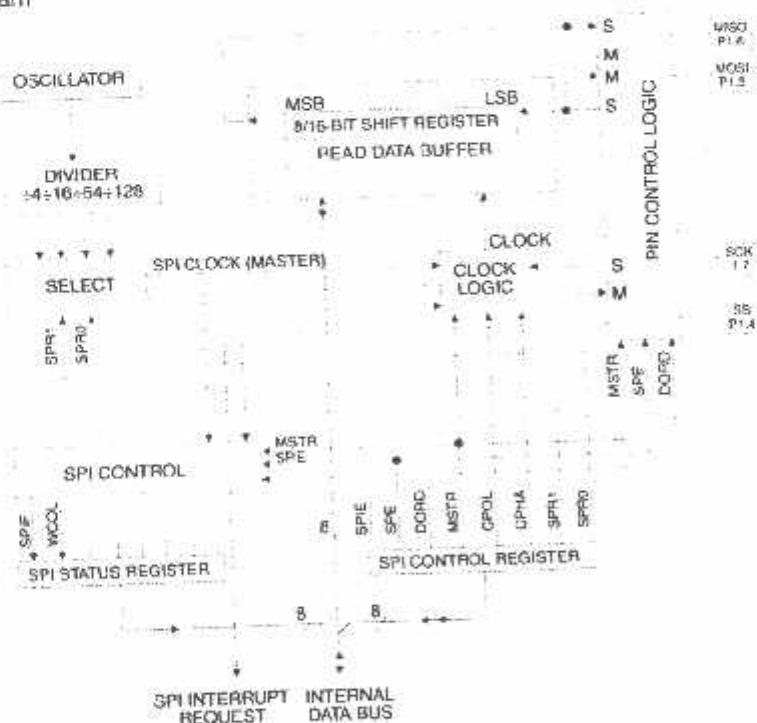


Figure 6. SPI Block Diagram



RT

The UART in the AT89S8252 operates the same way as the UART in the AT89C51 and AT89C52. For further information on the UART operation, refer to the Atmel web site (<http://www.atmel.com>). From the home page, select "Products", then "Microcontrollers", then "8051-Architecture". Click on "Documentation", then on "Other Documents". Open the document "AT89 Series Hardware Description".

Serial Peripheral Interface

The serial peripheral interface (SPI) allows high-speed synchronous data transfer between the AT89S8252 and peripheral devices or between several AT89S8252 devices. The AT89S8252 SPI features include the following:

- Full-Duplex, 3-Wire Synchronous Data Transfer
- Master or Slave Operation
- 1.5 MHz Bit Frequency (max.)
- LSB First or MSB First Data Transfer
- Four Programmable Bit Rates
- End of Transmission Interrupt Flag
- Write Collision Flag Protection
- Wakeup from Idle Mode (Slave Mode Only)

The interconnection between master and slave CPUs with SPI is shown in the following figure. The SCK pin is the clock output in the master mode but is the clock input in the slave mode. Writing to the SPI data register of the master CPU starts the SPI clock generator, and the data written shifts out of the MOSI pin and into the MOSI pin of the slave CPU. After shifting one byte, the SPI clock generator stops, setting the end of transmission flag (SPIF). If both the SPI interrupt enable bit (SPIE) and the serial port interrupt enable bit (ES) are set, an interrupt is requested.

The Slave Select input, SS/P1.4, is set low to select an individual SPI device as a slave. When SS/P1.4 is set high, the SPI port is deactivated and the MOSI/P1.5 pin can be used as an input.

There are four combinations of SCK phase and polarity with respect to serial data, which are determined by control bits CPHA and CPOL. The SPI data transfer formats are shown in Figure 8 and Figure 9.

Figure 7. SPI Master-slave Interconnection

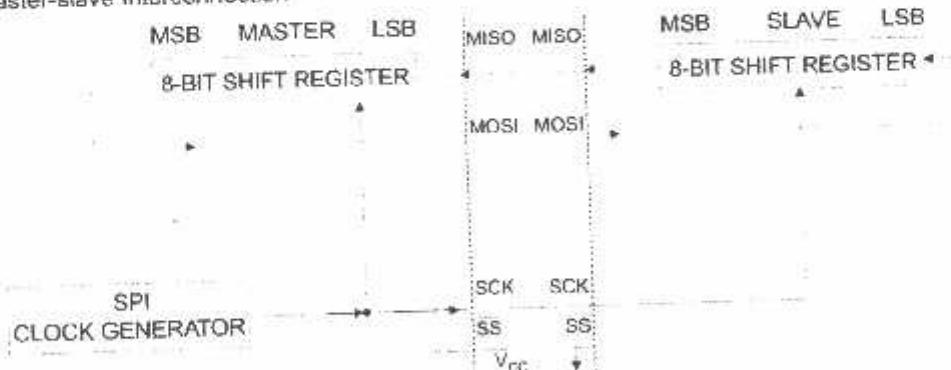
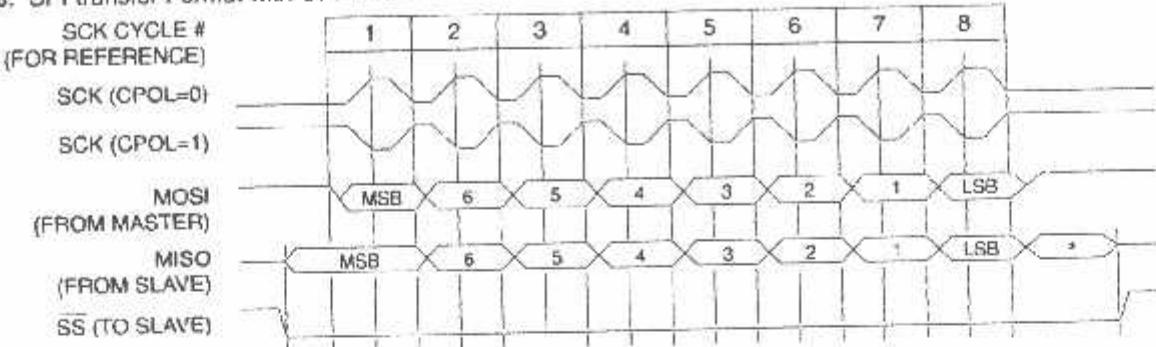
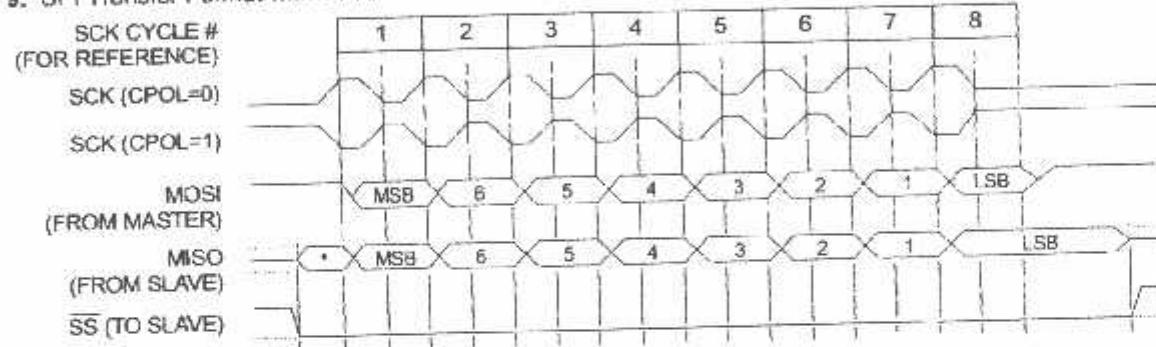


Figure 8. SPI transfer Format with CPHA = 0



Note: *Not defined but normally MSB of character just received

Figure 9. SPI Transfer Format with CPHA = 1



Note: *Not defined but normally LSB of previously transmitted character.

Interrupts

The AT89S8252 has a total of six interrupt vectors, two external interrupts (INT0 and INT1), three timer interrupts (Timers 0, 1, and 2), and the serial port interrupt. These interrupts are all shown in Figure 10.

Each of these interrupt sources can be individually enabled or disabled by setting or clearing a bit in Special Function Register IE. IE also contains a global disable bit, EA, which disables all interrupts at once.

Note that Table 10 shows that bit position IE.6 is unimplemented. In the AT89C51, bit position IE.5 is also unimplemented. User software should not write 1s to these bit positions, since they may be used in future AT89 products.

Timer 2 interrupt is generated by the logical OR of bits TF2 and EXF2 in register T2CON. Neither of these flags is cleared by hardware when the service routine is vectored to. In fact, the service routine may have to determine whether it was TF2 or EXF2 that generated the interrupt, and that bit will have to be cleared in software.

The Timer 0 and Timer 1 flags, TF0 and TF1, are set at S5P2 of the cycle in which the timers overflow. The values are then polled by the circuitry in the next cycle. However, the Timer 2 flag, TF2, is set at S2P2 and is polled in the same cycle in which the timer overflows.

e 10. Interrupt Enable (IE) Register

(SB)(LSB)

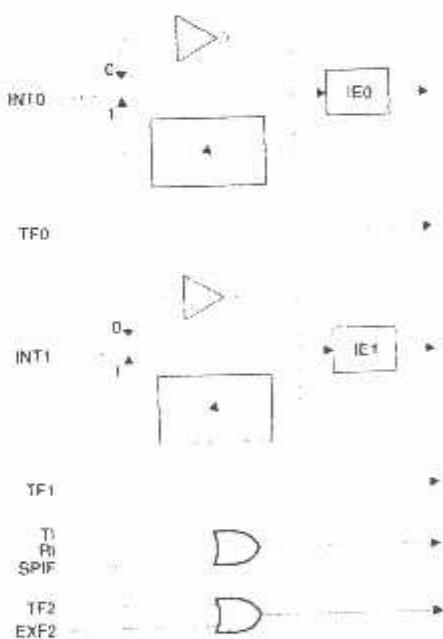
EA	-	ET2	ES	ET1	EX1	ET0	EX0
----	---	-----	----	-----	-----	-----	-----

Enable Bit = 1 enables the interrupt.

Enable Bit = 0 disables the interrupt.

Symbol	Position	Function
EA	IE.7	Disables all interrupts. If EA = 0, no interrupt is acknowledged. If EA = 1, each interrupt source is individually enabled or disabled by setting or clearing its enable bit.
-	IE.6	Reserved.
ET2	IE.5	Timer 2 interrupt enable bit.
ES	IE.4	SPI and UART interrupt enable bit.
ET1	IE.3	Timer 1 interrupt enable bit.
EX1	IE.2	External interrupt 1 enable bit.
ET0	IE.1	Timer 0 interrupt enable bit.
EX0	IE.0	External interrupt 0 enable bit.

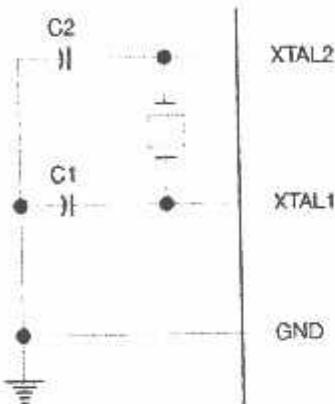
User software should never write 1s to unimplemented bits, because they may be used in future AT89 products.

Figure 10. Interrupt Sources

Oscillator Characteristics

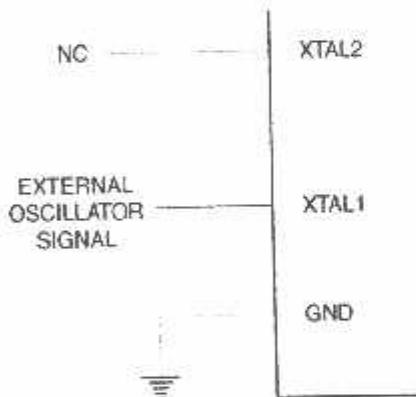
XTAL1 and XTAL2 are the input and output, respectively, of an inverting amplifier that can be configured for use as an on-chip oscillator, as shown in Figure 11. Either a quartz crystal or ceramic resonator may be used. To drive the device from an external clock source, XTAL2 should be left unconnected while XTAL1 is driven, as shown in Figure 12. There are no requirements on the duty cycle of the external clock signal, but since the input to the internal clocking circuitry is through a divide-by-two flip-flop, minimum and maximum voltage high and low time specifications must be observed.

Figure 11. Oscillator Connections



Note: C₁, C₂ = 30 pF ± 10 pF for Crystals
= 40 pF ± 10 pF for Ceramic Resonators

Figure 12. External Clock Drive Configuration





Idle Mode

In Idle mode, the CPU puts itself to sleep while all the on-chip peripherals remain active. The mode is invoked by software. The content of the on-chip RAM and all the special function registers remain unchanged during this mode. The idle mode can be terminated by any enabled interrupt or by a hardware reset.

Note that when idle mode is terminated by a hardware reset, the device normally resumes program execution from where it left off, up to two machine cycles before the internal reset algorithm takes control. On-chip hardware inhibits access to internal RAM in this event, but access to the port pins is not inhibited. To eliminate the possibility of an unexpected write to a port pin when idle mode is terminated by a reset, the instruction following the one that invokes idle mode should not write to a port pin or to external memory.

Status of External Pins During Idle and Power-down Modes

Mode	Program Memory	ALE	PSEN	PORT0	PORT1	PORT2	PORT3
Idle	Internal	1	1	Data	Data	Data	Data
Idle	External	1	1	Float	Data	Address	Data
Power-down	Internal	0	0	Data	Data	Data	Data
Power-down	External	0	0	Float	Data	Data	Data

Power-down Mode

In the power-down mode, the oscillator is stopped and the instruction that invokes power-down is the last instruction executed. The on-chip RAM and Special Function Registers retain their values until the power-down mode is terminated. Exit from power-down can be initiated either by a hardware reset or by an enabled external interrupt. Reset redefines the SFRs but does not change the on-chip RAM. The reset should not be activated before V_{CC} is restored to its normal operating level and must be held active long enough to allow the oscillator to restart and stabilize.

To exit power-down via an interrupt, the external interrupt must be enabled as level sensitive before entering power-down. The interrupt service routine starts at 16 ms (nominal) after the enabled interrupt pin is activated.

Program Memory Lock Bits

The AT89S8252 has three lock bits that can be left unprogrammed (U) or can be programmed (P) to obtain the additional features listed in the following table.

When lock bit 1 is programmed, the logic level at the EA pin is sampled and latched during reset. If the device is powered up without a reset, the latch initializes to a random value and holds that value until reset is activated. The latched value of EA must agree with the current logic level at that pin in order for the device to function properly.

Once programmed, the lock bits can only be unprogrammed with the Chip Erase operations in either the parallel or serial modes.

Lock Bit Protection Modes⁽¹⁾⁽²⁾

Program Lock Bits				Protection Type
	LB1	LB2	LB3	
1	U	U	U	No internal memory lock feature.
2	P	U	U	MOVC instructions executed from external program memory are disabled from fetching code bytes from internal memory. EA is sampled and latched on reset and further programming of the Flash memory (parallel or serial mode) is disabled.
3	P	P	U	Same as Mode 2, but parallel or serial verify are also disabled.
4	P	P	P	Same as Mode 3, but external execution is also disabled.

Notes:
1. U = Unprogrammed
2. P = Programmed

Programming the Flash and EEPROM

Atmel's AT89S8252 Flash Microcontroller offers 8K bytes of in-system reprogrammable Flash Code memory and 2K bytes of EEPROM Data memory.

The AT89S8252 is normally shipped with the on-chip Flash Code and EEPROM Data memory arrays in the erased state (i.e. contents = FFH) and ready to be programmed. This device supports a High-voltage (12-V V_{PP}) Parallel programming mode and a Low-voltage (5-V V_{CC}) Serial programming mode. The serial programming mode provides a convenient way to reprogram the AT89S8252 inside the user's system. The parallel programming mode is compatible with conventional third party Flash or EPROM programmers.

The Code and Data memory arrays are mapped via separate address spaces in the serial programming mode. In the parallel programming mode, the two arrays occupy one contiguous address space: 0000H to 1FFFH for the Code array and 2000H to 27FFH for the Data array.

The Code and Data memory arrays on the AT89S8252 are programmed byte-by-byte in either programming mode. An auto-erase cycle is provided with the self-timed programming operation in the serial programming mode. There is no need to perform the Chip Erase operation to reprogram any memory location in the serial programming mode unless any of the lock bits have been programmed.

In the parallel programming mode, there is no auto-erase cycle. To reprogram any non-blank byte, the user needs to use the Chip Erase operation first to erase both arrays.

Parallel Programming Algorithm: To program and verify the AT89S8252 in the parallel programming mode, the following sequence is recommended:

1. Power-up sequence:
Apply power between V_{CC} and GND pins.
Set RST pin to "H".
Apply a 3 MHz to 24 MHz clock to XTAL1 pin and wait for at least 10 milliseconds.
2. Set PSEN pin to "L"
ALE pin to "H"
EA pin to "H" and all other pins to "H".
3. Apply the appropriate combination of "H" or "L" logic levels to pins P2.6, P2.7, P3.6, P3.7 to select one of the programming operations shown in the Flash Programming Modes table.
4. Apply the desired byte address to pins P1.0 to P1.7 and P2.0 to P2.5.
Apply data to pins P0.0 to P0.7 for Write Code operation.
5. Raise EA/V_{PP} to 12V to enable Flash programming, erase or verification.
6. Pulse ALE/PROG once to program a byte in the Code memory array, the Data memory array or the lock bits. The byte-write cycle is self-timed and typically takes 1.5 ms.
7. To verify the byte just programmed, bring pin P2.7 to "L" and read the programmed data at pins P0.0 to P0.7.
8. Repeat steps 3 through 7 changing the address and data for the entire 2K or 8K bytes array or until the end of the object file is reached.
9. Power-off sequence:
Set XTAL1 to "L".
Set RST and EA pins to "L".
Turn V_{CC} power off.





In the parallel programming mode, there is no auto-erase cycle and to reprogram any non-blank byte, the user needs to use the Chip Erase operation first to erase both arrays.

Data Polling: The AT89S8252 features DATA Polling to indicate the end of a byte write cycle. During a byte write cycle in the parallel or serial programming mode, an attempted read of the last byte written will result in the complement of the written datum on P0.7 (parallel mode), and on the MSB of the serial output byte on MISO (serial mode). Once the write cycle has been completed, true data are valid on all outputs, and the next cycle may begin. DATA Polling may begin any time after a write cycle has been initiated.

Ready/Busy: The progress of byte programming in the parallel programming mode can also be monitored by the RDY/BSY output signal. Pin P3.4 is pulled Low after ALE goes High during programming to indicate BUSY. P3.4 is pulled High again when programming is done to indicate READY.

Program Verify: If lock bits LB1 and LB2 have not been programmed, the programmed Code or Data byte can be read back via the address and data lines for verification. The state of the lock bits can also be verified directly in the parallel programming mode. In the serial programming mode, the state of the lock bits can only be verified indirectly by observing that the lock bit features are enabled.

Chip Erase: Both Flash and EEPROM arrays are erased electrically at the same time. In the parallel programming mode, chip erase is initiated by using the proper combination of control signals and by holding ALE/PROG low for 10 ms. The Code and Data arrays are written with all "1"s in the Chip Erase operation.

In the serial programming mode, a chip erase operation is initiated by issuing the Chip Erase instruction. In this mode, chip erase is self-timed and takes about 16 ms.

During chip erase, a serial read from any address location will return 00H at the data outputs.

Serial Programming Fuse: A programmable fuse is available to disable Serial Programming if the user needs maximum system security. The Serial Programming Fuse can only be programmed or erased in the Parallel Programming Mode.

The AT89S8252 is shipped with the Serial Programming Mode enabled.

Reading the Signature Bytes: The signature bytes are read by the same procedure as a normal verification of locations 030H and 031H, except that P3.6 and P3.7 must be pulled to a logic low. The values returned are as follows:

(030H) = 1EH indicates manufactured by Atmel

(031H) = 72H indicates 89S8252

Programming Interface

Every code byte in the Flash and EEPROM arrays can be written, and the entire array can be erased, by using the appropriate combination of control signals. The write operation cycle is self-timed and once initiated, will automatically time itself to completion.

Most worldwide major programming vendors offer support for the Atmel AT89 microcontroller series. Please contact your local programming vendor for the appropriate software revision.



Serial Programming Instruction

The Instruction Set for Serial Programming follows a 3-byte protocol and is shown in the following table:

Instruction Set

Instruction	Input Format			Operation
	Byte 1	Byte 2	Byte 3	
Programming Enable	1010 1100	0101 0011	xxxx xxxx	Enable serial programming interface after RST goes high.
Chip Erase	1010 1100	xxxx x100	xxxx xxxx	Chip erase both 8K & 2K memory arrays.
Read Code Memory	aaaa a001	low addr	xxxx xxxx	Read data from Code memory array at the selected address. The 5 MSBs of the first byte are the high order address bits. The low order address bits are in the second byte. Data are available at pin MISO during the third byte.
Write Code Memory	aaaa a010	low addr	data in	Write data to Code memory location at selected address. The address bits are the 5 MSBs of the first byte together with the second byte.
Read Data Memory	00aa a101	low addr	xxxx xxxx	Read data from Data memory array at selected address. Data are available at pin MISO during the third byte.
Write Data Memory	00aa a110	low addr	data in	Write data to Data memory location at selected address.
Write Lock Bits	1010 1100	xxxx x111	xxxx xxxx	Write lock bits. Set LB1, LB2 or LB3 = "0" to program lock bits.

- Notes:
1. DATA polling is used to indicate the end of a byte write cycle which typically takes less than 2.5 ms at 5V.
 2. "aaaaa" = high order address.
 3. "X" = don't care.

Flash and EEPROM Parallel Programming Modes

Mode	RST	PSEN	ALE/PROG	\bar{E}/V_{PP}	P2.6	P2.7	P3.6	P3.7	Data I/O P0.7:0	Address P2.5:0 P1.7:0
Serial Prog. Modes	H	h ⁽¹⁾	h ⁽¹⁾	x						
Chip Erase	H	L	 (2)	12V	H	L	L	L	X	X
Write (10K bytes) Memory	H	L		12V	L	H	H	H	DIN	ADDR
Read (10K bytes) Memory	H	L	H	12V	L	L	H	H	DOUT	ADDR
Write Lock Bits:	H	L		12V	H	L	H	L	DIN	X
									P0.7 = 0	X
									P0.6 = 0	X
									P0.5 = 0	X
Read Lock Bits:	H	L	H	12V	H	H	L	L	DOUT	X
									@P0.2	X
									@P0.1	X
									@P0.0	X
Read Atmel Code	H	L	H	12V	L	L	L	L	DOUT	30H
Read Device Code	H	L	H	12V	L	L	L	L	DOUT	31H
Serial Prog. Enable	H	L	 (2)	12V	L	H	L	H	P0.0 = 0	X
Serial Prog. Disable	H	L	 (2)	12V	L	H	L	H	P0.0 = 1	X
Read Serial Prog. Fuse	H	L	H	12V	H	H	L	H	@P0.0	X

- Notes:
1. "h" = weakly pulled "High" internally.
 2. Chip Erase and Serial Programming Fuse require a 10 ms PROG pulse. Chip Erase needs to be performed first before reprogramming any byte with a content other than FFH.
 3. P3.4 is pulled Low during programming to indicate RDY/BSY.
 4. "X" = don't care

Figure 13. Programming the Flash/EEPROM Memory

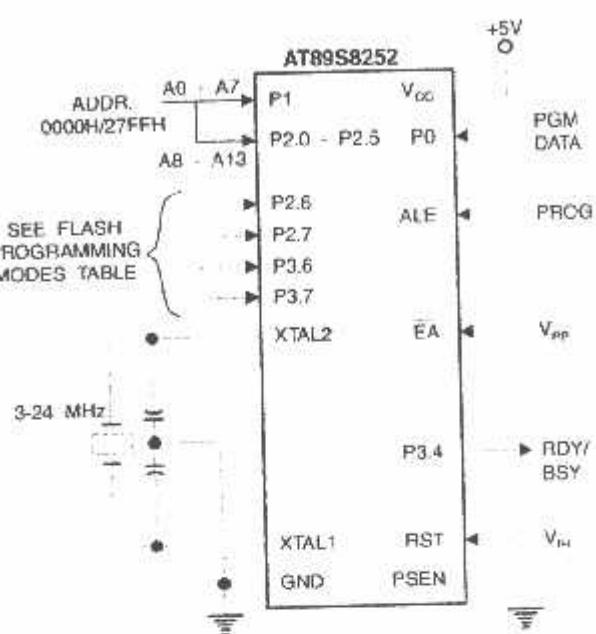


Figure 14. Verifying the Flash/EEPROM Memory

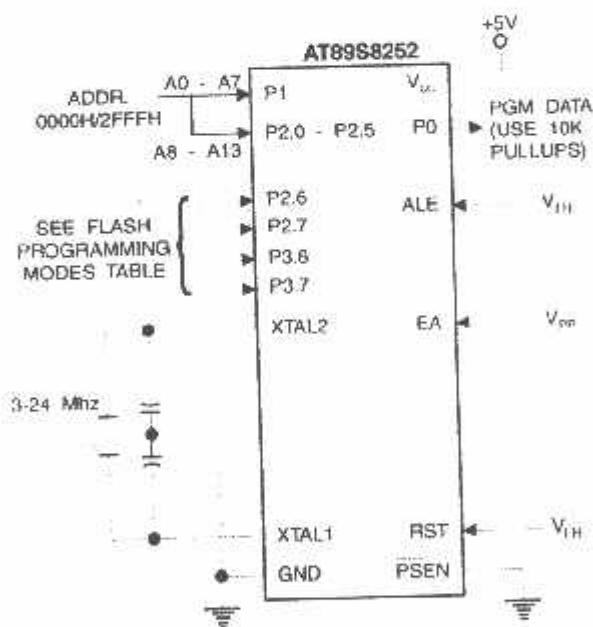
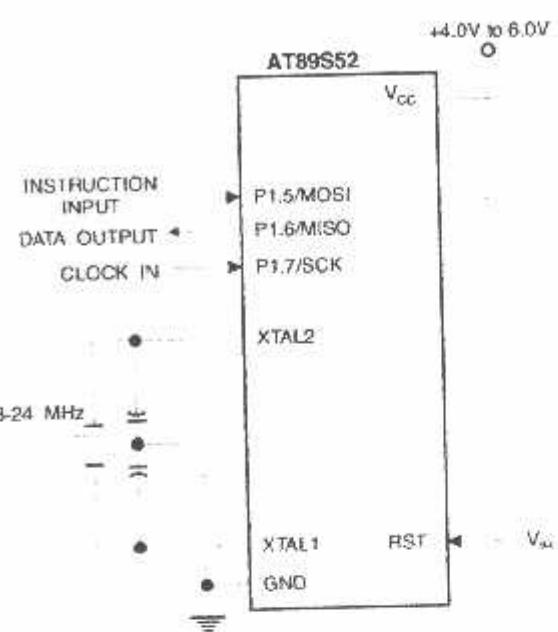


Figure 15. Flash/EEPROM Serial Downloading



Flash Programming and Verification Characteristics – Parallel Mode

 $T_A = 0^\circ\text{C}$ to 70°C , $V_{CC} = 5.0\text{V} \pm 10\%$

Symbol	Parameter	Min	Max	Units
V_{PP}	Programming Enable Voltage	11.5	12.5	V
I_{PP}	Programming Enable Current	1.0	mA	
t_{CLCL}	Oscillator Frequency	3	24	MHz
t_{AVOL}	Address Setup to PROG Low	$48t_{CLCL}$		
t_{GHAX}	Address Hold after PROG	$48t_{CLCL}$		
t_{DVOL}	Data Setup to PROG Low	$48t_{CLCL}$		
t_{GHDX}	Data Hold after PROG	$48t_{CLCL}$		
t_{EHSH}	P2.7 (ENABLE) High to V_{PP}	$48t_{CLCL}$		
t_{SHGL}	V_{PP} Setup to PROG Low	10		μs
t_{GLGH}	PROG Width	1	110	μs
t_{AVDV}	Address to Data Valid		$48t_{CLCL}$	
t_{ELOV}	ENABLE Low to Data Valid		$48t_{CLCL}$	
t_{EHQZ}	Data Float after ENABLE	0	$48t_{CLCL}$	
t_{GHBL}	PROG High to BUSY Low		1.0	μs
t_{WC}	Byte Write Cycle Time		2.0	ms

Flash/EEPROM Programming and Verification Waveforms – Parallel Mode

